

Es: $M=2$ $\mathcal{A} = \{0, 1\}$ $\mathcal{A} = \{A, B\}$

$M=3$ $\mathcal{A} = \{0, 1, * \}$

0 | * | 0 | * | 1 | 1 | 0 | ...

$M=8$

$\mathcal{A} = \{a_1, a_2, a_3, a_4, a_5, \dots, a_8\}$
 $\{000, 001, 010, \dots, 111\}$

000 | 010 | 111 | ...

BAUDRATE Freq. di simboli $\frac{1}{T_s}$ simboli/sec.

TASSO BINARIO BIT-RATE $B_2 = \frac{\log_2 M}{T_s}$ bit/sec

0 1 0 0 1 0 0 0 0 0 1 0 0 1

cont. inf. \uparrow
 diverso? \downarrow

T_s

$T_s = 1 \mu\text{sec}$

$B_2 = \frac{1}{T_s} \frac{\text{bit}}{\text{sec}} = 10^6 \text{ bit/sec}$

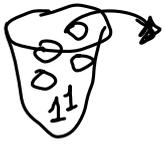
diverso? ↓

T_s

0 1 1 0 0 1 1 0 1 0

$$B_2 = \frac{1}{T_s}$$

$$s = 1 \mu s \\ \frac{\text{bit}}{s} = 10^6 \frac{\text{bit}}{s} \\ = 1 \text{ Mbit/s}$$



$$\mathcal{A} = \{a_1 a_2 \dots a_m\}$$

$$\pi = \{p_1 p_2 \dots p_m\}$$

$$0 \leq p_i \leq 1$$

$$\sum_{i=1}^m p_i = 1$$

Es. $M=3$ $\mathcal{A} = \{a_1 a_2 a_3\}$

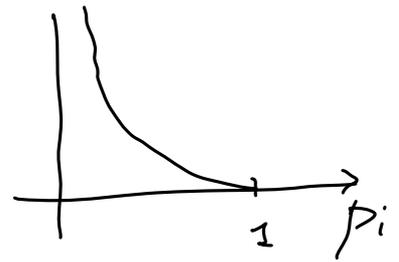
$$\pi = \{0.001, 0.5, 0.499\}$$

$$B_2 = \frac{\log_2 3}{T_s}$$

$$H(S) = E\left[\log_2 \frac{1}{p_i}\right] = \sum_{i=1}^m p_i \log_2 \frac{1}{p_i}$$

ENTROPIA bit

Autoformazione.



$$\log_3 \rightarrow \text{bit}$$

$$\ln \rightarrow \text{nat}$$

$$B_2 = \frac{\log_2 M}{T_s} \quad \text{FREQ. DI CIFRA} \quad R_s = \frac{H(S)}{T_s}$$

TASSO ENTROPICO $\frac{\text{bit}}{s}$

$$\mathcal{A} = \{a_1 \dots a_m\}$$

$$\pi = \left\{ \frac{1}{m}, \dots, \frac{1}{m} \right\}$$

$$H(S) = \sum_{i=1}^m p_i \log_2 \frac{1}{p_i} = \sum_{i=1}^m \frac{1}{m} \log_2 m = \log_2 m$$

$$B_2 = R_s$$

CODICE MORSE

E. T -
I .. M --

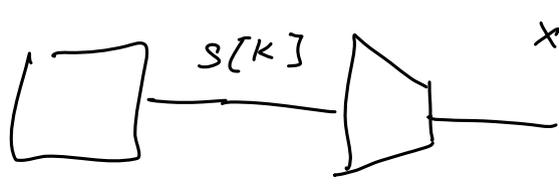
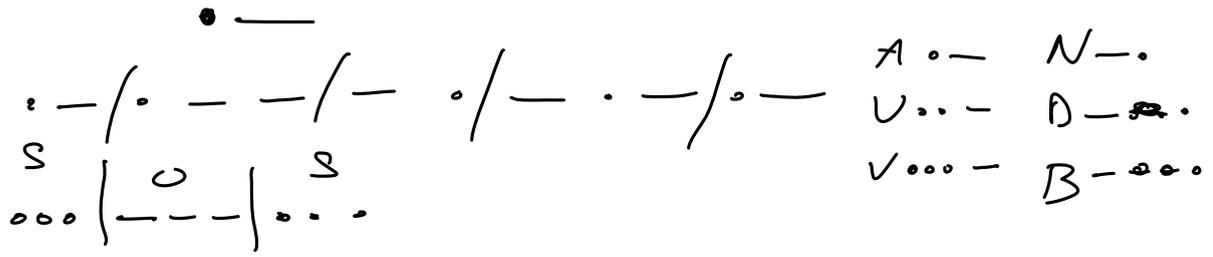
A B C D E - - - - -

S ... 0 ---

H ... C H - - - -

• —
/ / /

A — N —



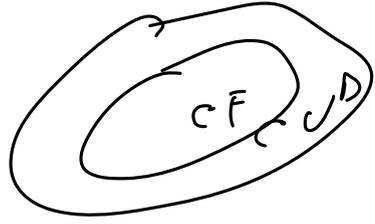
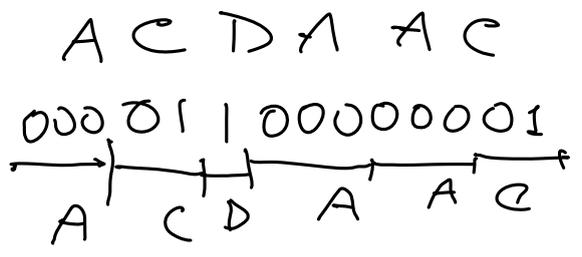
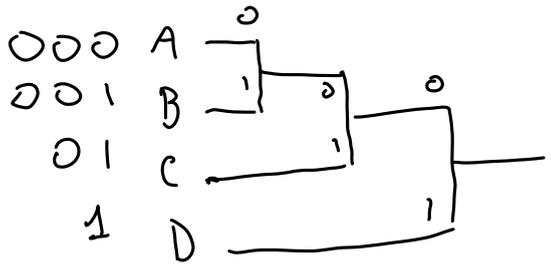
$x[k] \rightarrow$ le espressioni possibili
costituite da un solo
alfabeto base. . .

$$S[K] = \{a_1 \dots a_m\}$$

$\{0, 1\}$

- a_1 00
- a_2 011
- \vdots 0110
- a_n 011...10

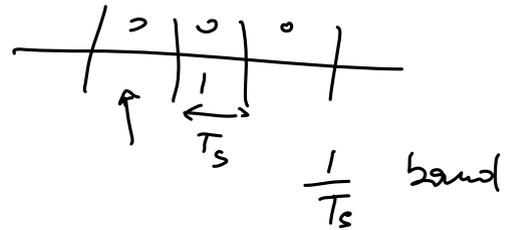
A B C D



codice A PREFISSO

$$S = \{s^{(k)} \in \{a_1, \dots, a_m\}\}$$

$$\pi = \{p_1, \dots, p_m\}$$



$$H(s) = \sum_{i=1}^n p_i \log_2 \frac{1}{p_i}$$

$$B_s = \frac{\log_2 M}{T_s} \text{ bit/sec}$$

$$R_s = \frac{H(s)}{T_s}$$

$$H(s) \leq \log_2 M$$

$$p_1 = p_2 = \dots = p_m = \frac{1}{M} \Rightarrow H(s) = \log_2 M$$

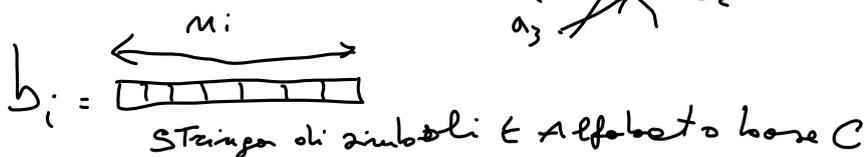
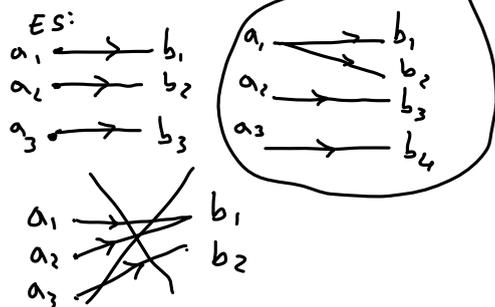
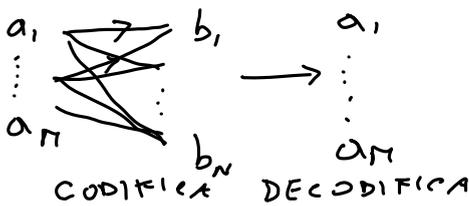
Sorgente binaria $\mathcal{A} = \{0, 1\}$ $\pi = \{0.1, 0.9\}$

$$H(s) = 0.1 \log_2 \frac{1}{0.1} + 0.9 \log_2 \frac{1}{0.9} = 0.47 \text{ bit}$$

0 0 1 0 0 1 1 1 0 1 1 1 0 1 1 0

CODICE UNIVOCAMENTE DECODIFICABILE

$$\mathcal{A} = \{a_1, \dots, a_n\} \longleftrightarrow \mathcal{B} = \{b_1, b_2, \dots, b_m\}$$



Lunghezza fissa $m_i = m \quad \forall i$
Lunghezza variabile

Lunghezza fissa $m_i = m \quad \forall i$
 Lunghezza variabile

7	0	0	0	0	1	1	1	1
6	0	0	1	1	0	0	1	1
4	3	2	1	5	0	1	0	1
0	0	0	0	NUL	DLE	SP	0	6
0	0	0	1	SOH	DC1	!	1	A
0	0	1	0	STX	DC2	"	2	B
0	0	1	1	ETX	DC3	#	3	C
0	1	0	0	EOT	DC4	\$	4	D
0	1	0	1	ENQ	NAK	%	5	E
0	1	1	0	ACK	SYN	&	6	F
0	1	1	1	BEL	ETB	'	7	G
1	0	0	0	BS	CAN	(8	H
1	0	0	1	HT	EM)	9	I
1	0	1	0	LF	SUB	*	:	J
1	0	1	1	VT	ESC	+	:	K
1	1	0	0	FF	FS	,	<	L
1	1	0	1	CR	GS	-	=	M
1	1	1	0	SO	RS	.	>	N
1	1	1	1	SI	US	/	?	O
								DEL

Tabella 3.1: Il codice ASCII

$C = \{0, 1\}$
 Lunghezza fissa $M = 128$

$m = 7$
 $A \rightarrow 1000001$
 $G \rightarrow 1110001$
 $N = 128$

↓ bit di parità

Amminoacido	Abbreviazione	Parola codice
1 Alanine	Ala	GCU, GCC, GCA, GCG
2 Arginine	Arg	AGA, AGG, CGU, CGC, CGA, CGG
3 Asparagine	Asn	AAU, AAC
4 Aspartic acid	Asp	GAU, GAC
5 Cysteine	Cys	UGU, UGC
6 Glutamine	Gln	CAA, CAG
7 Glutamic acid	Glu	GAA, GAG
8 Glycine	Gly	GCU, GGC, GGA, GGG
9 Histidine	His	CAU, CAC
10 Isoleucine	Ile	AUU, AUC, AUA
11 Leucine	Leu	UUA, UUG, CUU, CUC, CUA, CUG
12 Lysine	Lys	AAA, AAG
13 Methionine	Met	AUG
14 Phenylalanine	Phe	UUU, UUC
15 Proline	Pro	CCU, CCC, CCA, CCG
16 Serine	Ser	AGU, AGC, UCU, UCC, UCA, UCG
17 Threonine	Thr	ACU, ACC, ACA, AGC
18 Tryptophan	Trp	UGG
19 Tyrosine	Tyr	UAU, UAC
20 Valine	Val	GUU, GUC, GUA, GUG
- Term. codon	TC	UAA, UAG, UGA

Tabella 3.2: Il codice genetico

Lunghezza fissa $m = 3$
 $C = \{A, G, C, U\}$
 $4^3 = 64$
 Ans. Amm. → Stringa non univoca

A	B	A	B	A	B
A	---	1	----	----	----
B	---	2	----	----	----
C	---	3	----	----	----
D	---	4	----	----	----
E	---	5	----	----	----
F	---	6	----	----	----
G	---	7	----	----	----
H	---	8	----	----	----
I	---	9	----	----	----
J	---	0	----	----	----
K	---			Double dash (Break)	----
L	---			Error	----
M	---			/	----
N	---			End of message (AR)	----
O	---			End of Transmission (SK)	----
P	---			SOS	----
Q	---			Wait (AS)	----
R	---				----
S	---				----
T	---				----
U	---				----
V	---				----
W	---				----
X	---				----
Y	---				----
Z	---				----

Tabella 3.3: Il codice Morse

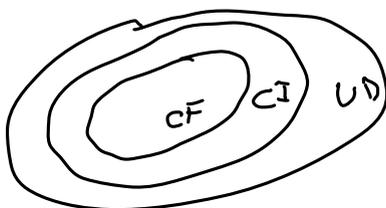
Lunghezza Variabile

$C = \{., -\}$

FEDE



UNIVOCAMENTE DECODIFICABILI



Seq. simboli di sorgente è univoca

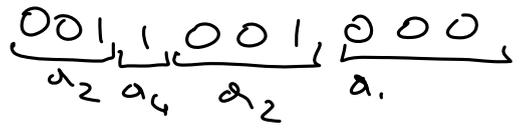
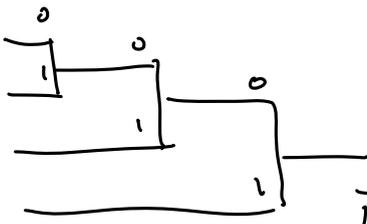
$a_1 a_2 a_1 a_1$
 $a_1 00$
 $a_2 111$
 CODICE I STANTANEO

CODICI A PREFISSO: nessuna parola codice è prefisso di un'altra

PAROLE CODICE ← TERMINAZIONI DI UN ALBERO

PAROLE CODICE \leftrightarrow TERMINAZIONI DI UN ALBERO

BINARIO
 0.1 000 a_1
 0.2 001 a_2
 0.5 01 a_3
 0.2 1 a_4

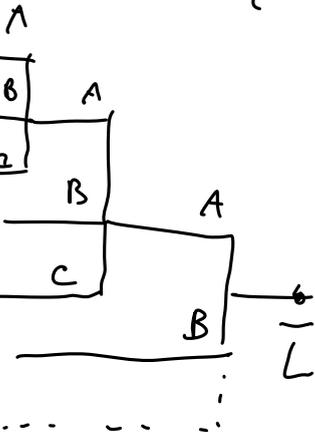


$$\bar{L} = 3 \cdot 0.1 + 3 \cdot 0.2 + 2 \cdot 0.5 + 0.2 = 2.1 \text{ bit}$$

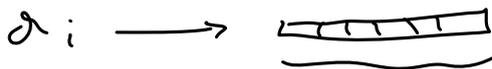
TERNARIO

$C = \{A, B, C\}$

0.1 AAA a_1
 0.2 AAB a_2
 0.5 AAC a_3
 0.1 AB a_4
 0.05 AC a_5
 0.05 B a_6



$$\bar{L} = 3 \cdot 0.1 + 3 \cdot 0.2 + 3 \cdot 0.5 + 2 \cdot 0.1 + 2 \cdot 0.05 + 0.05 = X$$



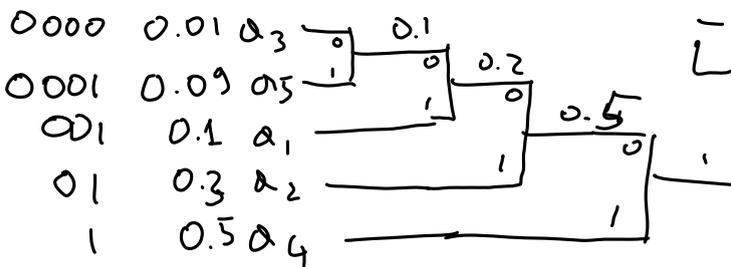
p_i l_i

$C = \{0, 1\}$

a_1, a_3, a_4, a_5
 $\underline{a} \quad \underline{a} \quad \underline{a}$

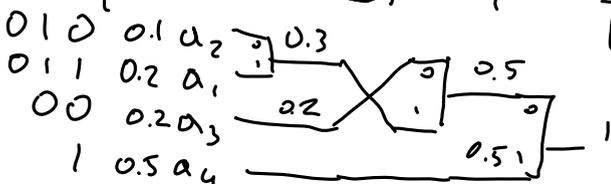
$$\bar{L} = \sum_{i=1}^n p_i l_i$$

$\{a_1, a_2, a_3, a_4, a_5\}$
 $\{0.1, 0.3, 0.01, 0.5, 0.09\}$

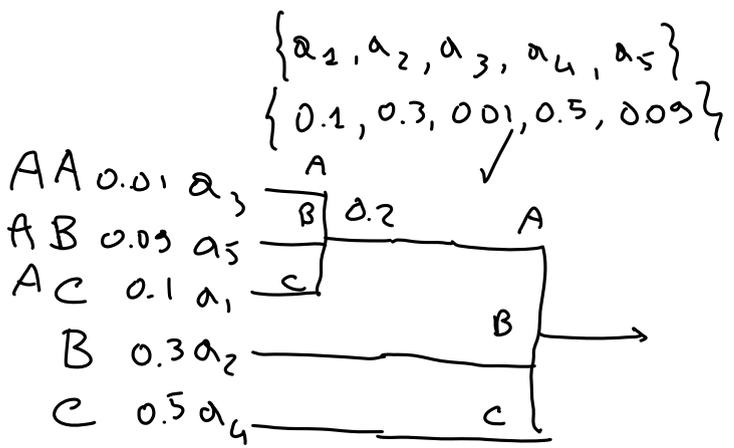


$$\bar{L} = 4 \cdot 0.01 + 4 \cdot 0.09 + 3 \cdot 0.1 + 2 \cdot 0.3 + 0.5 = X \text{ bit}$$

$\{a_1, a_2, a_3, a_4\}$
 $\{0.2, 0.1, 0.3, 0.5\}$



$\bar{L} = \dots$



l_i

Cond. nec. per C.P.

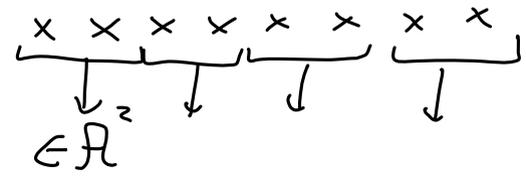
$$\sum_{i=1}^n 2^{-l_i} \leq 1$$

DIS. DI KRAFT-M. MILLMAN

3 nel caso binario



$s[k] \in \mathcal{A} = \{a_1, \dots, a_n\}$

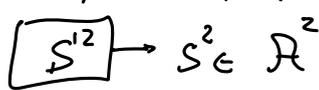


SORGENTE ESTESA

ordine 2
 $\mathcal{A} = \{a_1, a_2, a_3\}$



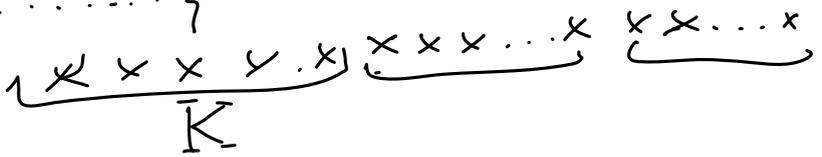
$\mathcal{A}^2 = \{a_1 a_1, a_1 a_2, a_1 a_3, a_2 a_1, a_2 a_2, a_2 a_3, a_3 a_1, a_3 a_2, a_3 a_3\}$
 $= \mathcal{A} \times \mathcal{A}$



$\mathcal{A}^3 = \mathcal{A} \times \mathcal{A} \times \mathcal{A} = \mathcal{A}^2 \times \mathcal{A}$



$= \{a_1 a_1 a_1, a_1 a_1 a_2, a_1 a_1 a_3, \dots\}$



$s^k \in \mathcal{A}^k = \underbrace{\mathcal{A} \times \mathcal{A} \times \dots \times \mathcal{A}}_k$

$\mathcal{A} = \{a_1, a_2, a_3\}$

$\Pi = \{p_1, p_2, p_3\}$

$$\mathcal{A}^2 = \{a_1 a_1, a_1 a_2, a_1 a_3, a_2 a_1, a_2 a_2, a_2 a_3, a_3 a_1, a_3 a_2, a_3 a_3\}$$

$$\Pi^2 = \{p_1^2, p_1 p_2, p_1 p_3, p_2 p_1, p_2^2, p_2 p_3, p_3 p_1, p_3 p_2, p_3^2\}$$

$$= \{p_1, p_2, p_3\} \otimes \{p_1, p_2, p_3\} = \{p_1 \{p_1 p_2 p_3\}, p_2 \{p_1 p_2 p_3\}, p_3 \{p_1 p_2 p_3\}\}$$

$$\mathcal{A} = \{a_1, a_2, \dots, a_n\} \quad n\text{-ario}$$

$$\Pi = \{p_1, p_2, \dots, p_m\}$$

$$S^K \quad s^i \in \underbrace{\mathcal{A} \times \mathcal{A} \times \dots \times \mathcal{A}}_K = \mathcal{A}^K$$

$$\Pi^K = \underbrace{\Pi \otimes \Pi \otimes \dots \otimes \Pi}_K$$

Esempio

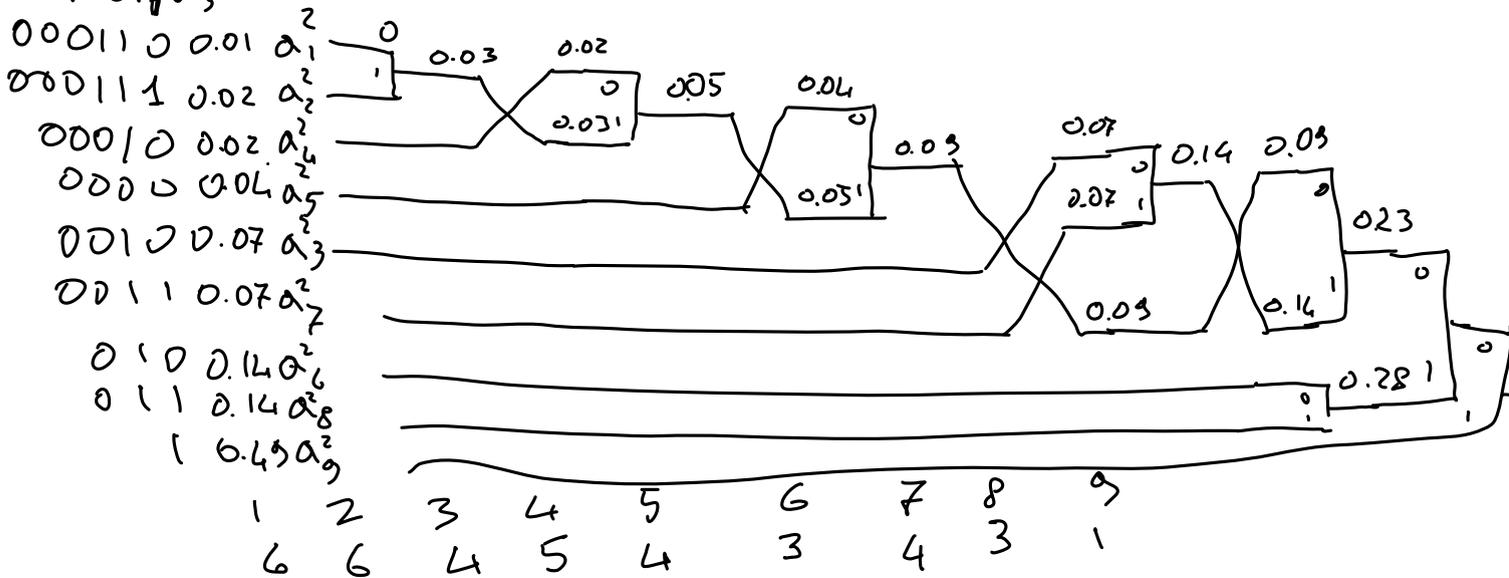
$$\mathcal{A} = \{a_1, a_2, a_3\} \quad H(\Pi) = 1.568 \text{ bit}$$

$$\Pi = \{0.1, 0.2, 0.7\}$$

$$\mathcal{A}^2 = \{a_1^2, a_1 a_2, a_1 a_3, a_2 a_1, a_2 a_2, a_2 a_3, a_3 a_1, a_3 a_2, a_3 a_3\}$$

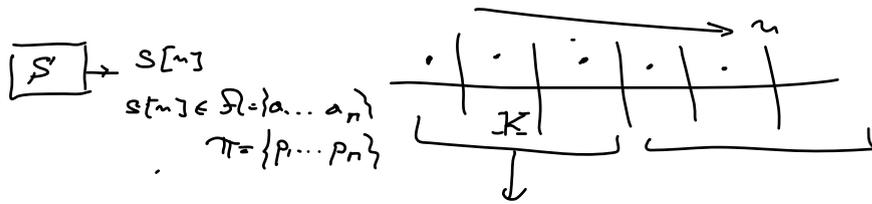
$$\Pi^2 = \Pi \otimes \Pi = \{0.0100, 0.0200, 0.0700, 0.0200, 0.0400, 0.1400, 0.0700, 0.1400, 0.4900\}$$

$$\bar{L} = 8 \cdot 0.1 + 2 \cdot 0.2 + 0.7 = 1.3 \text{ bit}$$

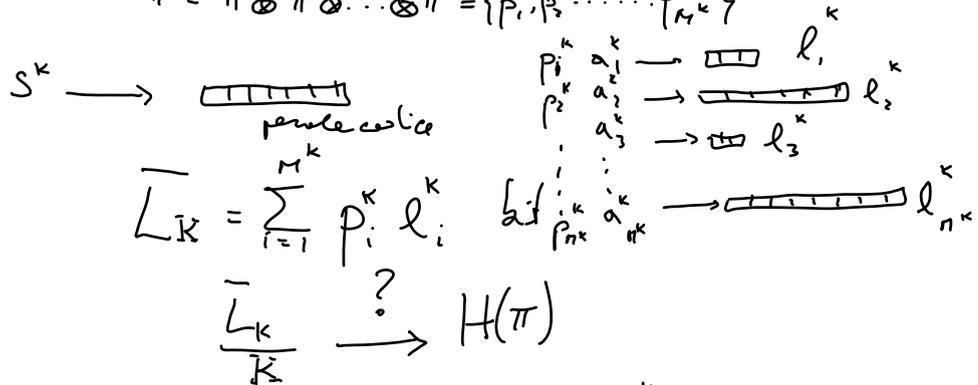


$$\bar{L} = 2.33 \text{ bit}$$

$$\frac{\bar{L}}{L_2} = 1.16 \text{ bit}$$



$S^K \rightarrow s^k \in \mathcal{R}^k = \mathcal{R} \times \mathcal{R} \times \dots \times \mathcal{R} = \{a_1^k, a_2^k, \dots, a_n^k\}$
 $\pi^k = \pi \otimes \pi \otimes \dots \otimes \pi = \{p_1^k, p_2^k, \dots, p_n^k\}$



$l_i^k = \left\lceil \log_2 \frac{1}{p_i^k} \right\rceil \quad \log_2 \frac{1}{p_i^k} < l_i^k < \log_2 \frac{1}{p_i^k} + 1$

Disuguaglianza Kraft-Sardis-Pfaff

$\sum_{i=1}^n 2^{-l_i} \leq 1$

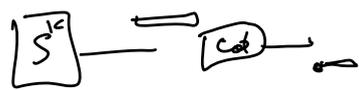
$\log_2 \frac{1}{p_i^k} < l_i^k < \log_2 \frac{1}{p_i^k} + 1$

$\log_2 \frac{1}{p_i^k} \cdot p_i^k l_i^k < p_i^k \log_2 \frac{1}{p_i^k} + p_i^k$

$\sum_{i=1}^{n^k} p_i^k \frac{1}{p_i^k} \cdot l_i^k < \sum_{i=1}^{n^k} p_i^k \log_2 \frac{1}{p_i^k} + 1$

$\frac{H(S^K)}{K} < \frac{\bar{L}_K}{K} < \frac{H(S^K)}{K} + \frac{1}{K}$

$K \rightarrow \frac{\bar{L}_K}{K} \rightarrow \frac{H(S^K)}{K} = \frac{K H(S)}{K}$



I teoremi

Esercizio Serpente binario

$\mathcal{R} = \{a, b\}$
 $\pi = \{0.1, 0.9\}$
 $H(S) = 0.1 \log_2 \frac{1}{0.1} + 0.9 \log_2 \frac{1}{0.9} = 0.4690$

$$\mathcal{R} = \{a, b\} \quad H(S) = 0.1 \log_{2} \frac{1}{0.1} + 0.9 \log_{2} \frac{1}{0.9} = 0.4690$$

$$\pi = \{0.1, 0.9\}$$

$$\bar{L}_1 = 1 \cdot 0.1 + 1 \cdot 0.9 = 1 \text{ bit}$$

$$\mathcal{R}^2 = \{aa, ab, ba, bb\} = \mathcal{R} \times \mathcal{R}$$

$$\pi^2 = \pi \otimes \pi = \{0.01, 0.09, 0.09, 0.81\}$$

$$-\log_2 p_i = \{6.64, 3.47, 3.47, 0.3\}$$

$$= \{7, 4, 4, 1\}$$

$$\bar{L}_2 = 7 \cdot 0.01 + 4 \cdot 0.09 + 4 \cdot 0.09 + 1 \cdot 0.81 = 1.6 \text{ bit}$$

$$\frac{\bar{L}_2}{2} = 0.8$$

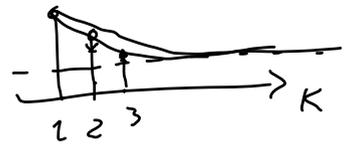
$$\mathcal{R}^3 = \mathcal{R} \times \mathcal{R}^2 = \{aaa, aab, aba, abb, baa, bab, bba, bbb\}$$

$$\pi^3 = \{0.0010, 0.0090, 0.0090, 0.0810, 0.0090, 0.0810, 0.0810, 0.7290\}$$

$$-\log_2 p_i = \{10, 7, 7, 4, 7, 4, 4, 1\}$$

$$\bar{L}_3 = 1.9$$

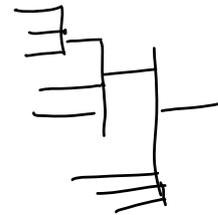
$$\frac{\bar{L}_3}{3} = 0.63$$



$$\sum_{i=1}^N 2^{-l_i} \leq 1$$



$$\sum_{i=1}^N M^{-l_i} \leq 1$$



$$\pi^K = \{p_1^K, p_2^K, \dots, p_{M^K}\}$$

$$H(S^K) = K H(S)$$

HUFFMAN →



→ CODES

a
b
c
d
e

a a
a b
a a

5

LEMPERL-ZIV

(ZIP)

$\Sigma = \{abc\}$

INPUT	OUTPUT	CODEBOOK
bcababbcbcbaaaabbc		{a = 1, b = 2, c = 3}

b|c|a|b|a|b|b|c|b|c|b|a|a|a|a|b|b|c

2|3|1|2|6|4|9|1|11|8|3

- 1 a
- 2 b
- 3 c

- 4 bc
- 5 ca
- 6 ab
- 7 ba
- 8 abb
- 9 bcb
- 10 bcba
- 11 aa

- 12 aaa
- 13 abbe

- 1 a
- 2 b
- 3 c
- 4 bc
- 5 ca
- 6 ab
- 7 ba
- 8 abb
- 9 bcb
- 10 bcba
- 11 aa
- 12 aaa

2 3 1 2 6 4 9 | 1 11 8 3

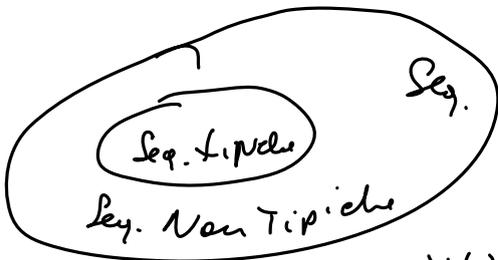
bcababbcbcbaaaabbc

SEQUENZE TIPICHE

$\mathcal{X} = \{0, 1\}$
 $\pi = \left\{ \frac{8}{10}, \frac{2}{10} \right\}$

$\left\{ \begin{array}{l} 000100110001 \\ 010100010000 \\ 001000001001 \end{array} \right. \left(\frac{8}{10} \right)^{10} \cdot \left(\frac{2}{10} \right)^3$

$\left\{ \begin{array}{l} \rightarrow 1111111111 \left(\frac{2}{10} \right)^{10} \\ \rightarrow 100111100110 \end{array} \right.$
 Prob $\rightarrow 0$



seq. tipiche $\approx 2^{mH(\pi)}$
 di lunghezza m
 (simboli indipendenti)

E.s. caso binario
 $\pi = \{p, 1-p\}$
 $H(\pi) = p \log_2 \frac{1}{p} + (1-p) \log_2 \frac{1}{1-p}$
 $p = \frac{1}{2} \Rightarrow H(\pi) = 1$
 seq. tipiche = 2^m

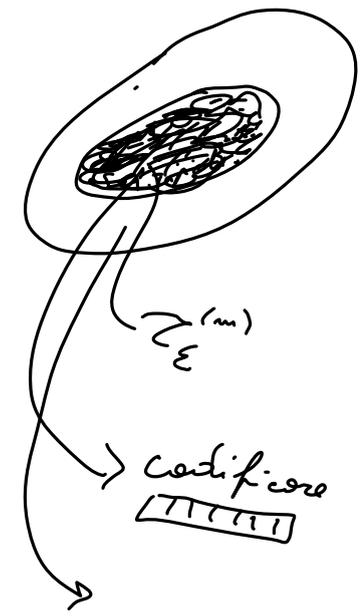
Definizione: l'insieme tipico $\mathcal{Z}_\epsilon^{(m)}$ rispetto a $p(s)$, è l'insieme delle sequenze $(s_1, s_2, \dots, s_m) \in \mathcal{X}^m$ con la proprietà che

$2^{-m(H(s)+\epsilon)} \leq p(s_1, s_2, \dots, s_m) \leq 2^{-m(H(s)-\epsilon)}$

Le sequenze sono dette ϵ -tipiche

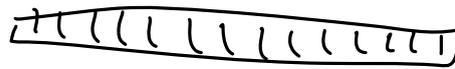
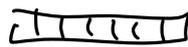
PROPRIETA' di $\mathcal{Z}_\epsilon^{(m)}$

- $H(s) - \epsilon \leq -\frac{1}{m} \log p(s_1, s_2, \dots, s_m) \leq H(s) + \epsilon$
- $P\{\mathcal{Z}_\epsilon^{(m)}\} > 1 - \epsilon$ per m sufficientemente grande
- $|\mathcal{Z}_\epsilon^{(m)}| \leq 2^{m(H(s)+\epsilon)}$ $|\mathcal{Z}_\epsilon^{(m)}|$ numero degli elementi di $\mathcal{Z}_\epsilon^{(m)}$
- $|\mathcal{Z}_\epsilon^{(m)}| \geq (1-\epsilon) 2^{m(H(s)-\epsilon)}$ per m sufficientemente grande



seq. tip. 111111

seq. tipiche
 seq. atipiche

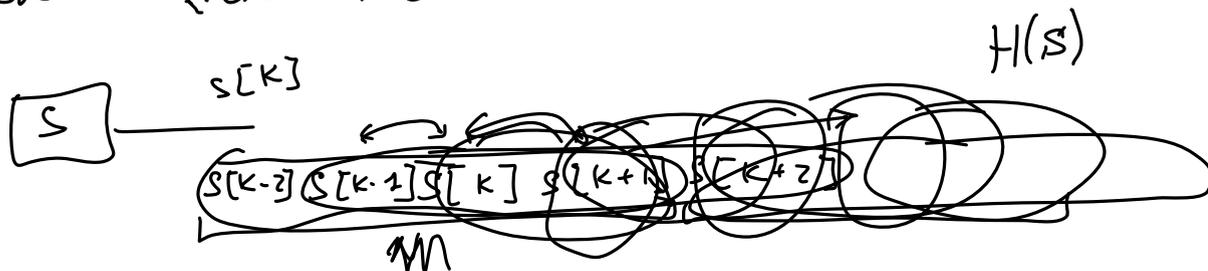


$$\bar{L} \leq m H(\pi)$$

$$\frac{\bar{L}}{m} \rightarrow H(\pi)$$



SORGENTI CON MEMORIA



$x_1, x_2, x_3, x_4, x_5, x_6$

$P(x_1, x_2, x_3, x_4, x_5, x_6)$
 $P_{x_1, x_2, x_3, x_4, x_5, x_6}$

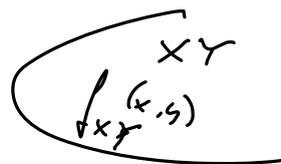
Seq. di ordine m



$$S^m = (S[K], S[K+1], \dots, S[K+m-1])$$

$$P_{S^m}(s_0, s_1, \dots, s_{m-1})$$

$$H(S^m) = \sum_{\substack{s_0, s_1, \dots, s_{m-1} \\ \in A^m}} P_{S^m}(s_0, \dots, s_{m-1}) \log_2 \frac{1}{P_{S^m}(s_0, \dots, s_{m-1})} \text{ bit.}$$



$$H(S) = \lim_{m \rightarrow \infty} \frac{H(S^m)}{m}$$

TASSO ENTROPICO

$$\frac{m H(S)}{m}$$

$S[1], S[2], \dots, S[n]$

$$H_\infty(S) = \lim_{n \rightarrow \infty} H(S[n] | S[n-1], \dots, S[1]) \text{ bit}$$

Seq. stazionarie $\Rightarrow H_\infty(S) = H(S)$

$$\boxed{\text{Seq. stationary} \Rightarrow H_\infty(S) = H(S)}$$

$$H(Y|X) \triangleq \sum_{(x,y) \in X \times Y} p(x,y) \log_2 \frac{1}{p(y|x)}$$

$(p(y|x)p(x))$

$$\begin{array}{l} x \in X \quad y \in Y \\ p(y|x) \\ p(x) \\ p(x,y) \end{array}$$

$$H(Y|X) \leq H(Y)$$

$$\begin{array}{l} H(S[2]|S[1]) \\ \leq H(S[2]) \end{array}$$

REGOLA A CATENA

$$p(x_1, x_2) = p(x_2|x_1)p(x_1)$$

$$p(x_1, x_2, x_3) = p(x_3|x_1, x_2)p(x_1, x_2) = p(x_3|x_1, x_2)p(x_2|x_1)p(x_1)$$

$$H(x_1, x_2, x_3) = H(x_3|x_1, x_2) + H(x_2|x_1) + H(x_1)$$

$$\underbrace{H(S[1] \dots S[m])}_m = \frac{1}{m} \sum_{i=1}^m \underbrace{H(S[i]|S[1] \dots S[i-1])}_{\text{eq. non costante}}$$

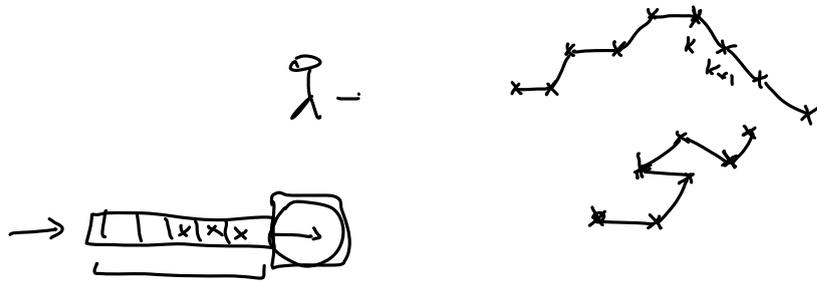
lim
 $m \rightarrow \infty$

$\longrightarrow H_\infty(S)$

$H_\infty(S)$

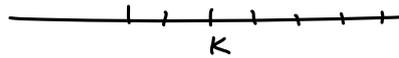
$$\lim_{m \rightarrow \infty} \frac{H(S[1] \dots S[m])}{m} = H_\infty(S) = \lim_{m \rightarrow \infty} H(S[m]|S[1] \dots S[m-1])$$

PROCESSI DI MARKOV



$$S(k+1) = f(S(k), u(k)) :$$

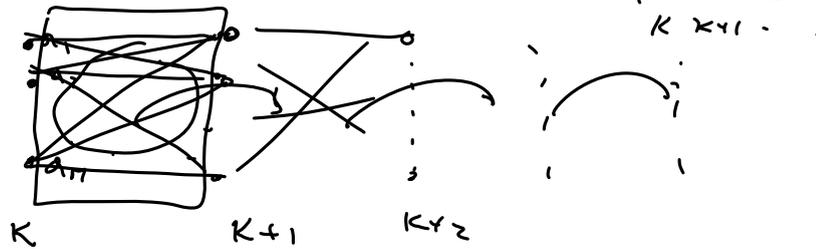
$\{S(k)\}$



$$P_2 \left\{ S[k] = s_k \mid S[k-1] = s_{k-1}, S[k-2] = s_{k-2}, S[k-3] = s_{k-3}, \dots \right\}$$

$$= P_2 \left\{ S[k] = s_k \mid S[k-1] = s_{k-1} \right\} \Rightarrow \text{Processo di Markov.}$$

$$S[k] \in \mathcal{X} = \{a_1, \dots, a_m\}$$



TRELLIS (TRAZIOCCIO)

$$P_2 \{S[k+1] \mid S[k]\} = P_2 \{S[k+2] \mid S[k+1]\} = P_2 \{S[k+m] \mid S[k+m-1]\}$$

STAZIONARIA O OMOGENEA $\forall m$

$$\begin{pmatrix} a_1 & a_2 & \dots & a_m \\ \cdot & \cdot & \cdot & \cdot \\ a_2 & \cdot & \cdot & \cdot \\ \cdot & \cdot & \cdot & \cdot \\ a_m & \cdot & \cdot & \cdot \end{pmatrix} = \begin{bmatrix} P\{S[k+1]=a_1 \mid S[k]=a_1\} & P\{S[k+1]=a_2 \mid S[k]=a_1\} & \dots & P\{S[k+1]=a_m \mid S[k]=a_1\} \\ P\{S[k+1]=a_1 \mid S[k]=a_2\} & P\{S[k+1]=a_2 \mid S[k]=a_2\} & \dots & \dots \\ \dots & \dots & \dots & \dots \end{bmatrix}$$

$X[k+1]$

$$\underbrace{X(k)}_D \begin{bmatrix} p(1|1) & p(2|1) & \dots & p(m|1) \\ p(1|2) & p(2|2) & \dots & p(m|2) \\ \dots & \dots & \dots & \dots \end{bmatrix} > 1$$

EQU-
valori > 1.

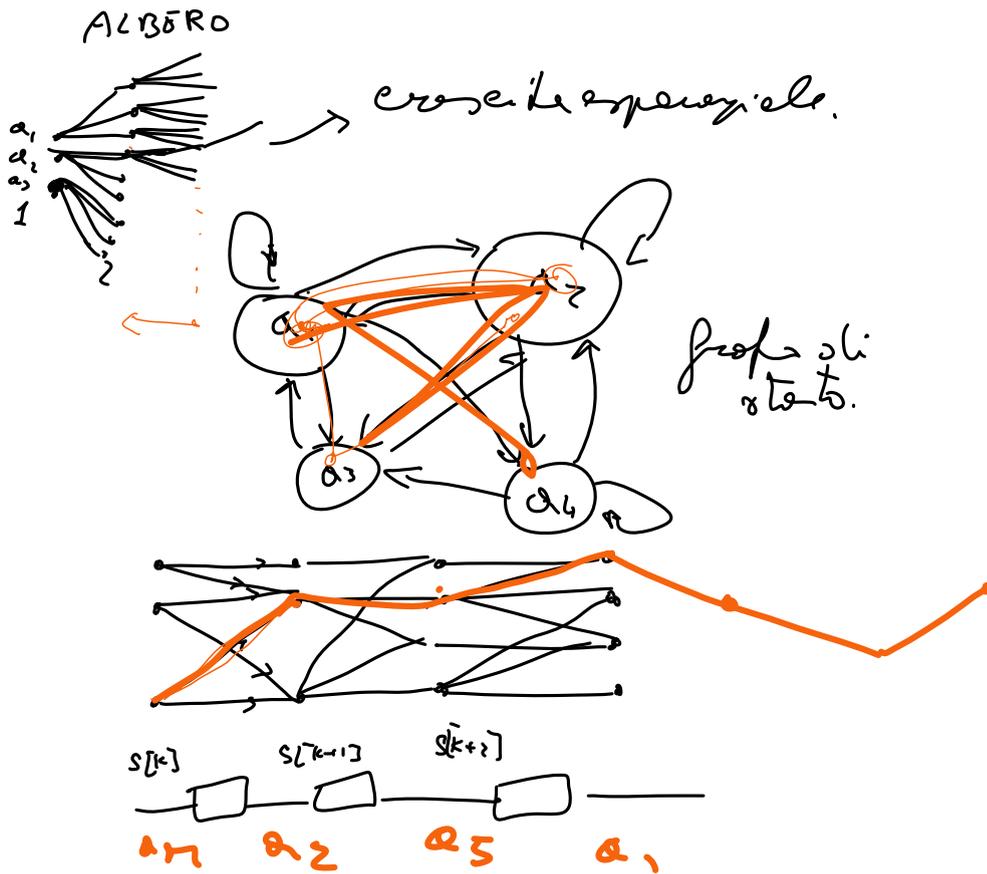
$$P = \begin{bmatrix} p(1|1) & p(2|1) & p(3|1) \\ p(1|2) & p(2|2) & p(3|2) \\ p(1|3) & p(2|3) & p(3|3) \end{bmatrix}$$

row-stochastic

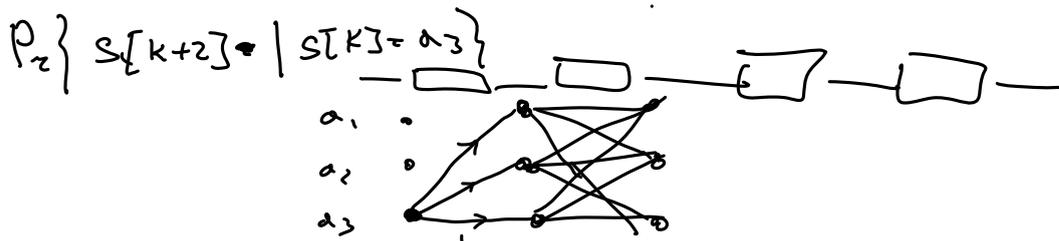
① 2 3 ... k ...

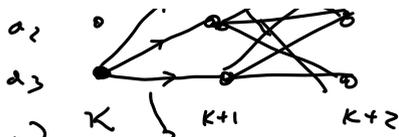
$$P\{X[2]\} = P \begin{bmatrix} P_2\{X[2]=a_1\} \\ P_2\{X[2]=a_2\} \\ P_2\{X[2]=a_3\} \end{bmatrix}$$

POSSIBILI RAPPRESENTAZIONI



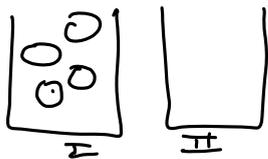
CATENA ERGODICA o IRRIDUCIBILE





$$P_2 \{ S[k+2] | S[k] \} = P^2 p(S[k])$$

$$P_2 \{ S[k+m] | S[k] \} = \left(P^2 \right)^m p(S[k])$$



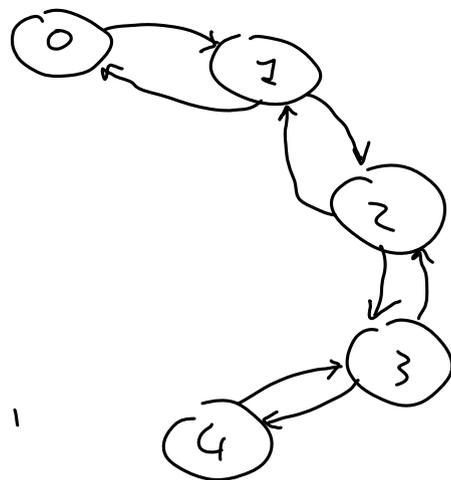
4 palle 2 Urne

ad ogni passo prendo una palla a caso e la metto nell'altra urna

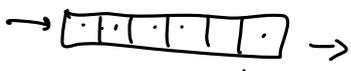
$S[k] = \#$ palle nella urna I



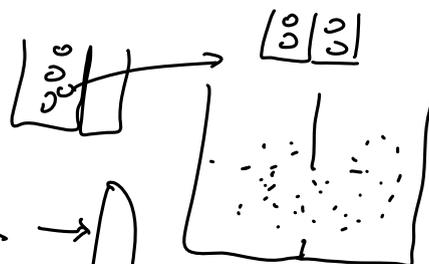
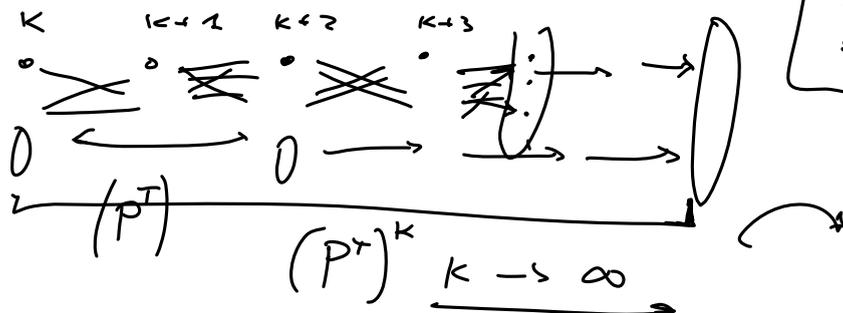
$S[k+1]$		0	1	2	3	4
$S[k]$		0	1	2	3	4
0		0	1	0	0	0
1		1/4	0	3/4	0	0
2		0	1/2	0	1/2	0
3		0	0	3/4	0	1/4
4		0	0	0	1	0



BUFFER



$$S[k] = \{0, 1, 2, 3, \dots, m\}$$



$$P^k \rightarrow \begin{bmatrix} \vdots \\ \vdots \\ \vdots \\ \vdots \\ \vdots \end{bmatrix}$$

$$P^k = (P^T)^k P^k$$

$$P \rightarrow \begin{bmatrix} \epsilon_1 & \epsilon_2 & \dots & \epsilon_n \end{bmatrix}$$

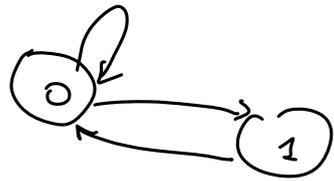
$$P^{-1} = P$$

$P[\infty] = P^T P[\infty]$

$$\left. \begin{array}{l} \pi = P^T \pi \\ \pi \text{ stochastico.} \end{array} \right\}$$

$$\pi = \begin{bmatrix} \pi_1 \\ \pi_2 \\ \vdots \\ \pi_n \end{bmatrix} \quad \sum_{i=1}^n \pi_i = 1$$

$\left. \begin{array}{l} | \quad \square \quad | \\ \pi \end{array} \right\}$

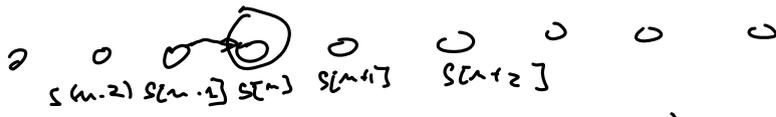


$$P = \begin{bmatrix} 0.1 & 0.9 \\ 1 & 0 \end{bmatrix}$$

001001010

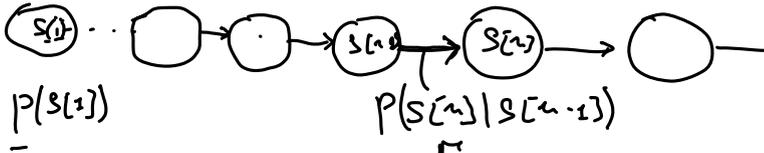
$$0 \rightarrow \rightarrow \rightarrow 1$$

$$[\cdot] = (P^T)^3 \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$



$$p(s^{(n)} | s^{(n-1)} s^{(n-2)} \dots s^{(1)})$$

$$= p(s^{(n)} | s^{(n-1)})$$

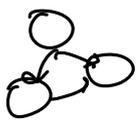


$$p(s^{(1)})$$

$$p(s^{(n)} | s^{(n-1)})$$



$$p(s^{(n)}) = (P^T)^{n-1} p(s^{(1)})$$



Esposizione

Rappresentazione

$$p(\infty) = \lim_{k \rightarrow \infty} (P^T)^k$$

$$p(\infty) = P^T p(\infty)$$

$$H_{\infty}(S) = \lim_{n \rightarrow \infty} H(s^{(n)} | s^{(n-1)} s^{(n-2)} \dots s^{(1)})$$

$$\lim_{k \rightarrow \infty} \frac{H(s^k)}{k} = H_{\infty}(S) \quad \text{se } \{s^{(n)}\} \text{ erg.$$

CATENA DI MARKOV

$$H_{\infty}(S) = \lim_{n \rightarrow \infty} H(s^{(n)} | s^{(n-1)})$$

$$= H(s^{(n)} | s^{(n-1)}) \quad \forall n > n_0$$

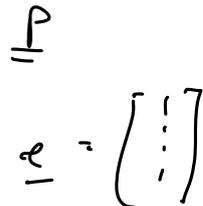
$$H(X|Y) = \sum_{(x,y)} \frac{p(x,y)p(y)}{p(x,y)} \log_2 \frac{1}{p(x,y)} = E \left[\log_2 \frac{1}{p(X|Y)} \right]$$

$$s^{(n)} \in \{a, \dots, o, n\} \subset \mathcal{X}^2$$

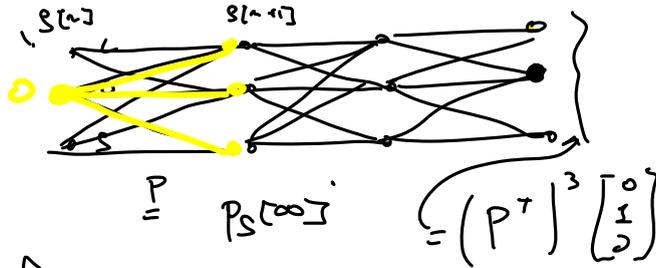
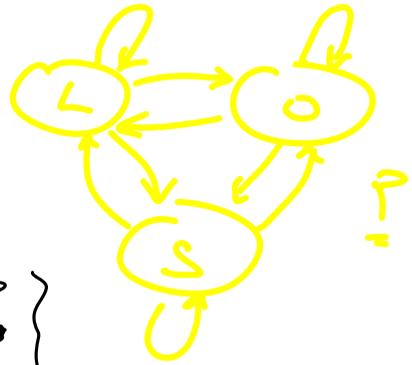
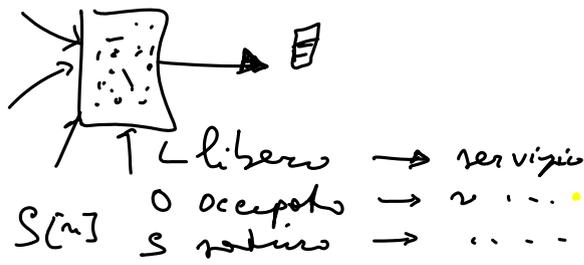
$$H(s^{(n)} | s^{(n-1)}) = \sum_{(s^{(n)}, s^{(n-1)}) \in \mathcal{X}^2} \frac{p(s^{(n)} | s^{(n-1)}) p(s^{(n-1)})}{p(s^{(n-1)})} \log_2 \frac{1}{p(s^{(n)} | s^{(n-1)})}$$

=

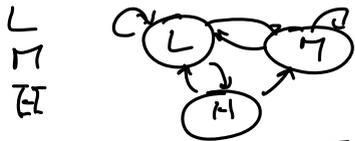
$$H_{\infty}(S) = - p^{T[\infty]} (P \circ \log_2 P) e \quad \text{bit}$$



CATENE DI MARKOV NASCOSTE

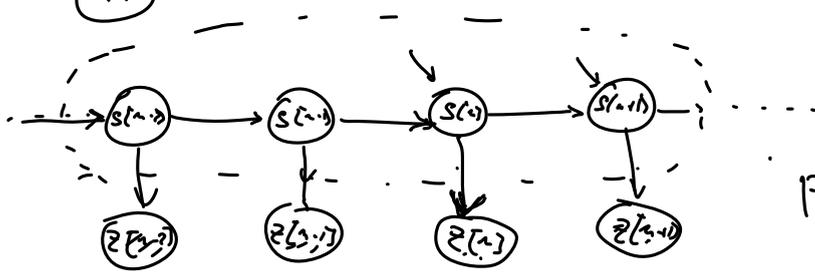


Temperatura del Processore



$$P = (P^T)^{10} \begin{bmatrix} 1/2 \\ 1/2 \\ 0 \end{bmatrix}$$

HMM
HIDDEN MARKOV MODEL

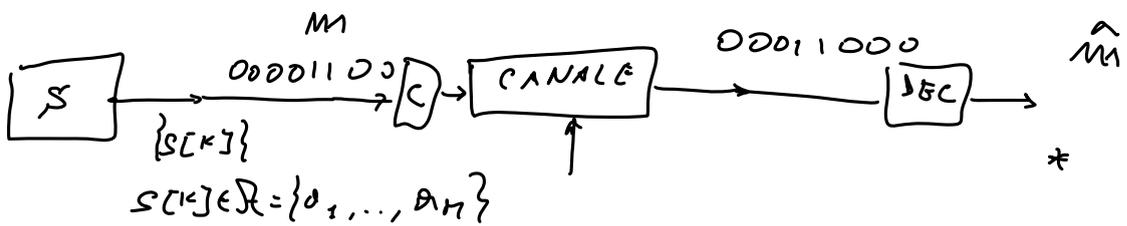


$$Z[n] = \{z_1, \dots, z_n\}$$

$$= \{b_1, \dots, b_n\}$$

$$P(z/s) = \{P_i\}_{z=b_i/s=o_j}$$

$$\begin{cases} s[n+1] = f(s[n], w[n]) \\ z[n] = g(s[n]) \end{cases}$$



RIDONDANZA



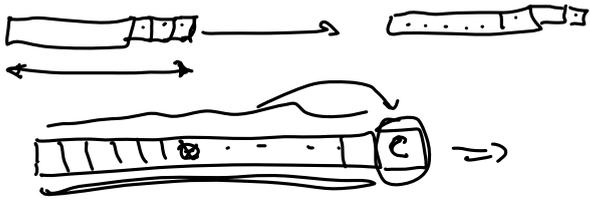
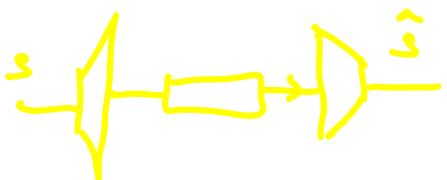
Trasmissione	0	0	1	0	1
Ricevuta	0	1	0	0	1

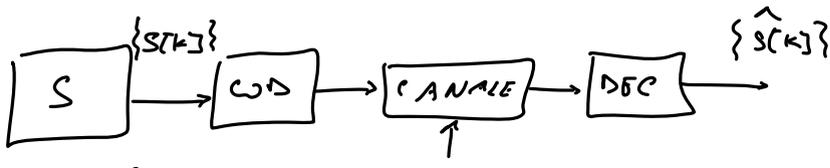
CODICE A RIPETIZIONE

Ricevuta	0000001111	000111	000111
Trasmissione	0000001111	000111	000111

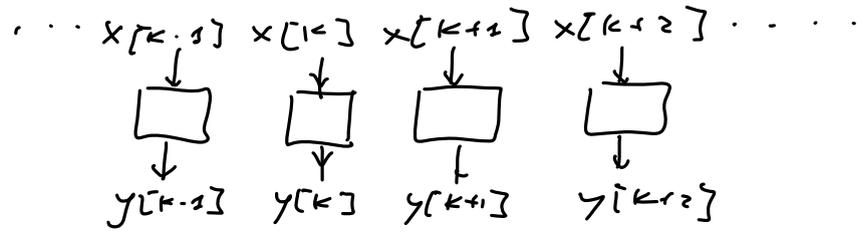
RIPETITIVO

	Körner		1		1		1		1	
	000	000	111	000	111	000	111	000	111	111
Pro.	010	000	011	000	000	100				
	0	0	1	0	0	0				

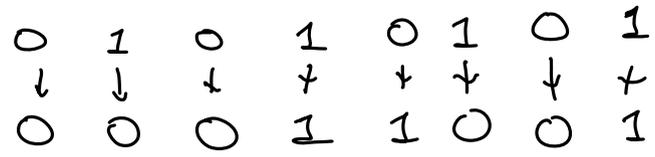




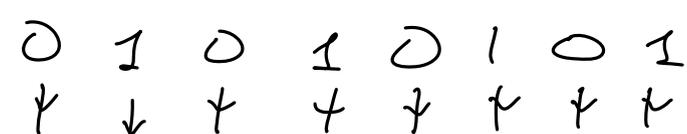
Semplificazioni =



Esempio



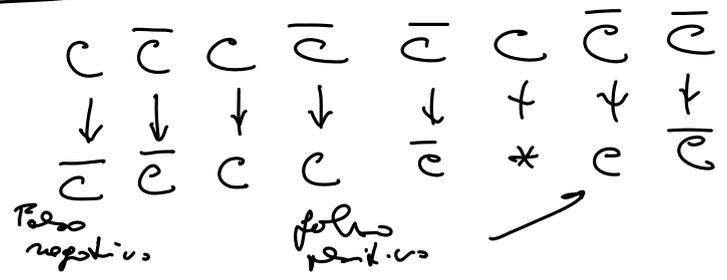
$X = \{0, 1\}$



$Y = \{0, 1, * \}$

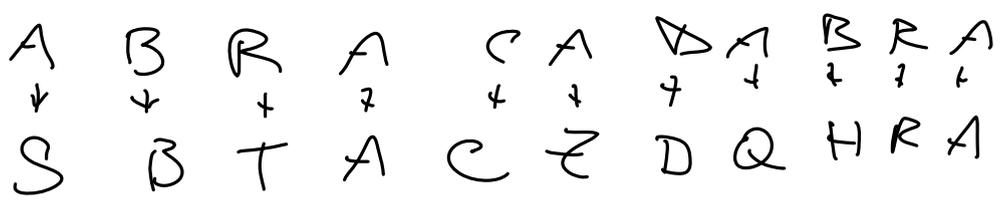
TEST COVID:

$X = \{c, \bar{c}\}$



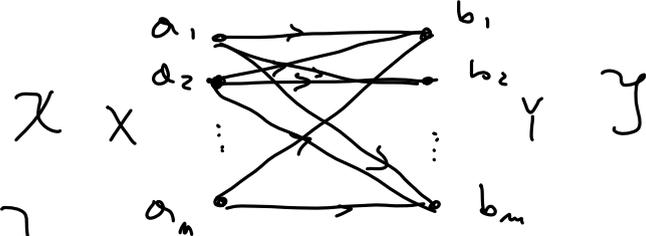
$Y = \{c, \bar{c}, * \}$

Testo
 $X = \{ \text{-----} \}$
 128 caratteri.



$Y = \{ \text{-----} \}$
 128 caratteri.

CANALE

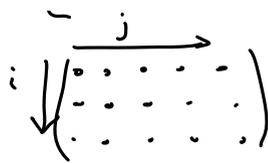


$\left[\begin{matrix} p_1 \\ p_2 \\ \vdots \\ p_n \end{matrix} \right] X = \{ a_i \}$

$$P_e = \left[\begin{matrix} p_1 \\ p_2 \\ \vdots \\ p_n \end{matrix} \right] Y = \{ b_j | X = a_i \}$$

$i: 1:n$
 $j: 1:m$

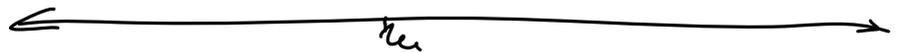
$$\left[\begin{matrix} P_{11} \\ \vdots \\ P_{1n} \end{matrix} \right] X = a_n$$



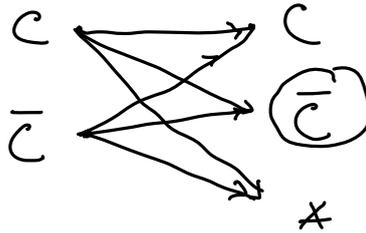
$j = 1, \dots, m$

MATRICE DI CANALE

$$P_c = \begin{bmatrix} P_1 \{Y=b_1 | X=a_1\} & P_1 \{Y=b_2 | X=a_1\} & \dots & P_1 \{Y=b_m | X=a_1\} \\ P_2 \{Y=b_1 | X=a_2\} & P_2 \{Y=b_2 | X=a_2\} & \dots & P_2 \{Y=b_m | X=a_2\} \\ \vdots & \vdots & \ddots & \vdots \\ P_n \{Y=b_1 | X=a_n\} & \dots & \dots & P_n \{Y=b_m | X=a_n\} \end{bmatrix}$$



COVID
 $X = \{c, \bar{c}\}$
 $Y = \{c, \bar{c}, * \}$



$$P_c = \begin{bmatrix} P(c|c) & P(\bar{c}|c) & P(*|c) \\ P(c|\bar{c}) & P(\bar{c}|\bar{c}) & P(*|\bar{c}) \end{bmatrix}$$

$$= \begin{bmatrix} 0.9 & 0.08 & 0.02 \\ 0.2 & 0.7 & 0.1 \end{bmatrix}$$

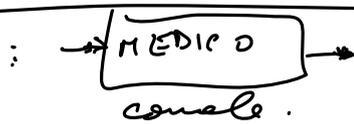
$$P_2 \{Y=c, X=c\} ?$$

$$P_2 \{Y=\bar{c}, X=\bar{c}\} ?$$

$$P_1 \{Y=* \} = ?$$

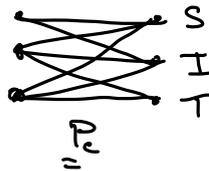
$$P_1 \{Y=c, X=\bar{c}\} ?$$

$$P(Y=c, X=c) = P(Y=c | X=c) P\{X=c\}$$



$P(S) = 0.4$
 $P(I) = 0.4$
 $P(T) = 0.2$

S STRESS
 I IPERTENSIONE
 T TUMORE

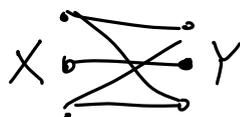


	S	I	T
S	0.5	0.5	0
I	0.19	0.8	0.01
T	0.2	0.2	0.4

$X = \{a_1, \dots, a_n\}$

$Y = \{b_1, \dots, b_m\}$

$$\pi_x = \begin{bmatrix} P_1 \{X=a_1\} \\ \vdots \\ P_n \{X=a_n\} \end{bmatrix}$$



$$P_c = [P_1 \{Y=b_j | X=a_i\}]_{i,j=1, \dots, m}$$

$$[P\{X=a_i\}] \quad P_c = [P\{Y=b_j | X=a_i\}]_{i=1..n}^{j=1..m}$$

$$P\{Y=b_j\} = \sum_{i=1}^n P\{Y=b_j | X=a_i\} P\{X=a_i\}$$

$$= \sum_{i=1}^n P\{Y=b_j, X=a_i\}$$

$$\pi_y = \begin{bmatrix} P\{Y=b_1\} \\ P\{Y=b_2\} \\ \vdots \\ P\{Y=b_m\} \end{bmatrix} = P_c^T \pi_x \quad \pi_y = P_c^T \pi_x$$

$$P\{X=a_i | Y=b_j\} = \frac{P\{X=a_i, Y=b_j\}}{P\{Y=b_j\}}$$

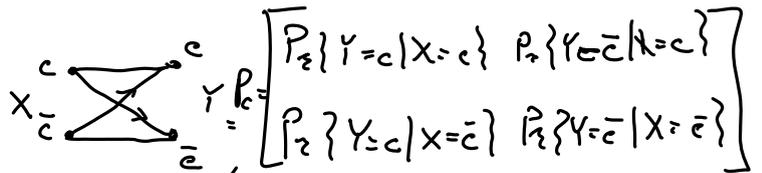
PROB. A POSTERIORI

$$= \frac{P(Y=b_j | X=a_i) P(X=a_i)}{\sum_{e=1}^n P(Y=b_j | X=a_e) P(X=a_e)}$$

T. DI BAYES

ESEMPIO

$$\pi_x = \begin{bmatrix} P\{X=c\} \\ P\{X=\bar{c}\} \end{bmatrix} = \begin{bmatrix} 0.3 \\ 0.7 \end{bmatrix}$$



$$\pi_y = \begin{bmatrix} P\{Y=c\} \\ P\{Y=\bar{c}\} \end{bmatrix} = \begin{bmatrix} 0.7 & 0.3 \\ 0.35 & 0.65 \end{bmatrix} \begin{bmatrix} 0.3 \\ 0.7 \end{bmatrix} = \begin{bmatrix} 0.455 \\ 0.545 \end{bmatrix} = \begin{bmatrix} 0.7 & 0.3 \\ 0.35 & 0.65 \end{bmatrix}$$

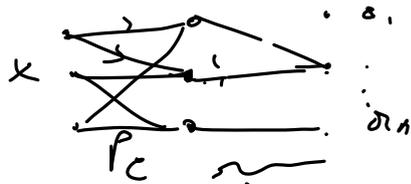
$$P\{X=c | Y=\bar{c}\} = \frac{P\{Y=\bar{c} | X=c\} P(X=c)}{P(Y=\bar{c})} = \frac{0.3 \cdot 0.3}{0.545} = 0.16$$

$$P\{X=\bar{c} | Y=\bar{c}\} = 0.84$$

$$P = \begin{bmatrix} P(X=a_i | Y=b_j) \\ \text{prob. a posteriori} \end{bmatrix}$$

MASSIMA PROB. A POSTERIORI

$$\hat{a}_e = \underset{i=1..n}{\text{argmax}} P\{X=a_i | Y=b_j\} \quad \text{MAP}$$



Esempio

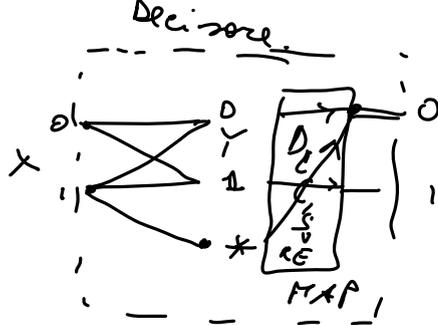
$$\mathcal{X} = \{0, 1\}$$

$$\mathcal{Y} = \{0, 1, * \}$$

$$\pi_X = \begin{bmatrix} 0.8 \\ 0.2 \end{bmatrix}$$

$$\rightarrow 00001000110000$$

$$P_C = \begin{bmatrix} 0.8 & 0.1 & 0.1 \\ 0.1 & 0.8 & 0.1 \end{bmatrix}$$



$$P = \begin{bmatrix} P(X=0|Y=0) & P(X=0|Y=1) & P(X=0|Y=*) \\ P(X=1|Y=0) & P(X=1|Y=1) & P(X=1|Y=*) \end{bmatrix}$$

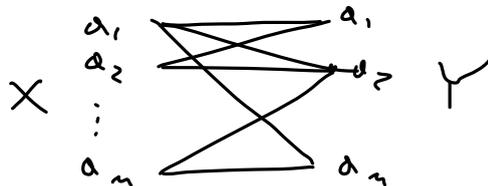
$$= \begin{bmatrix} 0.9667 & 0.3333 & 0.8000 \\ 0.0333 & 0.6667 & 0.2000 \end{bmatrix}$$

PROB. DI ERRORI

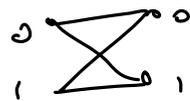
$$M = M$$

$$\mathcal{X} = \{a_1 \dots a_n\}$$

$$\mathcal{Y} = \{a_1 \dots a_n\}$$



$$\begin{matrix} 0 & 1 & 1 & 0 & 0 & 1 & 0 \\ \downarrow & \downarrow & \downarrow & \downarrow & \downarrow & \downarrow & \downarrow \\ 0 & 1 & 0 & 0 & 1 & 0 & 0 \end{matrix}$$



ABRACAD

ANRECA D

$$P_{\{error\}} = P_{\{X \neq Y\}} = \sum_{\substack{i \neq j \\ i, j=1 \\ i, j=n}}^n P_{\{X=a_i, Y=a_j\}}$$

$$= 1 - P_{\{non\ error\}} = 1 - \sum_{i=1}^n P_{\{X=a_i, Y=a_i\}}$$

$$= 1 - \sum_{i=1}^n P\{y=a_i | x=a_i\} P\{x=a_i\}$$

$$= 1 - (\text{diag } P_c)^T \pi_x$$

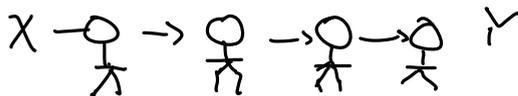
Esercizio

$$\pi_x = \begin{bmatrix} 0.8 \\ 0.2 \end{bmatrix} \quad \times \quad \begin{array}{c} \diagdown \\ \diagup \end{array} \quad Y$$

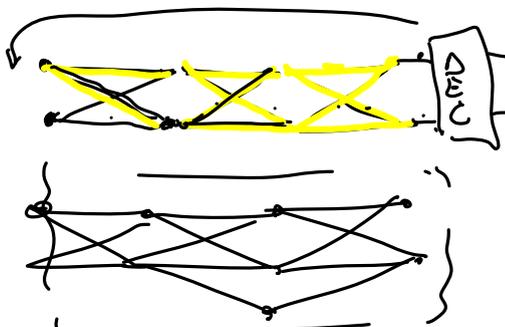
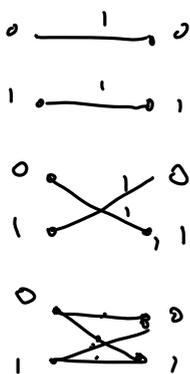
$$P_c = \begin{bmatrix} 0.8 & 0.2 \\ 0.1 & 0.9 \end{bmatrix}$$

$$P_e = 1 - [0.8 \ 0.2] \begin{bmatrix} 0.8 \\ 0.2 \end{bmatrix} = 0.18$$

CANALI IN CASCATA

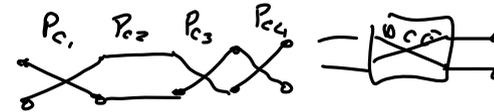


$$P_{ct} = P_{c1} P_{c2} \dots P_{cL}$$

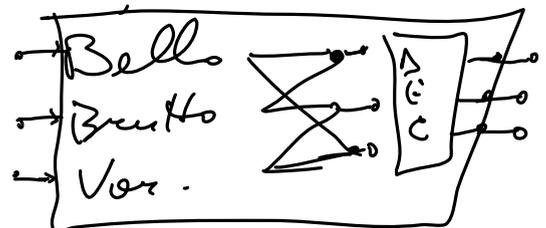
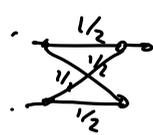


$$\pi_y = P_{c1}^T \pi_x =$$

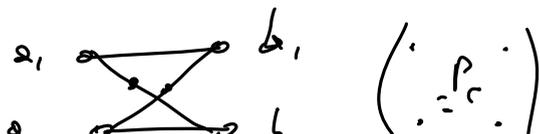
$$= P_{cL}^T P_{cL-1}^T \dots P_{c2}^T \pi_x$$

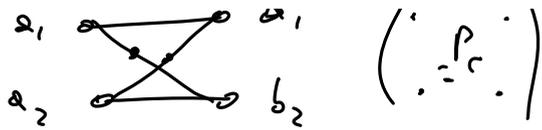


$$\begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix} = \begin{pmatrix} 0 & 1 \\ 1 & 0 \end{pmatrix}$$



CANALE BINARIO

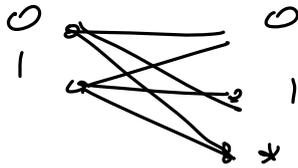




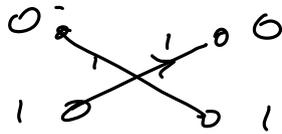
CANALE BINARIO SIMMETRICO

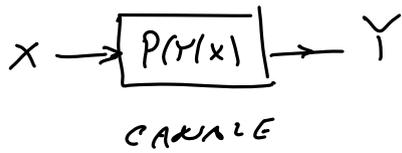
$$P_c = \begin{pmatrix} 1-p_e & p_e \\ p_e & 1-p_e \end{pmatrix} \quad \text{BSC}$$

CANALE CON CANCELLAZIONE



CANALE SEMPRE IN ERRORE





$$I(X;Y) = H(X) - H(X|Y)$$

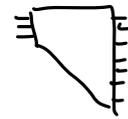
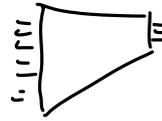
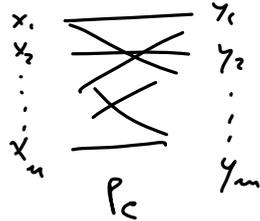
$$P(x), P(y|x)$$

$$C = \max_{P(x)} I(X;Y)$$

CAPACITÀ DI CANALE [bit]



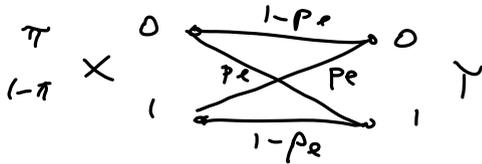
$C \geq 0$



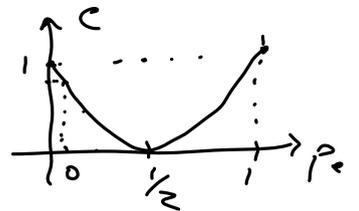
$$C \leq \log_2 m$$

$$C \leq \log_2 n$$

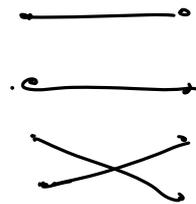
Esempio



$$C = 1 - H(p_e)$$

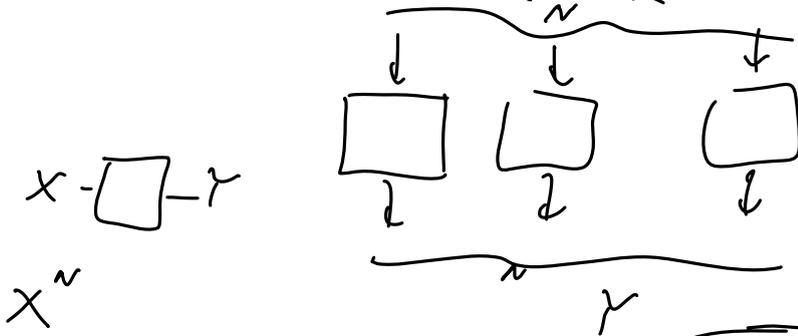
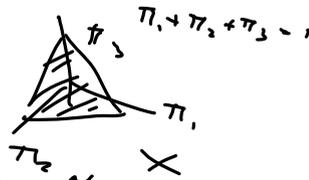


$p_e = 0.1$ $C \approx 0.53$



$$P_c, \pi_x \Rightarrow C$$

ARIMOTO/BLAUTH



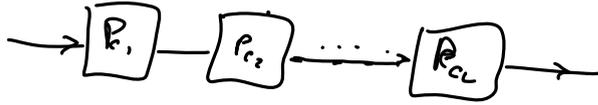
Concatenazione

$$P_c^N = \underbrace{P_c \otimes P_c \otimes \dots \otimes P_c}_N$$

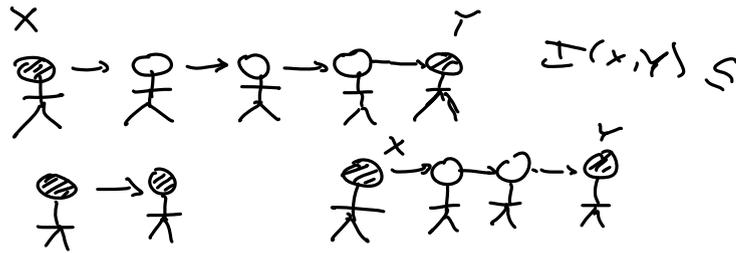
$$I(X^N; Y^N) = N I(X; Y)$$

$C^N = N C$





TEOREMA DEL DATA PROCESSING
TRATTAMENTO DATI



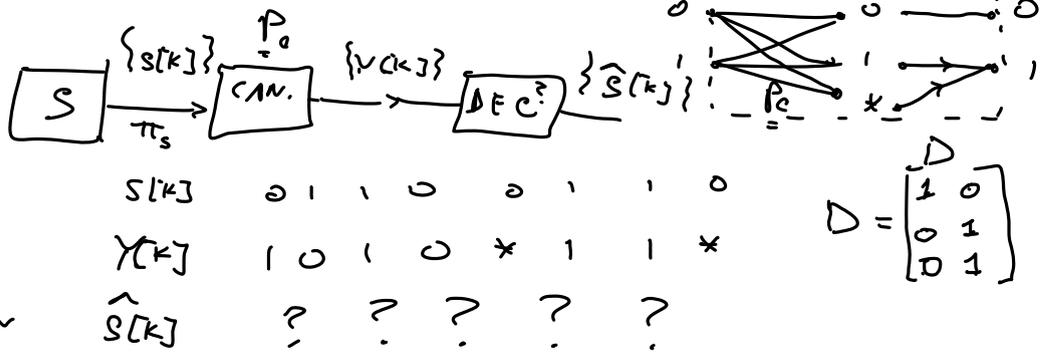
$$X \xrightarrow{P_1} Z \xrightarrow{P_2} Y$$

$$I(X; Y) \leq I(X; Z)$$

Divergenza di Kullback-Leibler $P \quad Q \quad \frac{P}{Q}$

$$D(P||Q) \triangleq \sum p \log \frac{p}{q} \geq 0$$

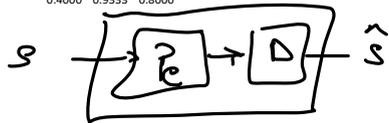
$$D(Q||P) \triangleq \sum q \log \frac{q}{p}$$



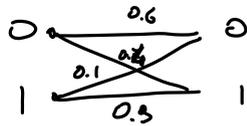
$$P_c = \begin{bmatrix} p(0|0) & p(1|0) & p(x|0) \\ p(0|1) & p(1|1) & p(x|1) \end{bmatrix} \Rightarrow P = \begin{bmatrix} P(X=0|Y=0) & P(X=0|Y=1) & P(X=0|Y=x) \\ P(X=1|Y=0) & P(X=1|Y=1) & P(X=1|Y=x) \end{bmatrix}$$

Esempio $\pi_x = \begin{bmatrix} 0.2 \\ 0.8 \end{bmatrix}$ $P_c = \begin{bmatrix} 0.6 & 0.2 & 0.2 \\ 0.1 & 0.7 & 0.2 \end{bmatrix}$

0.6000 0.0667 0.2000
0.4000 0.9333 0.8000



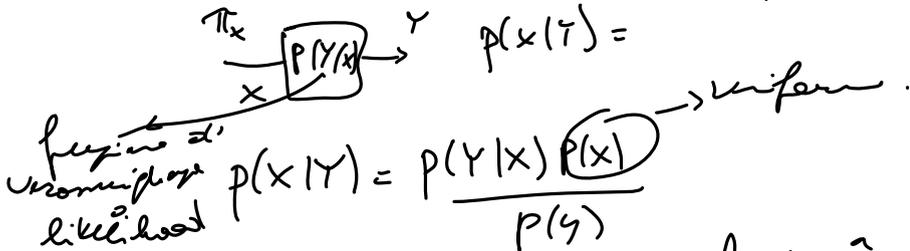
$$P_t = P_c \cdot D = \begin{bmatrix} 0.6 & 0.4 \\ 0.1 & 0.9 \end{bmatrix}$$



$$P_c = 1 - P_t = 1 - 0.2 \cdot 0.6 - 0.8 \cdot 0.9 = 1 - \pi_x^T \text{diag}(P_c) = 0.32$$

Decisione (i. di Bayes) MAP

MAXIMUM A POSTERIORI



$$p(x|Y) = \frac{p(Y|x) p(x)}{p(Y)}$$

MAP

$$p(x|Y=\bar{y}) \rightarrow \text{sceglie } x=\hat{x} : p(\hat{x}|Y=\bar{y}) > p(x|Y=\bar{y}) \quad \forall x$$

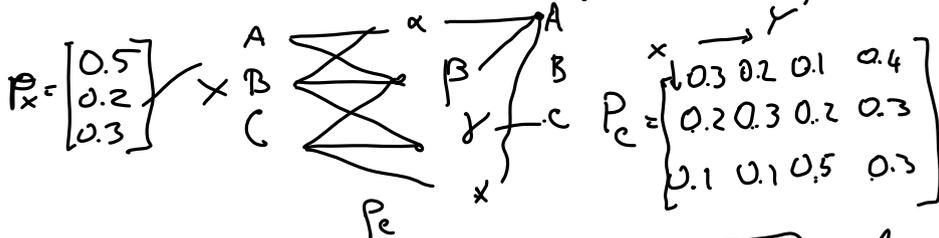
$$p(x|Y=\bar{y}) \propto p(Y=\bar{y}|x)$$

ML (Maximum Likelihood)

Massima verosimiglianza MV

$$\text{sceglie } x=\hat{x} : p(Y=\bar{y}|\hat{x}) > p(Y=\bar{y}|x) \quad \forall x$$

$$X \in \mathcal{X} = \{A, B, C\} \quad Y \in \{\alpha, \beta, \gamma, \delta\}$$



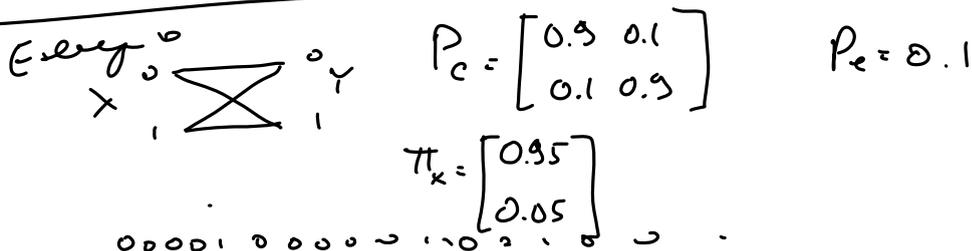
$$D = \begin{pmatrix} 1 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 1 \\ 1 & 0 & 0 \end{pmatrix}$$

$$P_t = P_c D = \text{MAP}$$

Pp =
0.6818 0.5263 0.2083 0.5714
0.1818 0.3158 0.1667 0.1714
0.1364 0.1579 0.6250 0.2571

Pt =
0.9000 0 0.1000
0.8000 0 0.2000
0.5000 0 0.5000

$$P_c = 1 - \pi_x^T \text{diag}(P_c) = 0.4$$



$\pi_x = \begin{bmatrix} 0.95 \\ 0.05 \end{bmatrix}$
 $P_p = \begin{bmatrix} 0.9942 & 0.6786 \\ 0.0058 & 0.3214 \end{bmatrix}$
 $P_c = \begin{bmatrix} 1 & 0 \\ 1 & 0 \end{bmatrix}$
 $D = \begin{bmatrix} 1 & 0 \\ 1 & 0 \end{bmatrix}$

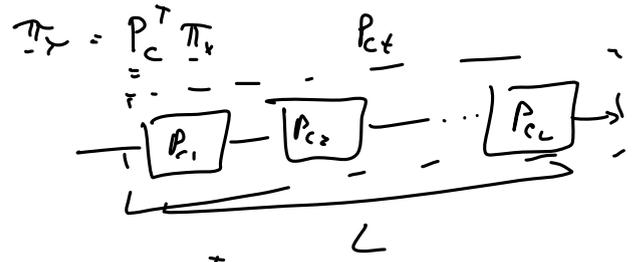
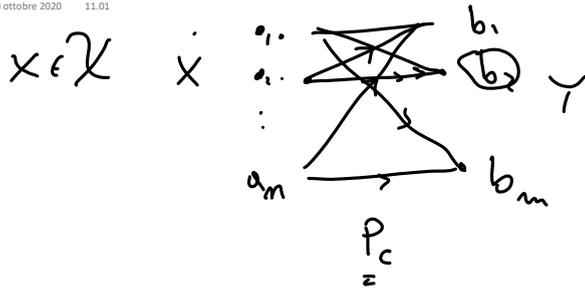
$$P_t = P_c D = \begin{bmatrix} 1 & 0 \\ 1 & 0 \end{bmatrix}$$

$$P_r = \mathbf{1} - \pi_x^T \text{diag}(P_c) = \mathbf{1} - (0.95 \ 0.05) \begin{pmatrix} 1 \\ 0 \end{pmatrix} = 0.05 \quad \text{!!!}$$

π_x
 $H(x)$

$P_c = \text{---}$ *Problema*
 Teoria della T.I

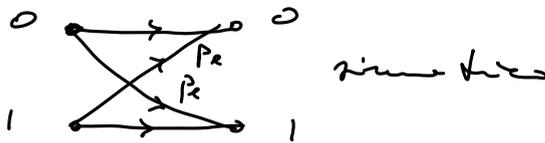
$I(x, t)$
 $C ?$



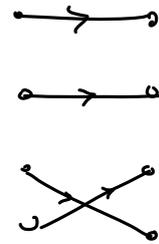
$P_c \{x | y = \bar{y}\}$ ↓ di Bayes.

$P_{c_t} = P_{c_1} P_{c_2} \dots P_{c_L}$

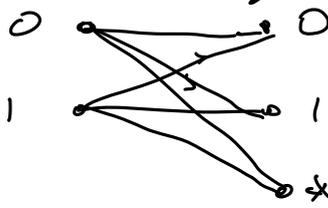
Esempi di canale discreto



$P_c = \begin{bmatrix} p(0|0) & p(1|0) \\ p(0|1) & p(1|1) \end{bmatrix} = \begin{pmatrix} 1-P_e & P_e \\ P_e & 1-P_e \end{pmatrix}$



Canale con cancellazione

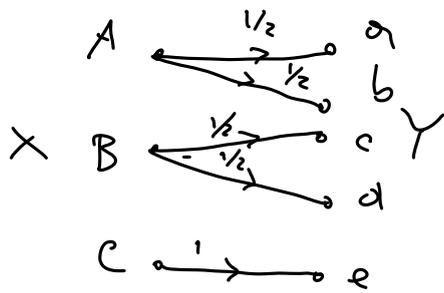


0110010001
 $100 * * 11001$

$P_c = \begin{pmatrix} p(0|0) & p(1|0) & p(*|0) \\ p(0|1) & p(1|1) & p(*|1) \end{pmatrix}$

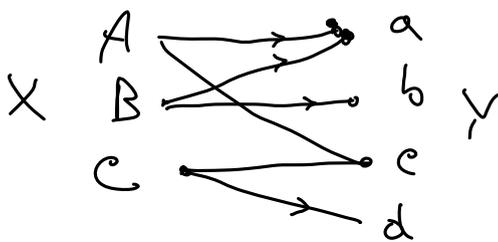
canale simmetrico

$= \begin{pmatrix} 1-P_e-P_c & P_e & P_c \\ P_e & 1-P_e-P_c & P_c \end{pmatrix}$



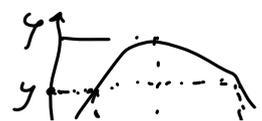
$P_c = \begin{pmatrix} 1/2 & 1/2 & 0 & 0 & 0 \\ 0 & 0 & 1/2 & 1/2 & 0 \\ 0 & 0 & 0 & 0 & 1 \end{pmatrix}$

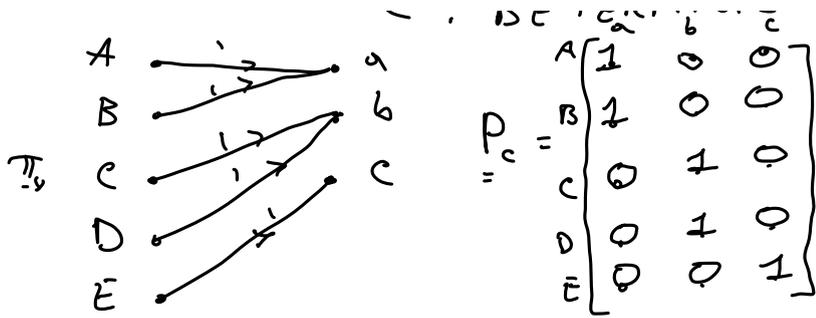
CANALE SENZA RUMORE



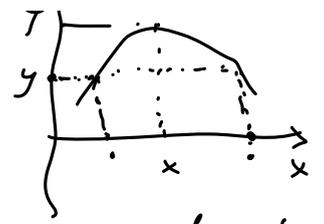
C. DETERMINISTICO

$A \begin{bmatrix} 1 & 0 & 0 \\ -1 & 0 & 0 \end{bmatrix}$

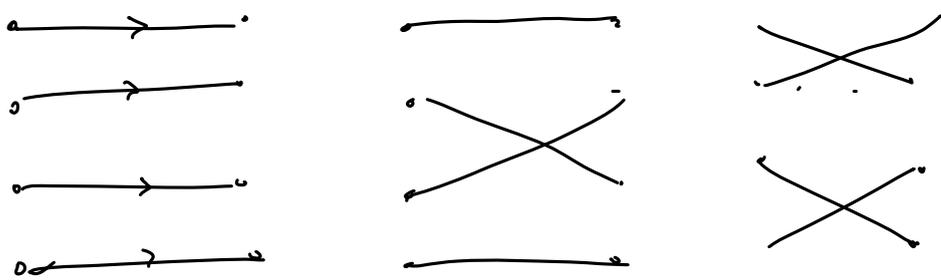




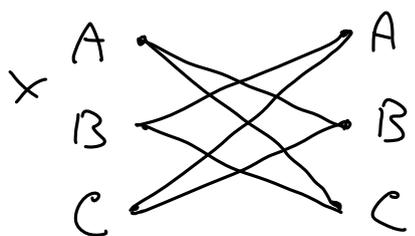
$$P_c = \begin{matrix} & a & b & c \\ \begin{matrix} A \\ B \\ C \\ D \\ E \end{matrix} & \begin{pmatrix} 1 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{pmatrix} \end{matrix}$$



inversione storica
di una funzione
non invertibile
(a più valori)

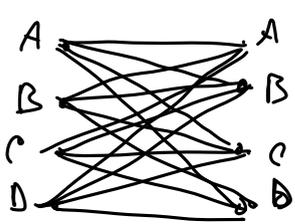
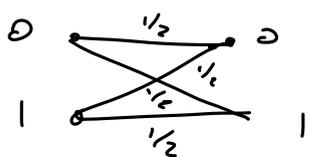


CANALE SEMPRE IN ERRORE

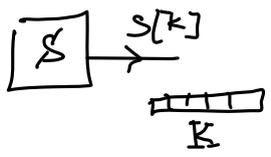


$$P_c = \begin{matrix} & A & B & C \\ \begin{matrix} A \\ B \\ C \end{matrix} & \begin{pmatrix} 0 & x & x \\ x & 0 & x \\ x & x & 0 \end{pmatrix} \end{matrix}$$

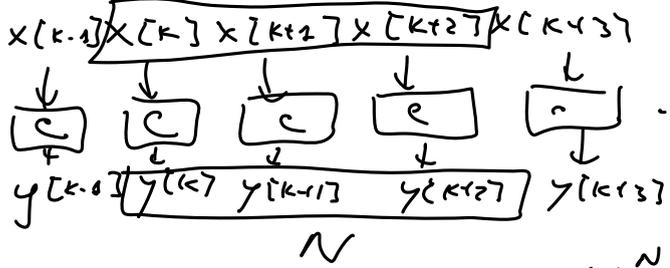
CANALE MASSIMAMENTE CONFUSO



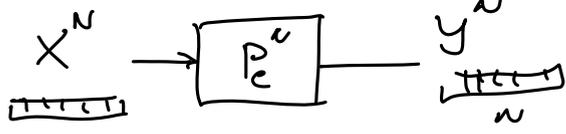
$$P_c = \begin{pmatrix} \frac{1}{4} & \frac{1}{4} & \frac{1}{4} & \frac{1}{4} \\ \frac{1}{4} & \frac{1}{4} & \frac{1}{4} & \frac{1}{4} \\ \frac{1}{4} & \frac{1}{4} & \frac{1}{4} & \frac{1}{4} \\ \frac{1}{4} & \frac{1}{4} & \frac{1}{4} & \frac{1}{4} \end{pmatrix}$$



$s \in \mathcal{R}$
 $s^k \in \mathcal{R}^k = \underbrace{\mathcal{R} \times \mathcal{R} \times \dots \times \mathcal{R}}_k$



$x^N \in \mathcal{X}^N$
 $\mathcal{X} = \{x_1, \dots, x_n\}$



$y^N \in \mathcal{Y}^N$
 $\mathcal{Y} = \{y_1, \dots, y_m\}$
 $y^N = \{ \underbrace{\dots}_{m^N} \}$

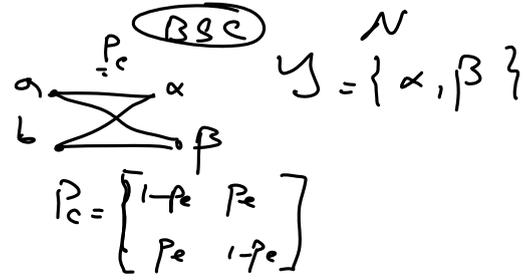
Ex.
 $\mathcal{X} = \{a, b, c\}$
 $\mathcal{X}^2 = \{aa, ab, ac, ba, bb, bc, ca, cb, cc\}$
 $\mathcal{X}^3 = \{aaa, \dots\}$
 $\mathcal{X}^N = \{ \dots \}$

$x_{1^N}^N, x_{2^N}^N, \dots, x_{n^N}^N$

$y_{1^N}^N, y_{2^N}^N, \dots, y_{m^N}^N$

$$P_c^N = \underbrace{P_c \otimes P_c \otimes \dots \otimes P_c}_N$$

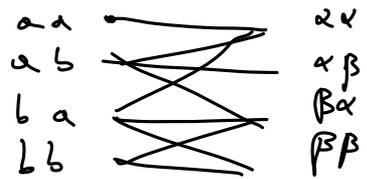
Example
 $\mathcal{X} = \{a, b\}$
 $N = 2$



$$P_c = \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix}$$

$\mathcal{X}^2 = \{aa, ab, ba, bb\}$

$\mathcal{Y}^2 = \{\alpha\alpha, \alpha\beta, \beta\alpha, \beta\beta\}$

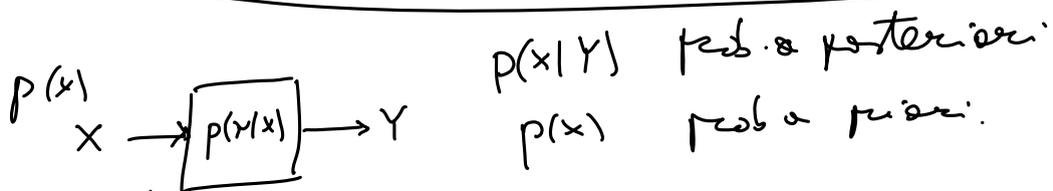


$$P_c^2 = P_c \otimes P_c = \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix} \otimes \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix}$$

$$= \begin{bmatrix} (1-p) \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix} & p \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix} \\ p \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix} & (1-p) \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix} \end{bmatrix}$$

$$\left[p \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix} (1-p) \begin{bmatrix} 1-p & p \\ p & 1-p \end{bmatrix} \right]$$

$$= \begin{bmatrix} (1-p)(1-p) & (1-p)p & p(1-p) & pp \\ (1-p)p & (1-p)(1-p) & pp & p(1-p) \\ p(1-p) & pp & (1-p)(1-p) & (1-p)p \\ pp & p(1-p) & (1-p)p & (1-p)(1-p) \end{bmatrix}$$



incertezza in x dato y

$$H(X|Y) \triangleq \sum_{(x,y) \in \mathcal{X} \times \mathcal{Y}} p(x,y) \log_2 \frac{1}{p(x|y)} =$$

$p(x|y) = \frac{p(x,y)}{p(y)}$

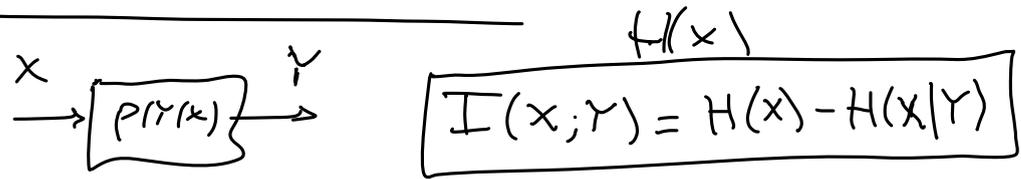
$= \frac{p(y|x)p(x)}{\sum_x p(y|x)p(x)}$

AMBITO UTM
 DEL CANALE
 bit.

$$= \sum_{x,y \in \mathcal{X} \times \mathcal{Y}} p(x,y) p(y) \log_2 \frac{1}{p(x|y)}$$

$$= \sum_{(x,y)} p(y|x)p(x) \log_2 \frac{\sum_x p(y|x)p(x)}{p(y|x)p(x)}$$

MUTUA INFORMAZIONE



$$I(x;Y) = I(Y;x) = H(Y) - H(Y|x)$$

$$I(x;Y) = H(x) - H(x|Y) \geq 0$$

$$= \sum_x p(x) \log_2 \frac{1}{p(x)} - \sum_{x,y} p(x,y) \log_2 \frac{1}{p(x|y)}$$

$$p(x) = \sum_y p(x,y)$$

$$= \sum_{x,y} p(x,y) \log_2 \frac{1}{p(x)} - \sum_{x,y} p(x,y) \log_2 \frac{1}{p(x,y)}$$

$$= \sum_{x,y} p(x,y) \log_2 \frac{p(x,y)}{p(x)}$$

$$= \sum_{x,y} p(x,y) \log_2 \frac{p(x|y)}{p(x)}$$

$$= \sum_{x,y} p(x,y) \log_2 \frac{p(x,y)}{p(y)p(x)}$$



$$I(x;Y) \geq 0 \quad I(x;Y) = I(Y;X)$$

$I(x;Y)$ concave rispetto a $p(x)$

$I(x;Y)$... U rispetto a $p(y|x)$

X, Y independenti:

$$I(x;Y) = H(x) - H(x|Y) = 0$$

||
 $H(x)$

$$p(x,y) = p(x)p(y)$$

$$p(y|x) = p(y)$$

$$p(x|y) = p(x)$$

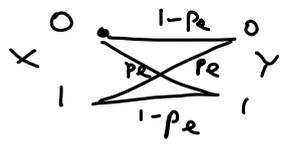
caso di uguaglianza



$$H(x|Y) = 0$$

$$I(x;Y) = H(x)$$

$$\pi_x = \begin{bmatrix} p \\ 1-p \end{bmatrix}$$



$$I(x;Y) = H(x) - H(x|Y)$$

1 1 2

$$H(x|Y) = \sum_x \sum_y p(x,y) \log_2 \frac{1}{p(x|y)}$$

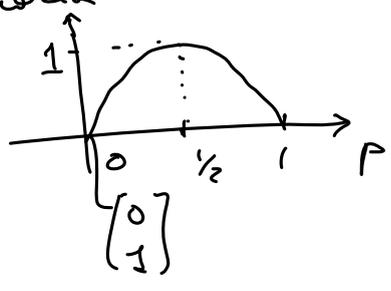
$$H(x) = \sum_x p(x) \log_2 \frac{1}{p(x)} = p \log_2 \frac{1}{p} + (1-p) \log_2 \frac{1}{1-p} \leq \log_2 2 = 1$$

$$\begin{bmatrix} p \\ 1-p \end{bmatrix} \log_2 \frac{1}{p_x} \quad \text{...} \quad \begin{bmatrix} \cdot \\ \cdot \end{bmatrix} \begin{bmatrix} \cdot \\ \cdot \end{bmatrix} \quad \text{...} \quad - p_x^T \log_2 p_x$$

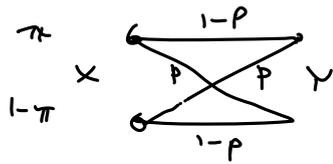
$H(x)$ sorgente binaria

$$p(x) = \begin{bmatrix} p \\ 1-p \end{bmatrix}$$

$$H(x) = p \log_2 \frac{1}{p} + (1-p) \log_2 \frac{1}{1-p}$$

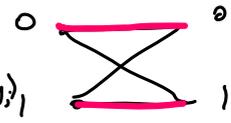
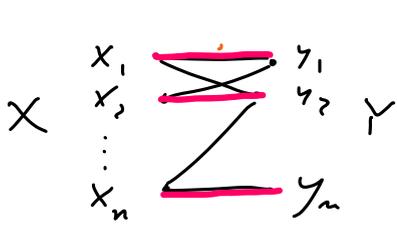


... 1-p



$$I(X; Y) = ((1-p)\pi + p(1-\pi)) \log_2 \frac{1}{(1-p)\pi + p(1-\pi)} + (p\pi + (1-p)(1-\pi)) \log_2 \frac{1}{p\pi + (1-p)(1-\pi)}$$

$$P(e) \leftrightarrow H(X), H(X|Y), I(X;Y), C$$

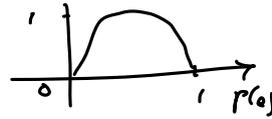


$$P(e) = P\{Y \neq X\} = \sum_{i=1}^n \sum_{j=1, j \neq i}^n p(x_i, y_j)$$

$$= 1 - P\{Y = X\} = 1 - \sum_{i=1}^n p(x_i, y_i)$$

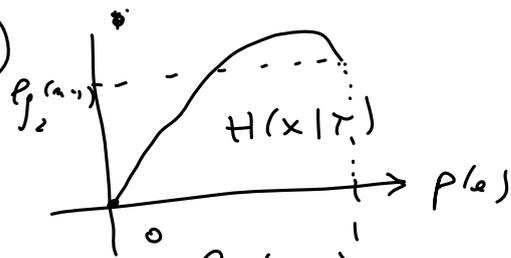
DIS. DI FANO:

$$H(X|Y) \leq H(e) + P(e) \log_2(n-1)$$



$$H(e) = -\left(p(e) \log_2 p(e) - (1-p(e)) \log_2 (1-p(e)) \right)$$

$F(p(e))$

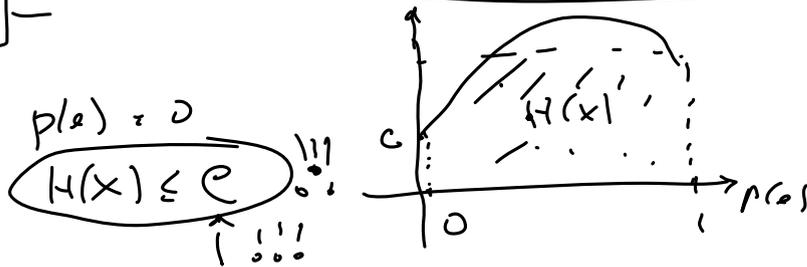


$$I(X;Y) = H(X) - H(X|Y)$$

$$H(X) - I(X;Y) \leq H(p(e)) + p(e) \log_2(n-1)$$

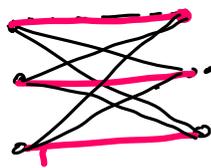
$$H(X) \leq I(X;Y) + H(p(e)) + p(e) \log_2(n-1)$$

$$H(X) \leq C + H(p(e)) + p(e) \log_2(n-1)$$



$p(e) = 0$
 $H(X) \leq C$

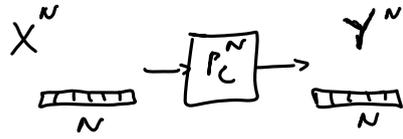
CANALI UNIFORMI



$$P_e = \begin{pmatrix} 1-p_e & \frac{p_e}{2} & \frac{p_e}{2} \\ \frac{p_e}{2} & 1-p_e & \frac{p_e}{2} \\ \frac{p_e}{2} & \frac{p_e}{2} & 1-p_e \end{pmatrix}$$

$$\begin{pmatrix} 1-p_e & \frac{p_e}{n-1} & \dots & \frac{p_e}{n-1} \\ \frac{p_e}{n-1} & 1-p_e & & \dots \\ \dots & & \dots & \dots \\ \dots & & & 1-p_e \end{pmatrix}$$

$$H(X|Y) \leq \underbrace{H(e_w)}_1 + \underbrace{P(e_w) \log_2(n-1)}_2$$



$$P(e_w) = P\{X^N \neq Y^N\}$$

D/S. DIFANO

$$H(X^N|Y^N) \leq H(e_w) + P(e_w) \log_2(n-1)$$

$$C^N = \max_{\pi(X^N)} I(X^N; Y^N) = NC$$

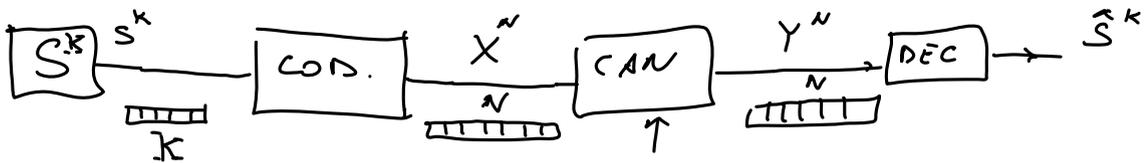
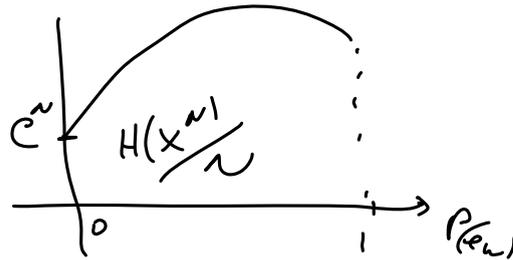
$$H(X^N) \leq C^N + H(e_w) + P(e_w) \log_2(n-1)$$

$$H(X^N) \leq NC + H(e_w) + P(e_w) \log_2(n-1)$$

$$\frac{H(X^N)}{N} \leq C + \frac{H(e_w)}{N} + \frac{P(e_w)}{N} \log_2(n-1)$$

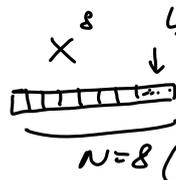
$P(e_w) = 0$

$$\frac{H(X^N)}{N} < C$$

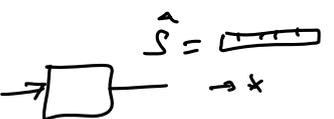
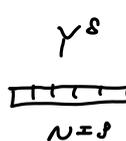


Esempio:
Controllo di parità (1 bit)

$K=7$ $n=2$

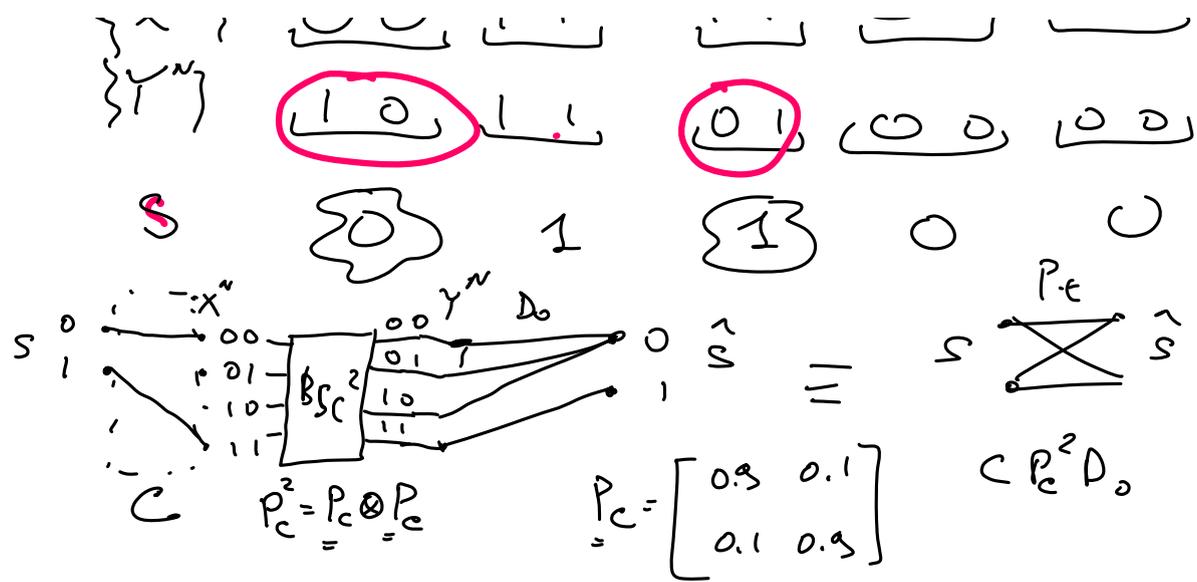


$N=8$ (256)



Codice a ripetizione (lineare)

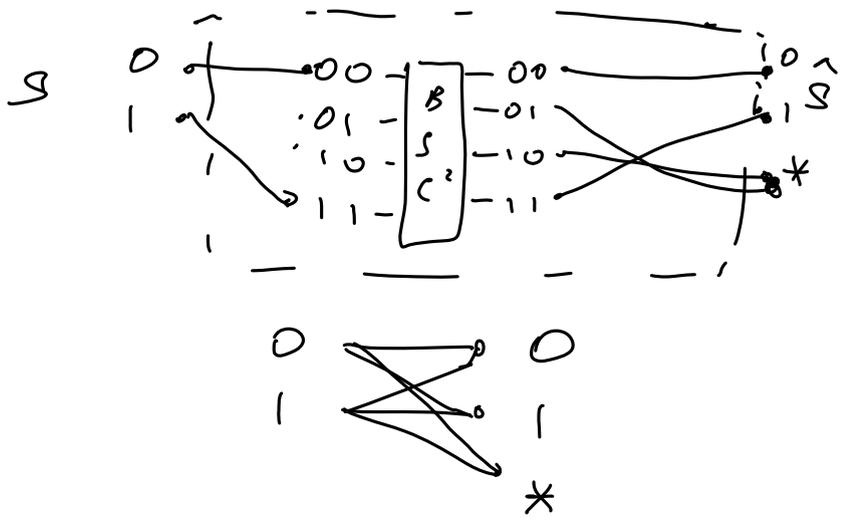




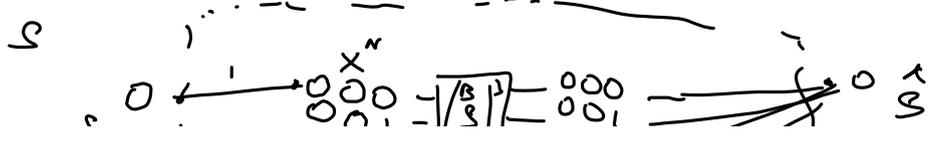
$$P_c^2 = P_c \otimes P_c = \begin{bmatrix} 0.9 \begin{bmatrix} 0.9 & 0.1 \\ 0.1 & 0.9 \end{bmatrix} & 0.1 \begin{bmatrix} 0.9 & 0.1 \\ 0.1 & 0.9 \end{bmatrix} \\ 0.1 \begin{bmatrix} 0.9 & 0.1 \\ 0.1 & 0.9 \end{bmatrix} & 0.9 \begin{bmatrix} 0.9 & 0.1 \\ 0.1 & 0.9 \end{bmatrix} \end{bmatrix}$$

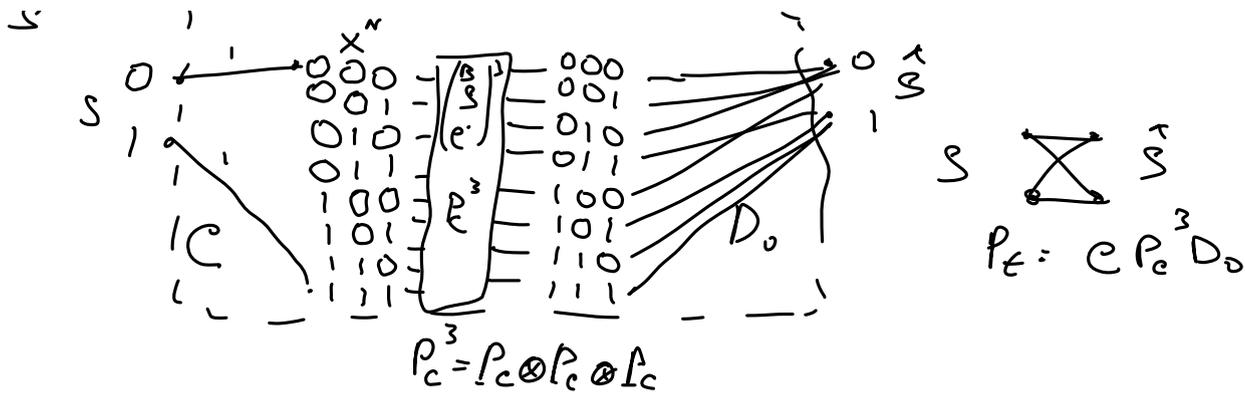
$$C = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

$$C P_c^2 = \begin{bmatrix} \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots \end{bmatrix}$$

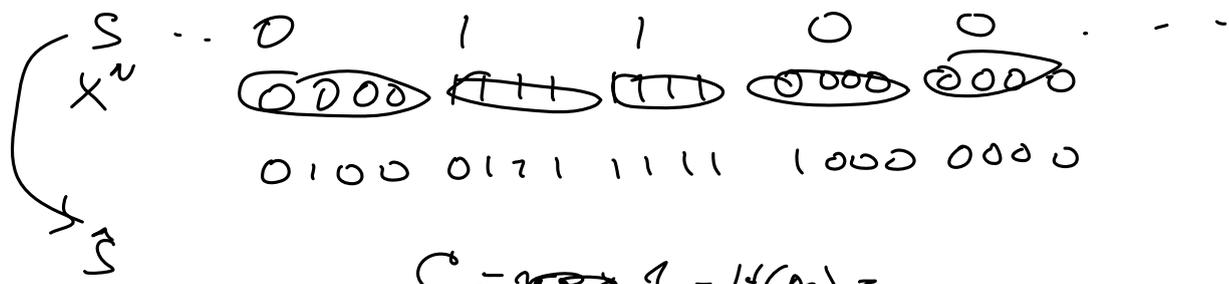


S	0	1	0	0	1
x^N	000	111	000	000	111
y^N	010	110	100	000	101





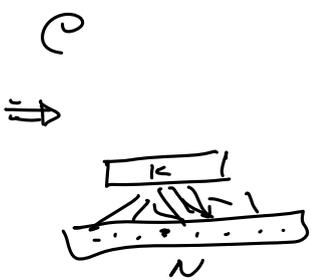
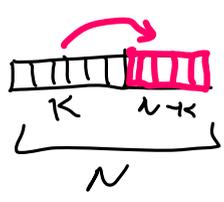
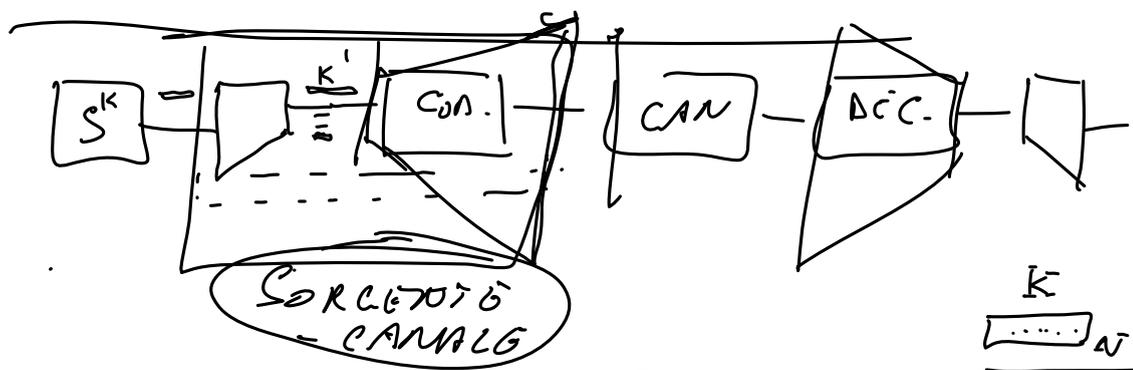
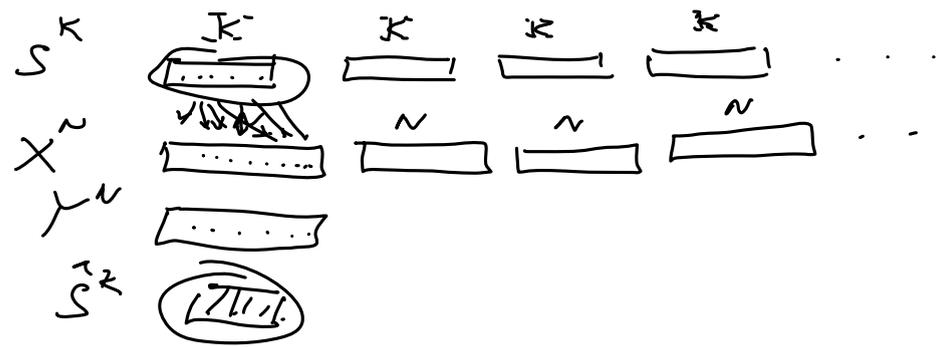
$$C P_e^3 = (\dots \dots \dots)$$



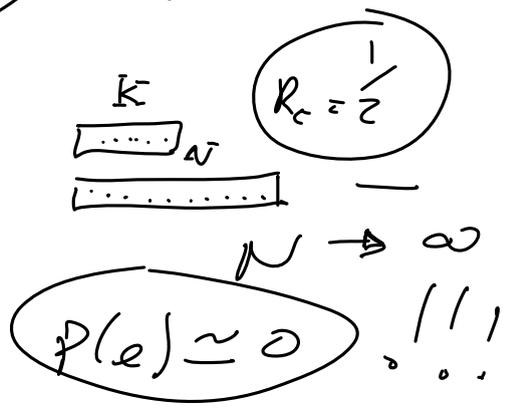
$$R_c = \frac{k}{N} < 1$$

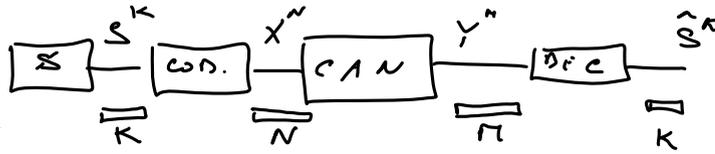
$$C \approx \frac{1}{2} \ln$$

$$R_c = \frac{1}{2}$$



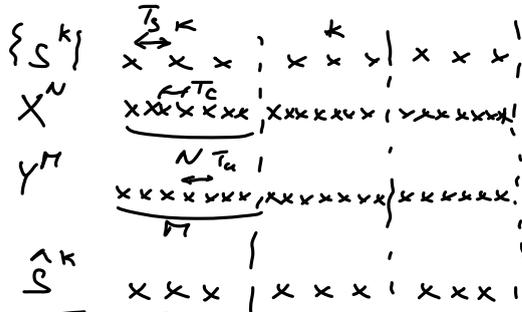
$$N = 100$$





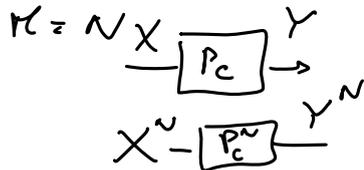
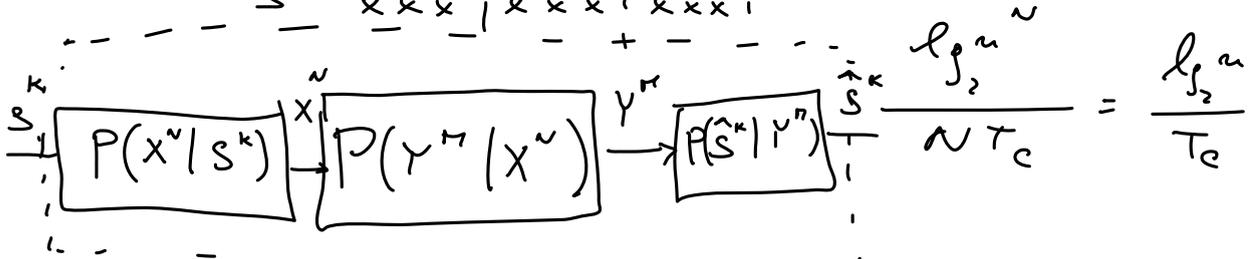
$S \in \mathcal{A} = \{a, \dots, a_d\}$

$X \in \{x_1, \dots, x_m\}$
 $Y \in \{y_1, \dots, y_n\}$

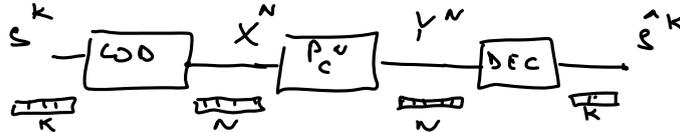


$kT_s = NT_c = \pi T_u$

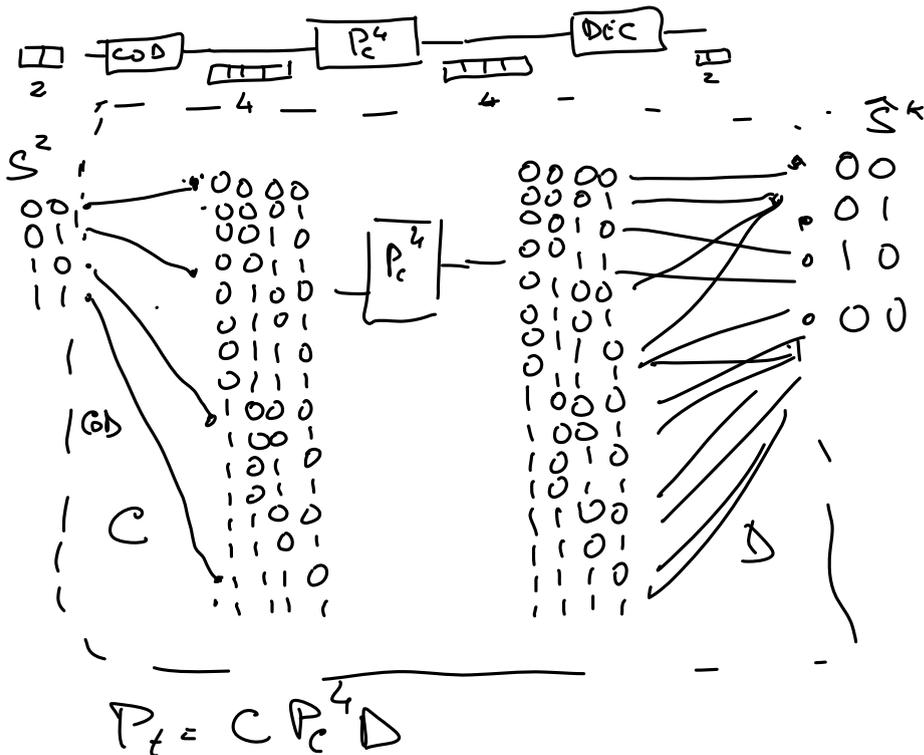
$R = \frac{\beta_2 d^k}{kT_s} = \frac{k \log_2 d}{kT_s}$



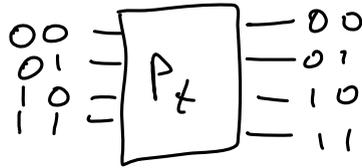
$\pi = N$ binario



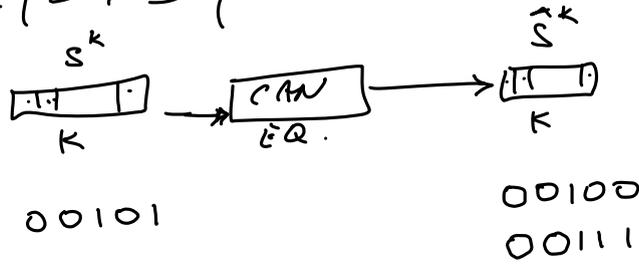
Esempio $k=2, d=2, N=4$



$P_t = C P_c^4 D$



$$P_{ew} = P_2 \{ \hat{S}^k \neq S^k \} = 1 - P_2 \{ \hat{S}^k = S^k \}$$



$P_2 \{ \text{errore di parola} \} \leftrightarrow P_2 \{ \text{errore per bit} \}$



$$P_2 \{ \hat{S}^k = s^i \mid S^k = s^i \}$$

$$P_{ew} = 1 - P_{ew} = 1 - P_2 \{ \hat{S}^k = S^k \}$$

$$= 1 - \sum_{i=1}^{2^k} P_2 \{ \hat{S}^k = s^i \mid S^k = s^i \} P_2 \{ S^k = s^i \}$$

o le prob. a priori sono uguali.
 $P(S^k = s^i) = 2^{-k}$

$$= 1 - \sum_{i=1}^{2^k} P_2 \{ \hat{S}^k = s^i \mid S^k = s^i \} 2^{-k}$$

$P_b =$ Prob. di errore a livello di bit.

\Rightarrow Quante bit sono mediate in errore \Rightarrow # bit diversi $\left(\begin{matrix} \hat{S}^k \\ \text{[1][1][1]} \end{matrix} - \begin{matrix} S^k \\ \text{[1][1][1]} \end{matrix} \right)$

$$P_b = \frac{E[\# \text{ bit in errore}]}{K} = \frac{E[d_H(\hat{S}^k, S^k)]}{K}$$

$$= \frac{1}{K} \sum_{i=1}^{2^k} \sum_{j=1}^{2^k} d_H(\hat{S}^k = s^i, S^k = s^j) P_2 \{ \hat{S}^k = s^i, S^k = s^j \}$$

$$= \frac{1}{K} \sum_{i=1}^{2^k} \sum_{j=1}^{2^k} d_H(\hat{S}^k = s^i, S^k = s^j) P_2 \{ \hat{S}^k = s^i \mid S^k = s^j \} P_2 \{ S^k = s^j \}$$

$$= \frac{1}{K} \sum_{i=1}^{2^k} \sum_{j=1}^{2^k} \alpha_H(\hat{s}_{=i}^k, s_{=j}^k) \{P\} \hat{s}_{=i}^k | s_{=j}^k \} \underbrace{\{H\} s_{=j}^k \}_{2^{-k}}$$

Esempio $k=3$

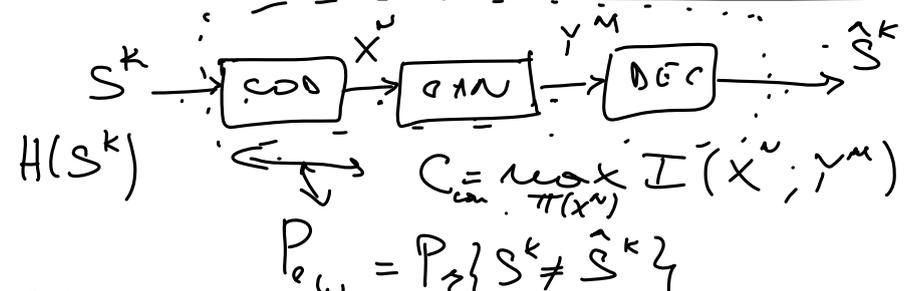
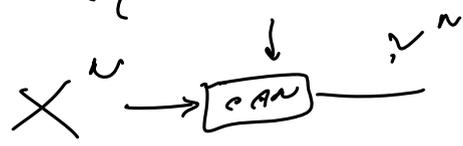
000	000
001	001
010	010
011	011
100	100
101	101
110	110
111	111

α_H	000	001	010	011	100	101	110	111
000	0	1	2	2	1	2	2	3
001	1	0	2	2				
010			0					
011				0				
100					0			
101						0		
110							0	
111								0

D_H

$$P_6 = \frac{1}{K} e^{-k} e^T (D_H \circ P_t) e$$

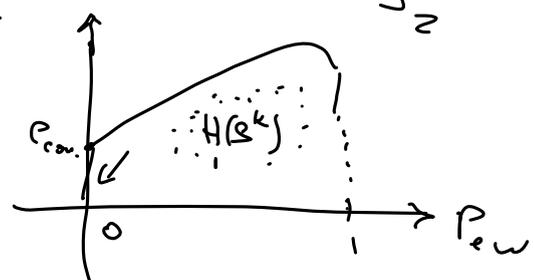
$r = \begin{pmatrix} 1 \\ \vdots \\ 1 \end{pmatrix}$



TEOREMA INVERSO DELLA CODIFICA

$$H(S^k) \leq C_{can} + H(P_{ew}) + P_{ew} \log_2(2^k - 1)$$

$P_{ew} = 0$
 $H(S^k) \leq C_{can}$



$$H(P_{ew}) = P_{ew} \log_2 \frac{1}{P_{ew}} + (1 - P_{ew}) \log_2 \frac{1}{1 - P_{ew}}$$

$$111 e^k | 2^k | - 11(P \gamma, P \circ (2^k - 1))$$

$$H(S^k | \hat{S}^k) \leq H(P_{ew}) + P_{ew} \log_2(d^{k-1})$$

$$I(S^k; \hat{S}^k) = H(S^k) - H(S^k | \hat{S}^k)$$

$$I(S^k; \hat{S}^k) \leq I(X^N; Y^M) \leq C_{can}$$

$$H(S^k) - H(S^k | \hat{S}^k) \leq C_{can}$$

$$H(S^k) \leq C_{can} + H(S^k | \hat{S}^k)$$

$$H(S^k) \leq C_{can} + H(P_{ew}) + P_{ew} \log_2(d^{k-1})$$

Con particolare $P_{ew} = 0$

$$H(S^k) \leq C_{can}$$

simboli di sorgente indipendenti.

$$KH(S) \leq C_{can}$$

$N=M$

$$N = \begin{matrix} \text{---} \\ \text{---} \end{matrix} N$$

$$P_c^N = P_c \otimes P_c \otimes \dots \otimes P_c$$

C capacità del canale
uso del canale.

$$KH(S) \leq N C$$

$$H(S) \leq \frac{N}{K} C$$

sorgente binaria
o simboli iid,
e uniformi: $H(S) = 1$

canale binario
o usi indipendenti:

$$1 \leq \frac{N}{K} C$$

$$\left(\frac{K}{N} \right) \leq C$$

Rapporto di codifica



canale: BSC P_e

k $\underbrace{\quad\quad}$ N $\underbrace{\quad\quad}$

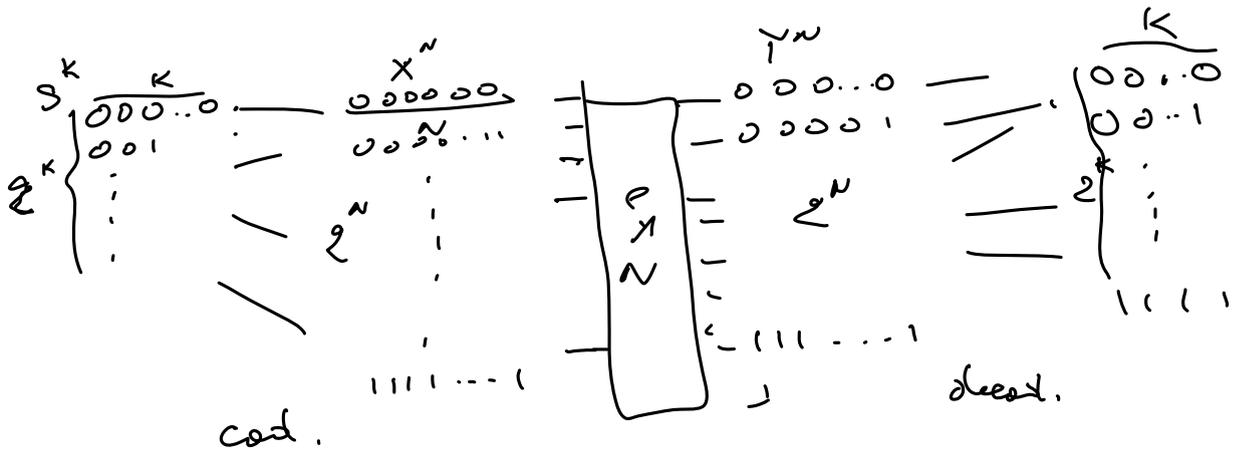
esempio: BSC p_e
 $C = 1 - H(p_e)$

$$\frac{k}{N} \leq 1 - H(p_e)$$

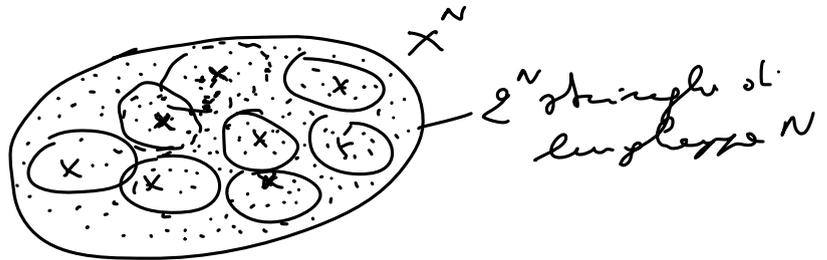
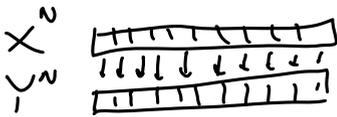
Es. $p_e = 0.1$ \downarrow 0.53

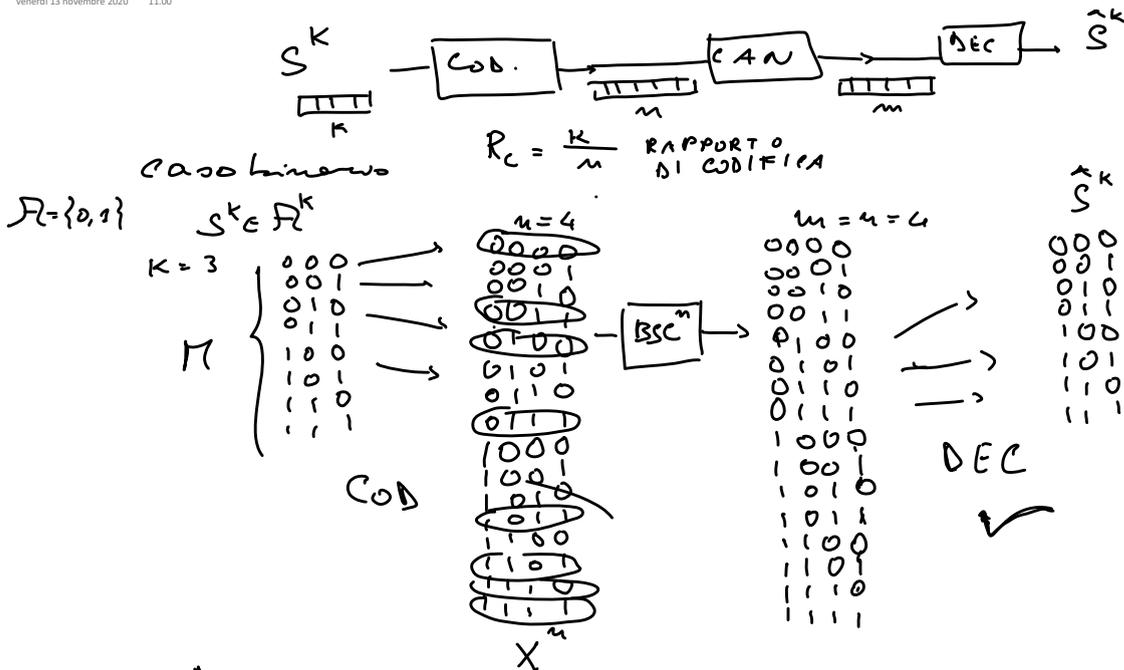
$$\frac{k}{N} \leq 0.53$$

00 0000
 000 000000



$CAN : (BSC)^N$



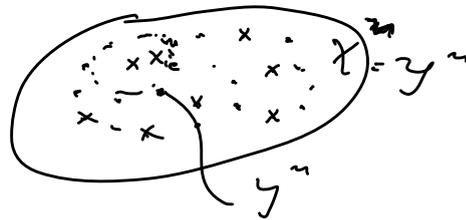


DEC : MAP = min. P(e)

$\pi(S^K)$ Uniforme.
CANALE BSC^m

$$S^K = \{s_1^K, s_2^K, \dots, s_{2^k}^K\} \quad X^m = \{x_1^m, x_2^m, \dots, x_{2^k}^m\}$$

$$X_c^m = \{x_{1c}^m, x_{2c}^m, \dots, x_{2^k c}^m\}$$



$$Y^m = X^m$$

MAP → MV

Teor.: Regola ottima è quella a minima distanza di Hamming

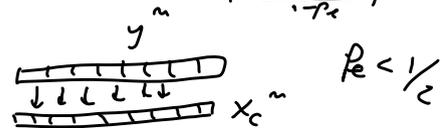
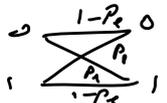
$$P_2(Y^m = y^m | X^m = x_c^m) = p^{d_H(y^m, x_c^m)} (1-p)^{n-d_H(y^m, x_c^m)}$$

$$Q = p^d (1-p)^{n-d} \quad p < \frac{1}{2}$$

$$d_1 < d_2$$

$$p^{d_1} (1-p)^{n-d_1} > p^{d_2} (1-p)^{n-d_2}$$

$$\rightarrow d_1 - d_2 \quad \dots \quad n - d_2 - n + d_1$$



$$p_e < \frac{1}{2}$$

1 2 1 1 1 1 1 1

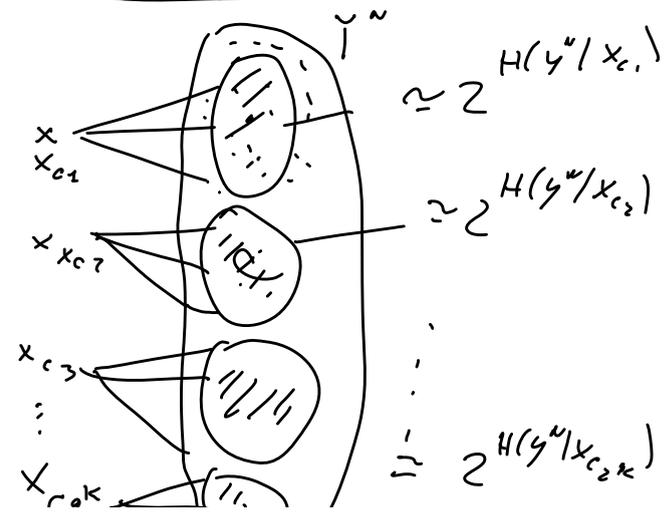
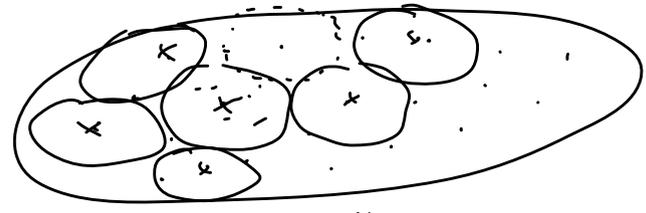
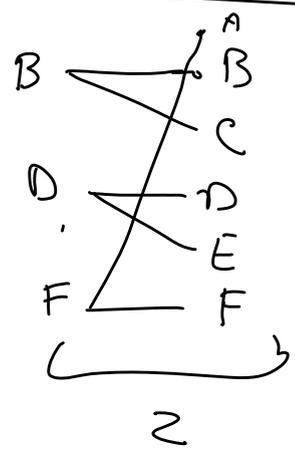
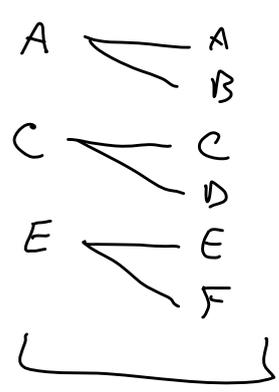
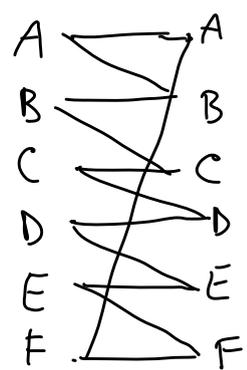
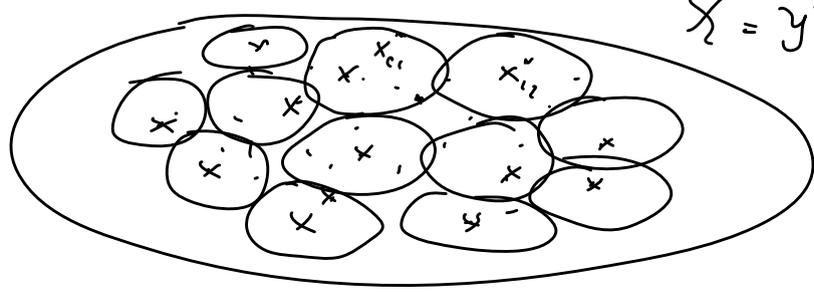
$$p^{d_1 - d_2} > (1-p)^{d_1 - d_2 - d_1}$$

$$p^{d_1 - d_2} > (1-p)^{d_1 - d_2}$$

$$\left(\frac{p}{1-p}\right)^{d_1 - d_2} > 1$$

.....

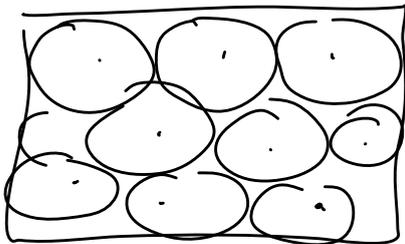
$$x^{\sim} = y^{\sim}$$



→ Dimensione media di ogni sfera $\propto H(y^u/x^u)$



Dimensione di y^N
 $\approx 2^{H(y^N)}$



$$\frac{2^{H(y^N)}}{2^{H(x^N/x^N)}} \approx \# \text{ di sfere non sovrapposte}$$

$$= \# \text{ di parole codice}$$

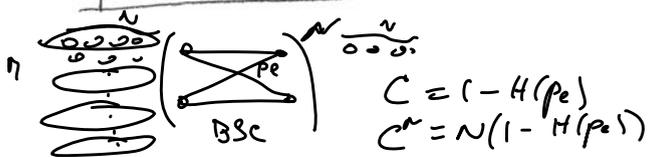
$$2^{H(y^N) - H(x^N/x^N)} = 2^{I(x^N; y^N)} < 2^{C_N}$$

$$C \approx \mathbb{R}^K \approx 2^C$$

TEOREMA DELLA CODIFICA (o II T. DI SHANNON)

TEOREMA:

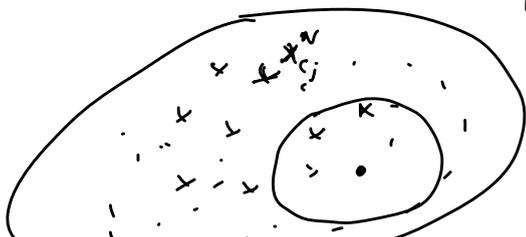
Per un BSC con prob. di errore p_e , ovvero capacità $C = 1 - H(p_e)$ bit,
 $\forall \epsilon > 0$ piccolo a piacere, $\exists N$ tale che è possibile
 trasmettere $M = 2^{N(C-\epsilon)}$ parole codice dalle 2^N all'ingresso
 del canale esteso tali che la probabilità di errore di
 decodifica ($P(e_w)$) sia arbitrariamente piccola.



DIMOSTRAZIONE

Dec ottima \Leftrightarrow Min distanza di Hamming
 \Downarrow
 Dec. subottima \Rightarrow funzione lontana

Prova dipendente dal codice specifico \Rightarrow Prova su un codice esteso
 \Downarrow
 Testando tutti i codici funzionali (T. della cod. canale)





Dec. SUB-OTTIMA (Decodifica a sfera, SPHERE-DECODING)



$$P(UA_i) \leq \sum_{i=1}^N P(A_i)$$

$$P(\text{ew} | X^N = x_{c_j}^N) = P_j(\text{ew})$$

$$P_j(\text{ew}) = \underbrace{P\{x_{c_j}^N \notin Y^E\}}_I + \underbrace{P\{x_{c_j}^N \in Y^E\}}_{\leq 1} \cdot \underbrace{P\{\text{almeno un'altra parola codice} \in Y^E\}}$$

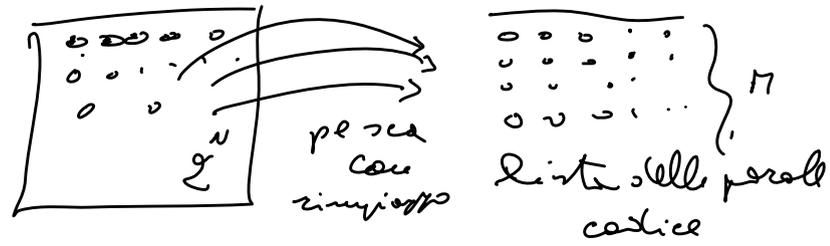
$$P_j(\text{ew}) \leq P\{x_{c_j}^N \notin Y^E\} + P\{\text{almeno un'altra} \in Y^E\}$$

$$\leq \underbrace{P\{x_{c_j}^N \notin Y^E\}}_I + \sum_{\substack{i=1 \\ i \neq j}}^M P\{x_{c_i}^N \in Y^E\}$$

$N \rightarrow \infty$ Prob. $x_{c_j}^N$ cada all'esterno della sfera. \rightarrow asintoticamente piccola.

$$P_j(\text{ew}) \leq \delta + \sum_{\substack{i=1 \\ i \neq j}}^M P\{x_{c_i}^N \in Y^E\}$$

Svincolarsi dal codice specifico



CODICE ALEATORIO
 $(2^N)^M$ # di possibili codici.

2^N non sufficientemente grande tale che la prob. di scegliere lo stesso stringa più volte non fosse trascurabile.

$$\rightarrow \dots \rightarrow \underbrace{P\{x^N \in Y^E\}}$$

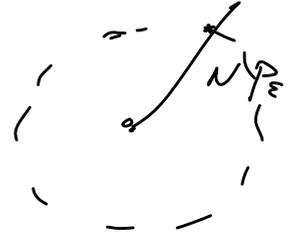
no senso stringa più volte no senso.

mediante
ai possibili
casi

$$\overline{P_j(\omega)} \leq \delta + \sum_{i=1, i \neq j}^M \overline{P_i \{x_c^N \in \mathcal{Y}^\varepsilon\}}$$

$$\overline{P_j(\omega)} \leq \delta + (M-1) \overline{P_2 \{x_c^N \in \mathcal{Y}^\varepsilon\}}$$

$$\overline{P_j(\omega)} \leq \delta + M \overline{P_2 \{x_c^N \in \mathcal{Y}^\varepsilon\}}$$



$$P_2 \{x_c^N \in \mathcal{Y}^\varepsilon\} = \frac{N(N\rho\varepsilon)}{2^N}$$

$$N \rightarrow \infty$$

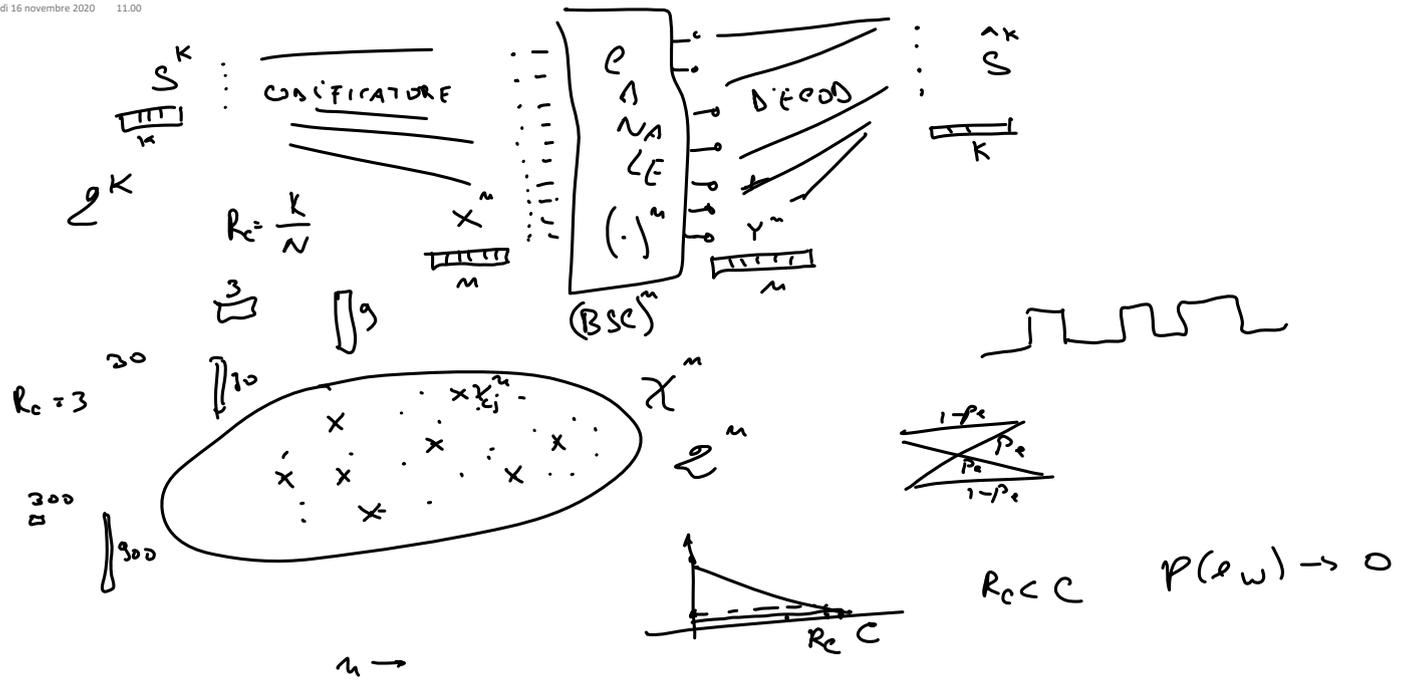
$$M < 2^{\frac{N(1-N(\rho\varepsilon))}{c}}$$

$$\overline{P(\omega)} \rightarrow 0$$

$$2^k < 2^{Nc}$$

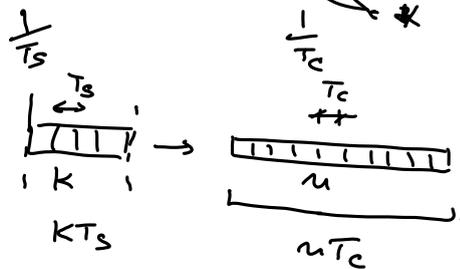
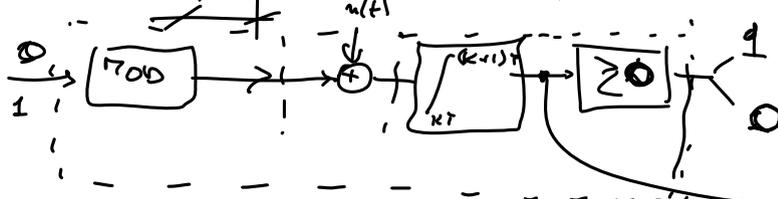
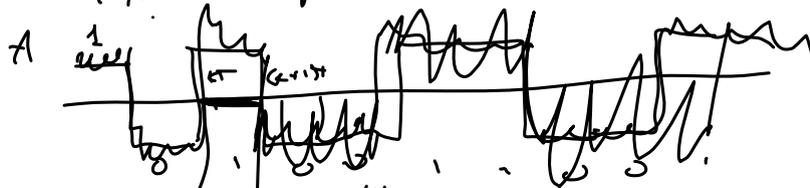
$$\frac{k}{N} < c$$

$$\xrightarrow{N \rightarrow \infty} \overline{P(\omega)} \rightarrow 0$$



Esempio di modulazione numerica.

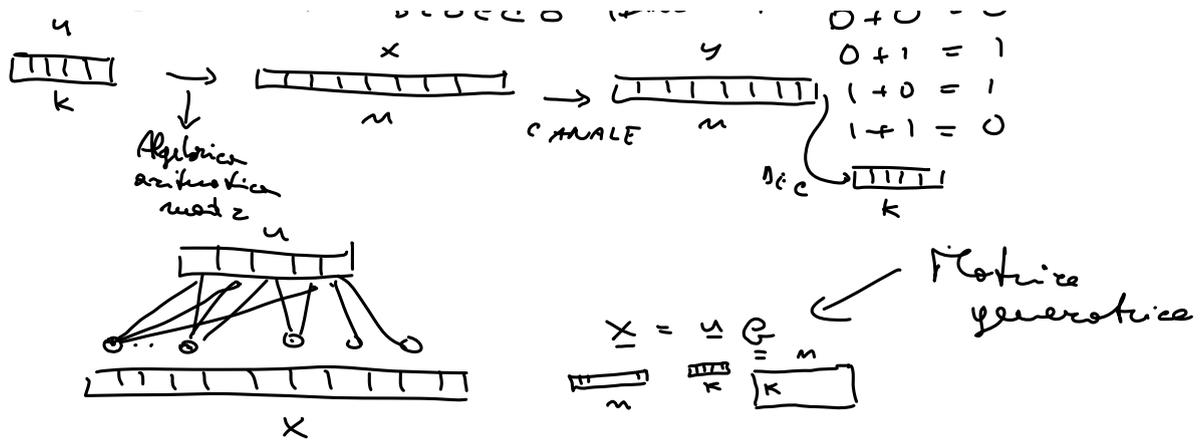
PAM Antipodale



KT_s	nT_c	$R_c = \frac{1}{2}$
4	8	
5	10	
8	16	

CODICI A BLOCCO (LINEARI)





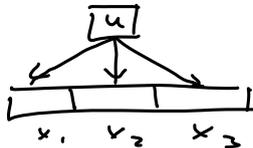
Codice sistematico



(n, k)

Esempio

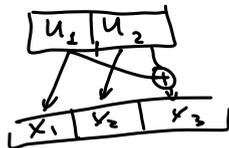
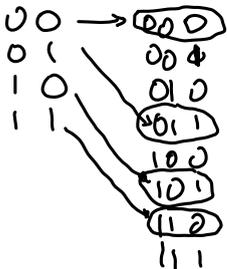
Codice a ripetizione
 $k=1$ $n=3$



$x = u [1 \ 1 \ 1]$

Codice a controllo di parità semplice

$k=2$, $n=3$



$\begin{cases} x_1 = u_1 \\ x_2 = u_2 \\ x_3 = u_1 + u_2 \end{cases}$

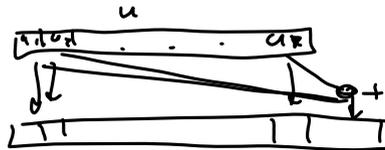
$[x_1 \ x_2 \ x_3] = [u_1 \ u_2] \begin{bmatrix} 1 & 0 & 1 \\ 0 & 1 & 1 \end{bmatrix}$

$G \uparrow \mathbb{F}_2$

Codice a controllo di parità

$(n, k) = (8, 7)$

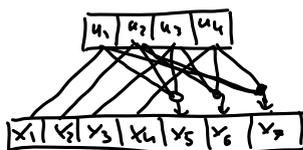
$\begin{cases} x_i = u_i \quad i=1, \dots, 7 \\ x_8 = \sum_{i=1}^7 u_i \end{cases}$



$[x_1 \ \dots \ x_8] =$

$[u_1 \ \dots \ u_7] \begin{bmatrix} 1 & 0 & \dots & 0 & \dots & 0 & 0 & 1 \\ 0 & 1 & \dots & 0 & \dots & 0 & 0 & 1 \\ \vdots & \vdots & \ddots & \vdots & \ddots & \vdots & \vdots & \vdots \\ 0 & 0 & \dots & 1 & \dots & 0 & 0 & 1 \end{bmatrix}$

Codice di Hamming (7, 4)



$\begin{cases} x_i = u_i \quad i=1, \dots, 4 \\ x_5 = u_1 + u_2 + u_3 \\ x_6 = u_2 + u_3 + u_4 \\ x_7 = u_1 + u_2 + u_4 \end{cases}$

$$[x_1, x_2, x_3, x_4, x_5, x_6, x_7] = [u_1, u_2, u_3, u_4] \begin{bmatrix} 1 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 1 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 1 & 0 & 1 & 1 \end{bmatrix}$$

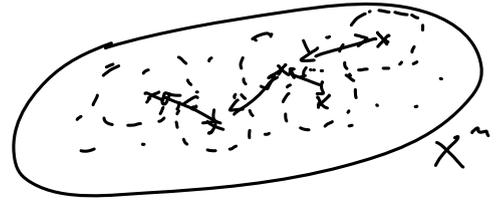
$$G = [I_k : P]$$

↑ Matrice di parità

Property 1 The block code consists of all possible sums of the rows of the generator matrix.

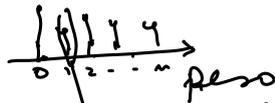
Property 2 The sum of two code words is still a code word.

Property 3 The n-tuple of all zeros is always a code word.



Peso di una parola codice
bit = 1

Distribuzione dei pesi

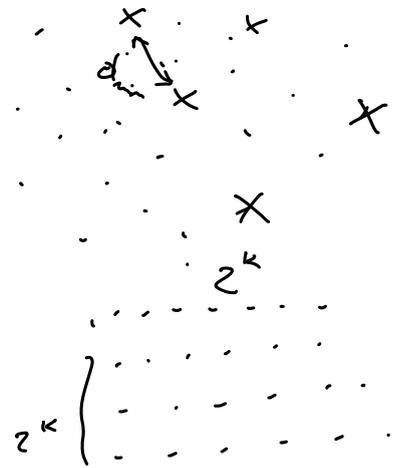


peso di una parola codice è la
distanza della parola 00...0

$d_H(i,j)$ = distanza di Hamming
tra la parola i -esima
e la parola j -esima

d_{min} = minimo peso
delle parole nulle

d_{min}



$$y = x + e$$

$$\begin{matrix} 0 & 1 \\ 1 & 1 \end{matrix}$$

$$\frac{x}{n} = \frac{u}{k} \begin{bmatrix} G_k & n-k \\ I_k & P \end{bmatrix}$$

SINDROME $\underline{s} = \underline{y} H^T = \underline{e} H^T$

$$\begin{matrix} n-k \\ \underline{y} \end{matrix} \begin{bmatrix} P \\ \dots \\ I_{n-k} \end{bmatrix}^k$$

$$H = \begin{bmatrix} P^T & I_{n-k} \\ \dots & \dots \\ P^T & I_{n-k} \end{bmatrix}$$

MATRICE DI
CONTROLLO
AL PARITÀ

$$G H^T = 0$$

$$\begin{matrix} k \\ \boxed{\quad} \end{matrix} \begin{matrix} n-k \\ \boxed{a} \end{matrix}$$

$$k \begin{bmatrix} I_k & P \end{bmatrix} \begin{bmatrix} P \\ \dots \\ I_{n-k} \end{bmatrix} = \begin{bmatrix} 0 \\ \dots \\ 0 \end{bmatrix}$$

$$0 \ 1 \ 1 \ 0$$

Esempio
 $H(7,4)$

$$\begin{cases} x_1 = u_1 \\ x_2 = u_2 \\ x_3 = u_3 \\ x_4 = u_4 \\ \dots \end{cases}$$

$$\begin{matrix} y_1 \\ y_2 \\ y_3 \\ y_4 \end{matrix} \rightarrow \text{codice} \begin{cases} s_1 = (y_1 + y_2 + y_3) + y_5 \\ s_2 = (y_2 + y_3 + y_4) + y_6 \\ s_3 = (y_1 + y_2 + y_4) + y_7 \end{cases}$$

$$\begin{cases} x_3 = u_3 \\ x_4 = u_4 \\ x_5 = u_1 + u_2 + u_3 \\ x_6 = u_2 + u_3 + u_4 \\ x_7 = u_1 + u_2 + u_4 \end{cases} \rightarrow \begin{matrix} y_3 \\ y_4 \\ y_5 \\ y_6 \\ y_7 \end{matrix} \quad (S_3 = (y_1 + y_2 + y_4) + y_7)$$

$$[S_1, S_2, S_3] = [y_1, y_2, y_3, y_4, y_5, y_6, y_7] \begin{bmatrix} 1 & 0 & 1 \\ 1 & 1 & 1 \\ 1 & 1 & 0 \\ 0 & 1 & 1 \\ 1 & 0 & 1 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

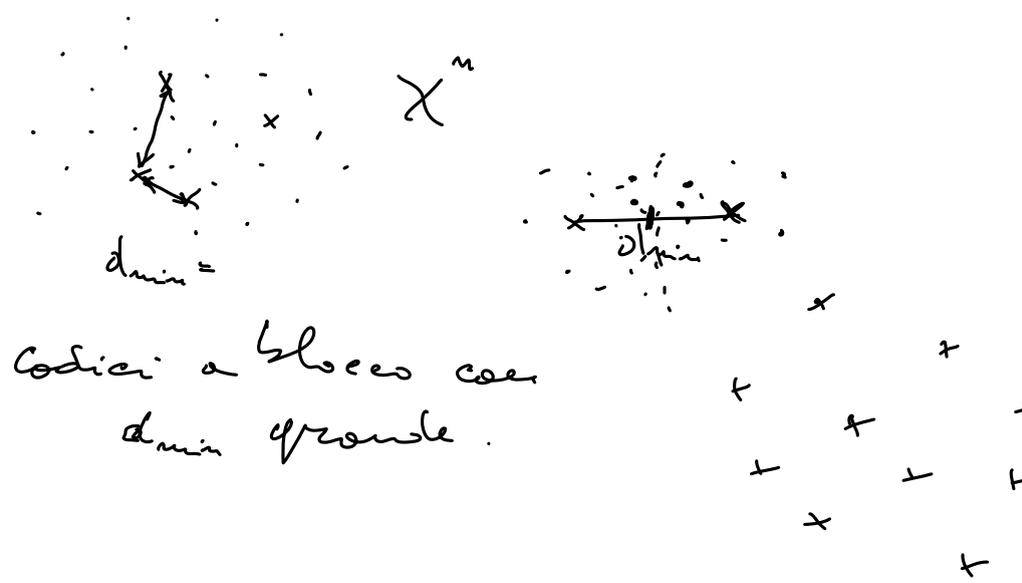
$$\underline{s} = \underline{y} \begin{bmatrix} P \\ \vdots \\ I_{n-k} \end{bmatrix} = \underline{y} H'$$

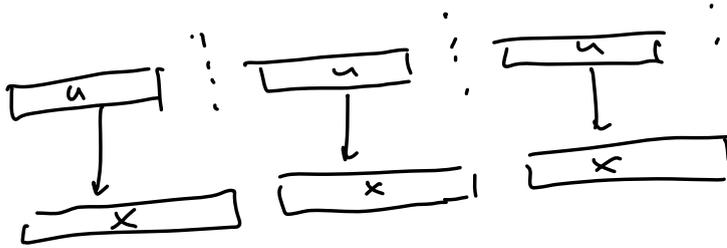
$\underline{s} = \underline{0}$ no problem?
 $\underline{s} \neq \underline{0}$ qualche problema certamente } \rightarrow $\begin{cases} \text{zerillo} \\ \text{corrozzetta} \end{cases}$

2: Standard array of the (7, 4) Hamming code.

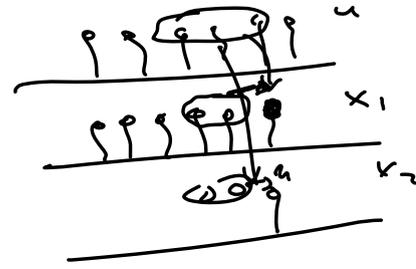
Syndromes	Coset leaders	Error patterns														
		Other errors														
000	0000000	1000101	0100111	0010110	0001011	1100010	1010011	1001110	0110001	0101100	0011101	0111010	1011000	1101001	1110100	1111111
001	0000001	1000100	0100110	0010111	0001010	1100011	1010010	1001111	0110000	0101101	0011100	0111011	1011001	1101000	1110101	1111110
010	0000010	1000111	0100101	0010100	0001001	1100000	1010001	1001100	0110011	0101110	0011111	0111000	1011010	1101011	1110110	1111101
011	0000100	1001101	0101111	0011110	0000011	1101010	1011011	1000110	0111001	0100100	0010101	0110010	1010000	1100001	1111100	1110111
100	0000100	1000001	0100011	0010010	0001111	1100110	1010111	1001010	0110101	0101000	0010011	0111110	1011100	1101101	1110000	1110111
101	1000000	0000101	1100111	1010110	1001011	0100010	0010011	0001110	1110001	1101100	1011101	1111010	0011000	0101001	0110100	0111111
110	0010000	1010101	0110111	0000110	0011011	1110010	1000011	1011110	0100001	0111100	0001101	0101010	1001000	1111001	1100100	1101111
111	0100000	1100101	0000111	0110110	0101011	1000010	1110011	1101110	0010001	0001100	0111101	0011010	1111000	1001001	1010100	1011111

2^{n-k}
 2^{n-k}
 2^7 parole codice
 $p = 0.01 \quad \left(\frac{1}{100}\right)^3 = \frac{1}{100^3}$
 p^3
 $d_H(\underline{y}, \underline{x}_e)$
 $\underline{y} = \underline{x} + \underline{e}$
 $\underline{y} + \underline{e} = \underline{x} \rightarrow \underline{u}$

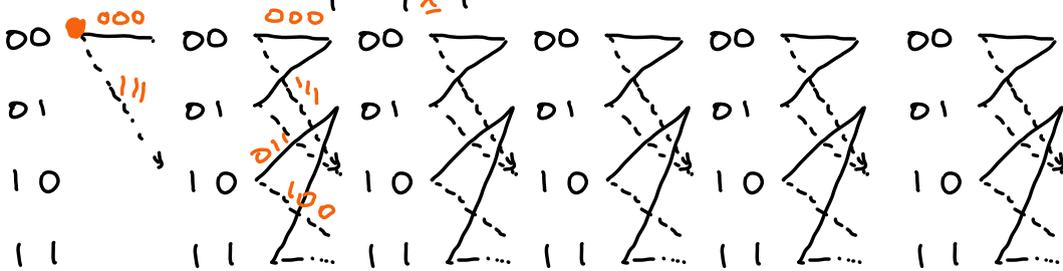
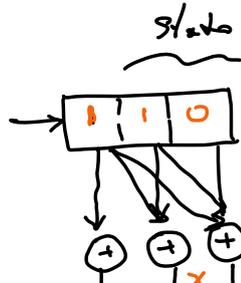




$$\frac{001}{20k} \quad \frac{0}{\dots}$$



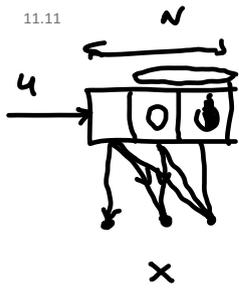
DIN



Z-trasformata

$$\begin{array}{cccc} 3 & 5 & 6 & 1 \\ \hline 0 & 1 & 2 & 3 \end{array} \rightarrow$$

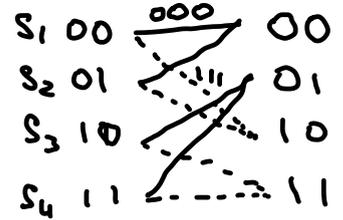
$$3 + 5Z^{-1} + 6Z^{-2} + 1Z^{-3}$$



$n \times k \times N$
 $(3, 1, 3)$

$\frac{K}{M}$

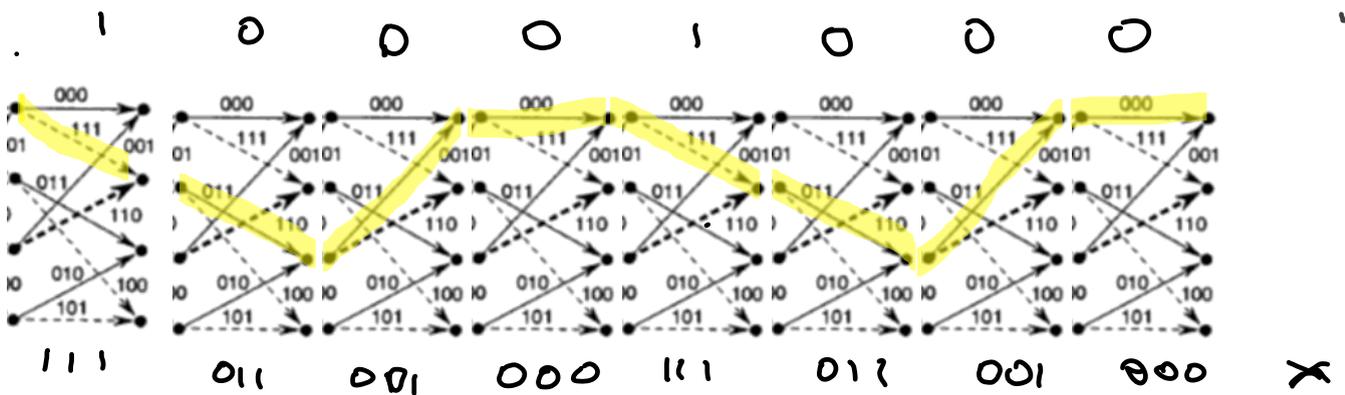
1 0 1



1 1 0 1 1

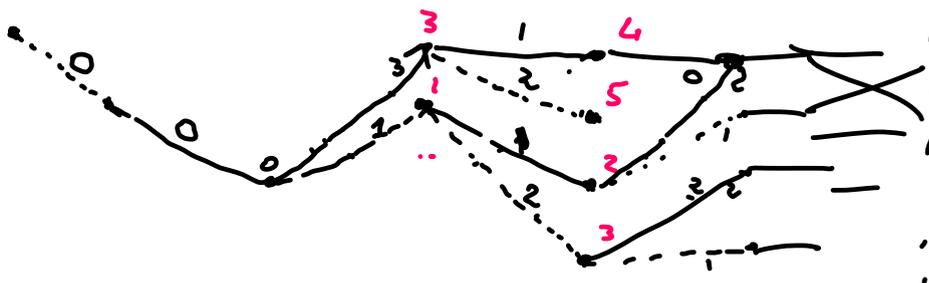
111 100 010 110 100

⇓ Consele (recursion unit)



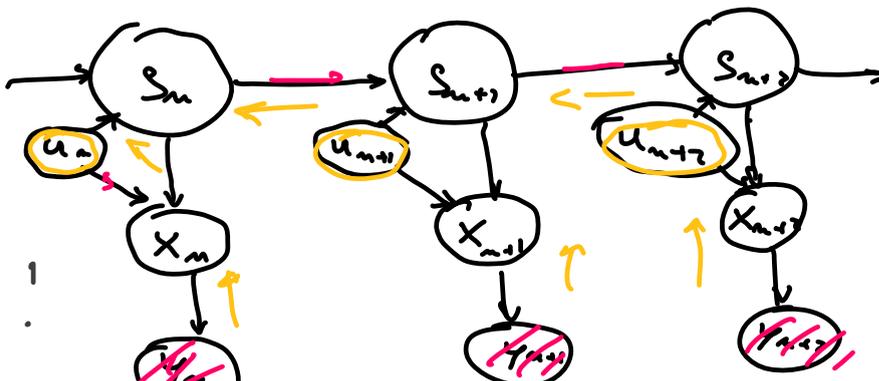
⇓ eseri

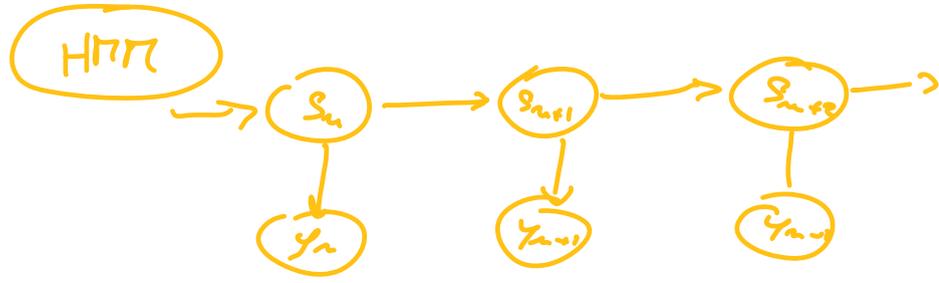
111 011 100 010 111 111 001 000



MODELLO BAYESIANO

Propagazione delle probabilità



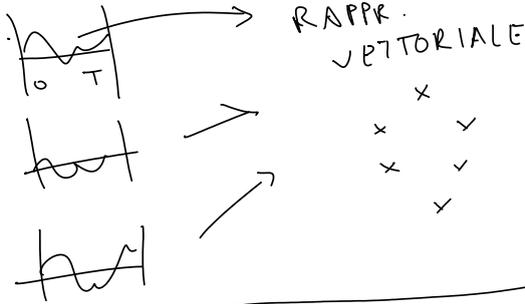
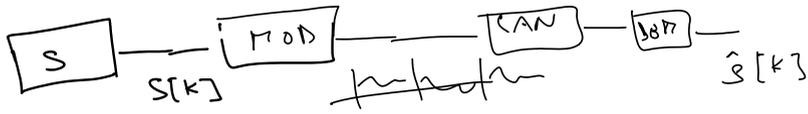


Lezione 16

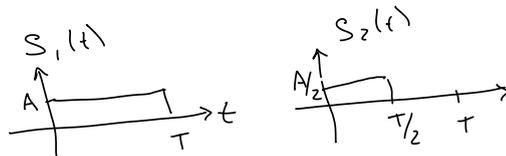
venerdì 27 novembre 2020 10:57

Introduzione alla modulazione numerica

Introduzione allo spazio dei segnali



ESEMPIO



$$\int_0^T S_1(t) S_2(t) dt \neq 0$$

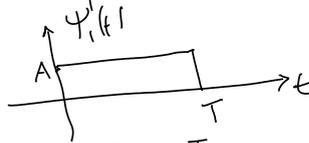
non ortogonali.

$c_1 S_1(t) + c_2 S_2(t) = 0$
 linearmente ind.

$$\| \cdot \| ^2 = \int_0^T (\cdot)^2 dt$$

Cerca base ortonormale di rappresentazione
 Proc. di Gram-Schmidt

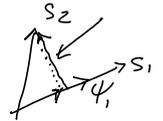
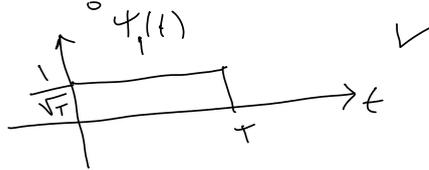
① $\psi'_1(t) = S_1(t)$



$$\psi_1(t) = \frac{\psi'_1(t)}{\sqrt{\|\psi'_1\|^2}}$$

$$\|\psi'_1\|^2 = \int_0^T \psi'_1(t)^2 dt = A^2 T$$

$$\psi_1(t) = \frac{\psi'_1(t)}{A\sqrt{T}}$$

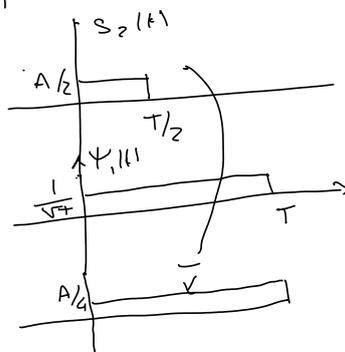
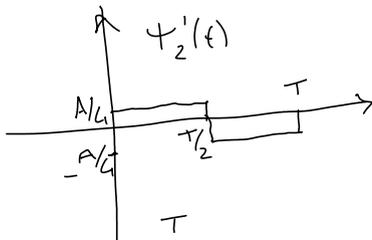


② $\psi'_2(t) = S_2(t) - \langle S_2, \psi_1 \rangle \psi_1(t)$

$$\langle S_2, \psi_1 \rangle = \int_0^T S_2(t) \psi_1(t) dt$$

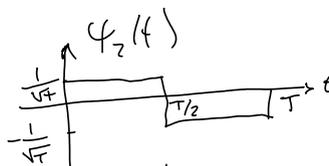
$$= S_2(t) - \frac{A\sqrt{T}}{4} \psi_1(t)$$

$$= \int_0^{T/2} \frac{A}{2} \frac{1}{\sqrt{T}} dt = \frac{A}{2\sqrt{T}} \frac{T}{2} = \frac{A\sqrt{T}}{4}$$



$$\|\psi'_2\|^2 = \int_0^T \frac{A^2}{16} dt = \frac{A^2 T}{16}$$

$$\psi_2(t) = \frac{\psi'_2(t)}{\sqrt{\|\psi'_2\|^2}} = \frac{\psi'_2(t) 4}{A\sqrt{T}}$$



Rappresenta $S_1(t)$ e $S_2(t)$ nella base $(\psi_1(t), \psi_2(t))$

$$S_1(t) \rightarrow \langle S_1, \psi_1 \rangle \psi_1(t) + \langle S_1, \psi_2 \rangle \psi_2(t)$$

$$S_2(t) \rightarrow \langle S_2, \psi_1 \rangle \psi_1(t) + \langle S_2, \psi_2 \rangle \psi_2(t)$$

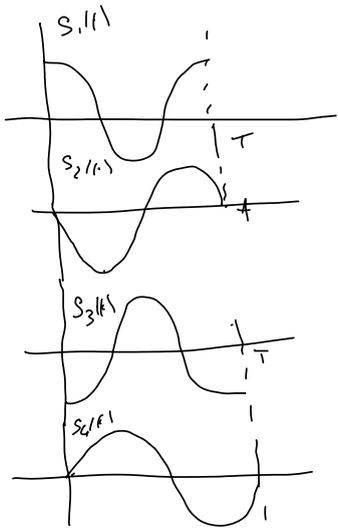
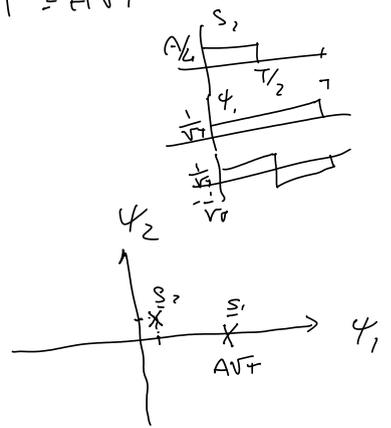
$$(T, 1/2) - A\sqrt{T} = A\sqrt{T}$$

$s_2(t) \rightarrow \langle s_2, \psi_1 \rangle \psi_1(t) + \langle s_2, \psi_2 \rangle \psi_2(t)$
 $\langle s_1, \psi_1 \rangle = \int_0^T s_1(t) \psi_1(t) dt = \int_0^T A \frac{1}{\sqrt{T}} dt = \frac{A}{\sqrt{T}} T = A\sqrt{T}$

$\langle s_1, \psi_2 \rangle = \int_0^T s_1(t) \psi_2(t) dt = 0$

$\langle s_2, \psi_1 \rangle = \int_0^{T/2} A \frac{1}{\sqrt{T}} dt = \frac{A}{\sqrt{T}} \frac{T}{2} = \frac{A\sqrt{T}}{2}$

$\langle s_2, \psi_2 \rangle = \int_0^{T/2} -A \frac{1}{\sqrt{T}} dt = -\frac{A}{\sqrt{T}} \frac{T}{2} = -\frac{A\sqrt{T}}{2}$



$\psi_1(t) = \sqrt{\frac{2}{T}} \cos \frac{2\pi}{T} t$
 $\psi_2(t) = -\sqrt{\frac{2}{T}} \sin \frac{2\pi}{T} t$

$t \in [0, T]$

$\psi_3'(t) = s_3(t) - \langle s_3, \psi_1 \rangle \psi_1(t) - \langle s_3, \psi_2 \rangle \psi_2(t)$

$= 0$

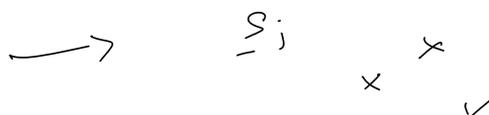
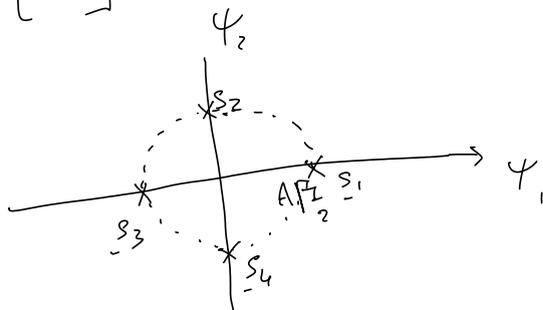
$\psi_4'(t) = s_4(t) - \langle s_4, \psi_1 \rangle \psi_1(t) - \langle s_4, \psi_2 \rangle \psi_2(t)$
 $\langle s_4, \psi_3 \rangle \psi_3(t)$

$s_1 = \begin{bmatrix} \langle s_1, \psi_1 \rangle \\ \langle s_1, \psi_2 \rangle \end{bmatrix} = \begin{bmatrix} \int_0^T s_1(t) \psi_1(t) dt \\ 0 \end{bmatrix} = \begin{bmatrix} \int_0^T A \cos \frac{2\pi}{T} t \sqrt{\frac{2}{T}} \cos \frac{2\pi}{T} t dt \\ 0 \end{bmatrix} = A\sqrt{\frac{2}{T}} \int_0^T \cos^2 \frac{2\pi}{T} t dt$
 $= \begin{bmatrix} A\sqrt{\frac{T}{2}} \\ 0 \end{bmatrix} = A\sqrt{\frac{2}{T}} \frac{T}{2} = A\sqrt{\frac{T}{2}}$

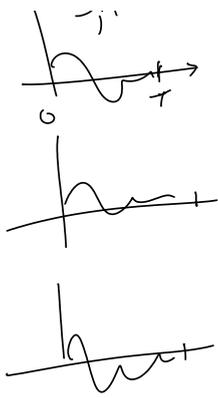
$s_2 = \begin{bmatrix} \langle s_2, \psi_1 \rangle \\ \langle s_2, \psi_2 \rangle \end{bmatrix} = \begin{bmatrix} 0 \\ A\sqrt{\frac{T}{2}} \end{bmatrix}$

$s_3 = \begin{bmatrix} \langle s_3, \psi_1 \rangle \\ \langle s_3, \psi_2 \rangle \end{bmatrix} = \begin{bmatrix} -A\sqrt{\frac{T}{2}} \\ 0 \end{bmatrix}$

$s_4 = \begin{bmatrix} 0 \\ -A\sqrt{\frac{T}{2}} \end{bmatrix}$
 PSK

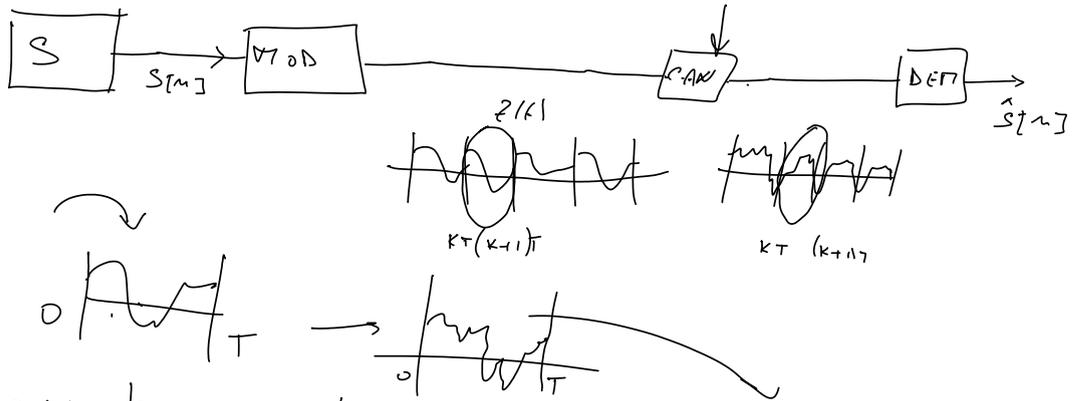


SPAZIO DEI



\rightarrow $=$ \times \times \times \times
 \times \times \times \times \times
 base
 ortogonale \times \times \times

SINCRONISMO
DEI
SEGNALI



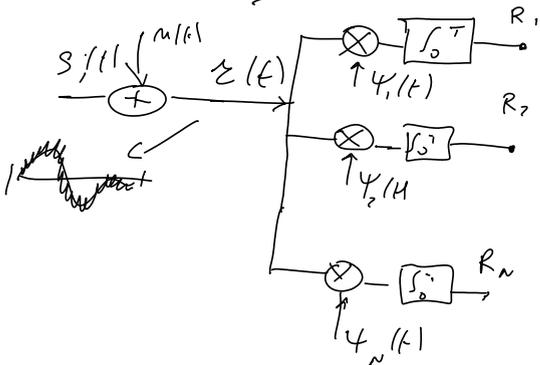
$a_1 \rightarrow s_1(t)$ \rightarrow \rightarrow \rightarrow
 $a_2 \rightarrow s_2(t)$ \rightarrow \rightarrow \rightarrow
 $a_n \rightarrow s_n(t)$ \rightarrow \rightarrow

$$z(t) = s_j(t) + n(t)$$

DESCRITTI
IN UNA BASE ORTONORMALE
COMPLETA PER TUTTI I SEGNALI

$$\{\psi_1(t), \psi_2(t), \dots, \psi_N(t)\} \quad N \leq M$$

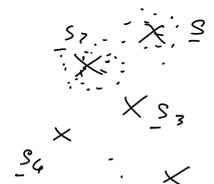
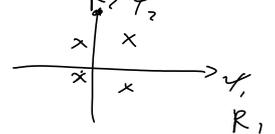
~~Non c'è rumore~~



$$\underline{R} = \begin{bmatrix} \int_0^T s_j(t) \psi_1(t) dt \\ \vdots \\ \int_0^T s_j(t) \psi_N(t) dt \end{bmatrix} = \underline{S}_j = \begin{bmatrix} s_{j1} \\ s_{j2} \\ \vdots \\ s_{jN} \end{bmatrix}$$

ESEMPIO

$M=4, N=2$
 $\{\psi_1(t), \psi_2(t)\}$



ADDITIVE WHITE GAUSSIAN NOISE - CHANNEL $\sum_{i=1}^M S_i$

DECISORE: $\underline{R} \rightarrow \hat{S}$

BASATO SU PARTIZIONE S_1, S_2, \dots, S_M : $\bigcup_{j=1}^M S_j = \mathbb{R}^N$, $S_i \cap S_j = \emptyset$

TROVARE LA PARTIZIONE OTTIMA

$$\begin{aligned}
 P(e) &= P\{\hat{S} \neq S\} = 1 - P(e) = 1 - P\{\hat{S} = S\} \\
 &= 1 - \sum_{j=1}^M P\{e | s = a_j\} \pi_j = 1 - \sum_{j=1}^M P\{\underline{R} \in S_j | s = a_j\} \pi_j \\
 &= 1 - \sum_{j=1}^M \int_{S_j} f(\underline{r} | a_j) d\underline{r} \pi_j
 \end{aligned}$$

Le regioni ottimali sono quelle che minimizzano questo secondo

$$S_j \text{ ottimali} = \left\{ \underline{r} : \int_{S_j} f(\underline{r} | a_j) \pi_j > \int_{S_i} f(\underline{r} | a_i) \pi_i, \forall i \neq j \right\}$$

CRITERIO MAP (MASSIMA PROB. A POSTERIORI)

$$\frac{\int_{S_j} f(\underline{r} | a_j) \pi_j}{\int_{S_j} f(\underline{r})} > \frac{\int_{S_i} f(\underline{r} | a_i) \pi_i}{\int_{S_i} f(\underline{r})}$$

$$P(a_j | \underline{r}) > P(a_i | \underline{r})$$

PROB. A POSTERIORI

SE $\pi_j = \pi_i = \frac{1}{M}$ Prob. a priori uniformi.

$$\int_{S_j} f(\underline{r} | a_j) > \int_{S_i} f(\underline{r} | a_i)$$

MASSIMA VEROSIMILIANZA

(ML MAXIMUM LIKELIHOOD)

USIAMO L'IPOTESI GAUSSIANA

$$\int_{S_j} f(\underline{r} | a_j) = N\left(\underline{r}; \underline{s}_j, \frac{\sigma^2}{2} \underline{I}_N\right) \quad j = 1, \dots, M$$

funzione N-dimensionale.

$$N(\underline{r}, \underline{\mu}, \underline{\Sigma}) = \frac{1}{(2\pi)^{N/2} |\underline{\Sigma}|^{1/2}} e^{-\frac{1}{2}(\underline{r}-\underline{\mu})^T \underline{\Sigma}^{-1} (\underline{r}-\underline{\mu})}$$



$N(\underline{x}; \underline{\mu}, \sigma^2 \mathbf{I}_N)$ ($N \times 1$) $(N \times N)$



$$\text{cov} \underline{x} = \sigma^2 \mathbf{I}_N$$



$$= \frac{1}{(2\pi)^{N/2} \sqrt{\sigma^2}^N} e^{-\frac{1}{2\sigma^2} (\underline{x} - \underline{\mu})^T (\underline{x} - \underline{\mu})}$$

$$= \prod_{i=1}^N \frac{1}{\sqrt{2\pi\sigma^2}} e^{-\frac{1}{2\sigma^2} (x_i - \mu_i)^2}$$

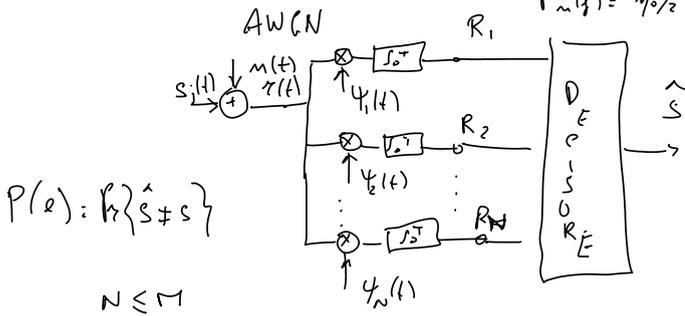
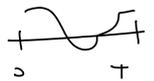
$$\int_{\mathbb{R}^N} (\underline{x} | \alpha_j) \pi_j > \int_{\mathbb{R}^N} (\underline{x} | \alpha_i) \pi_i$$

$$N(\underline{x}; \underline{s}_j, \frac{\gamma_0}{2} \mathbf{I}_N) \pi_j > N(\underline{x}; \underline{s}_i, \frac{\gamma_0}{2} \mathbf{I}_N)$$

$$\prod_{\ell=1}^N \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0} (x_\ell - s_{j\ell})^2} \pi_j > \prod_{\ell=1}^N \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0} (x_\ell - s_{i\ell})^2} \pi_i$$

$m(t)$ bianco $R_n(\tau) = \eta_0 \delta(\tau)$
 $P_n(f) = \eta_0/2$

$\{s_1(t) s_2(t) \dots s_M(t)\}$
 $a_1 a_2 \dots a_M$



$P(e) = P\{\hat{s} \neq s\}$

$N \leq M$

$\min P(e) \leftrightarrow \max P(c) \leftrightarrow \text{MAP}$

scegli a_j se $\int_{R^-} (\pi_j | a_j) \pi_j > \int_{R^-} (\pi_i | a_i) \pi_i$

$\frac{1}{\sqrt{\pi \eta_0}} e^{-\frac{(x_e - s_{je})^2}{\eta_0}} \pi_j > \frac{1}{\sqrt{\pi \eta_0}} e^{-\frac{(x_e - s_{ie})^2}{\eta_0}} \pi_i$

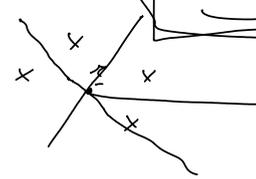
$\prod_{e=1}^N \frac{1}{\sqrt{\pi \eta_0}} e^{-\frac{(x_e - s_{je})^2}{\eta_0}} \pi_j > \prod_{e=1}^N \frac{1}{\sqrt{\pi \eta_0}} e^{-\frac{(x_e - s_{ie})^2}{\eta_0}} \pi_i$

\ln
 $-\frac{1}{\eta_0} \sum_{e=1}^N (x_e - s_{je})^2 + \ln \pi_j > -\frac{1}{\eta_0} \sum_{e=1}^N (x_e - s_{ie})^2 + \ln \pi_i$

NOTA: se $\pi_i = \frac{1}{M} \quad i=1..M \quad (ML)$
 $-\sum_{e=1}^N (x_e - s_{je})^2 > -\sum_{e=1}^N (x_e - s_{ie})^2$
 $\sum_{e=1}^N (x_e - s_{je})^2 < \sum_{e=1}^N (x_e - s_{ie})^2$

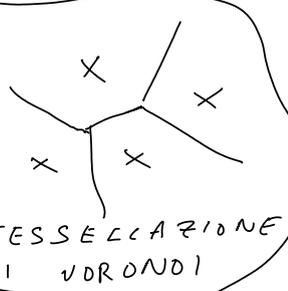
distance euclidea

DEFSORE MINIMA DISTANZA



$N(x; \mu, \Sigma)$
 $= \frac{1}{(2\pi)^{N/2} |\Sigma|^{1/2}} e^{-\frac{1}{2}(x-\mu)^T \Sigma^{-1} (x-\mu)}$
 $\Sigma = \sigma^2 I_N$
 $= \begin{pmatrix} \sigma^2 & & \\ & \ddots & \\ & & \sigma^2 \end{pmatrix}$

$N(x; \mu, \sigma^2 I_N)$
 $= \frac{1}{(2\pi)^{N/2} \sigma^N} e^{-\frac{1}{2\sigma^2} \sum_{e=1}^N (x_e - \mu_e)^2}$
 $= \prod_{e=1}^N \frac{1}{\sqrt{2\pi} \sigma} e^{-\frac{1}{2\sigma^2} (x_e - \mu_e)^2}$



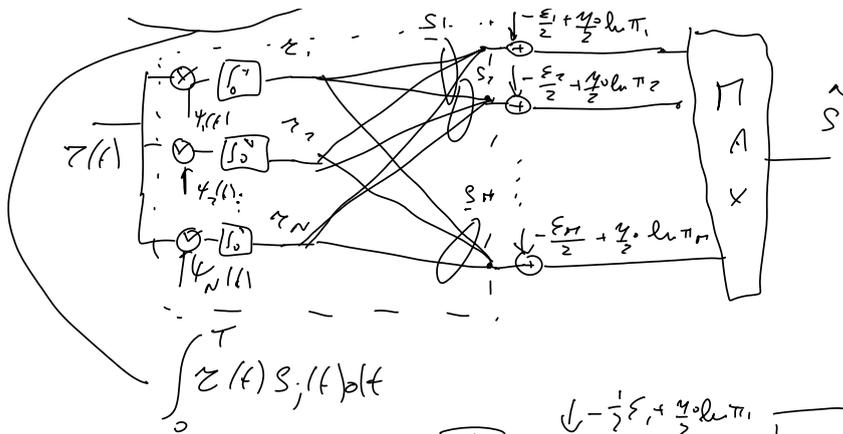
MAP

MV (ML)

$-\frac{1}{\eta_0} \sum_{e=1}^N (x_e^2 + s_{je}^2 - 2x_e s_{je}) + \ln \pi_j > -\frac{1}{\eta_0} \sum_{e=1}^N (x_e^2 + s_{ie}^2 - 2x_e s_{ie}) + \ln \pi_i$
 $\frac{2}{\eta_0} \sum_{e=1}^N x_e s_{je} - \frac{1}{\eta_0} \sum_{e=1}^N s_{je}^2 + \ln \pi_j > \frac{2}{\eta_0} \sum_{e=1}^N x_e s_{ie} - \frac{1}{\eta_0} \sum_{e=1}^N s_{ie}^2 + \ln \pi_i$

$\sum_{e=1}^N x_e s_{je} - \frac{1}{2} \epsilon_j + \frac{\eta_0}{2} \ln \pi_j > \sum_{e=1}^N x_e s_{ie} - \frac{1}{2} \epsilon_i + \frac{\eta_0}{2} \ln \pi_i$

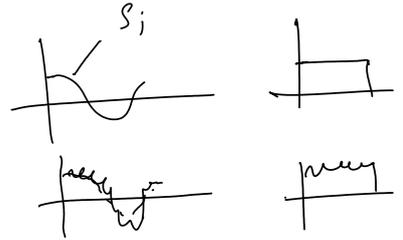
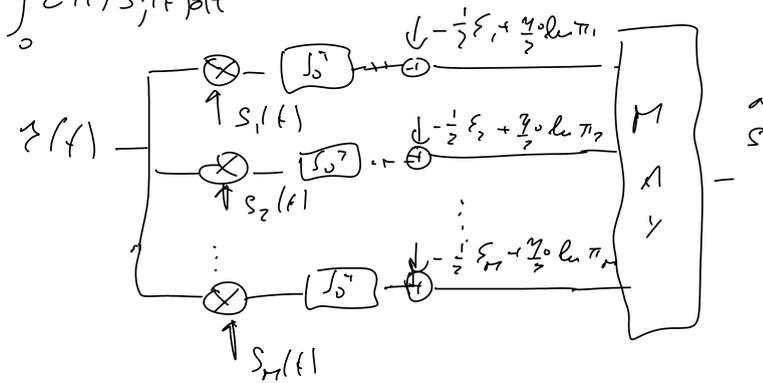
Spazio dei segnali
ordinario
 $\int s_j^2(t) dt = \|s_j\|^2$



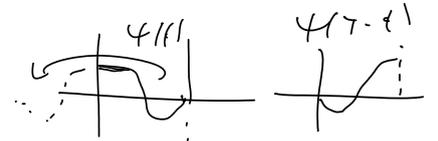
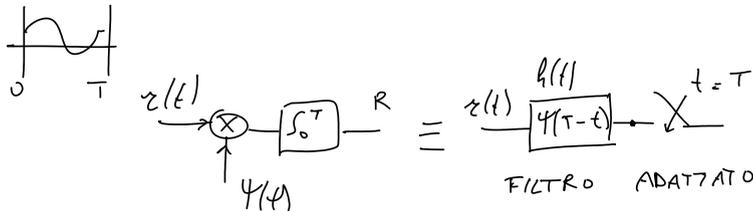
$$\int_0^T s_j^2(t) dt = \|s_j\|^2$$

$$E_j \leftarrow$$

RICE OTTIMO

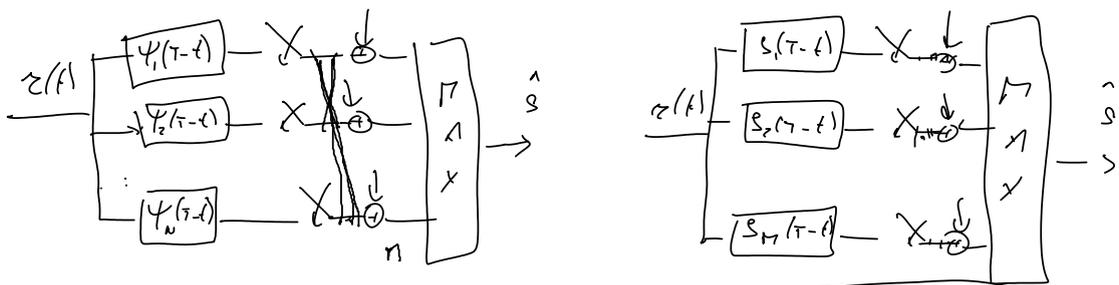


$$\psi(T-t) = \psi(-(t-T))$$



$$\int_0^T z(\tau) h(t-\tau) d\tau \Big|_{t=T} = \int_0^T z(\tau) \psi(T-t+\tau) d\tau \Big|_{t=T}$$

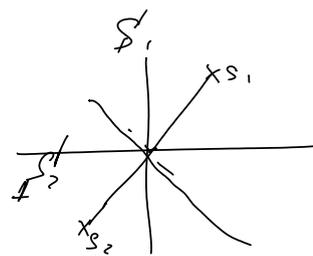
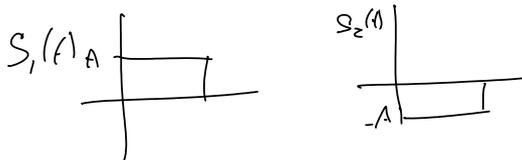
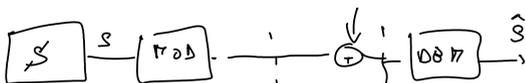
$$= \int_0^T z(\tau) \psi(\tau) d\tau$$



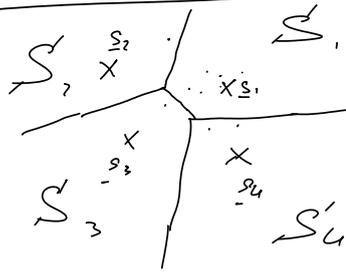
CASO BINARIO

$$S \in \{a_1, a_2\}$$

$$\pi \in \{p, 1-p\}$$

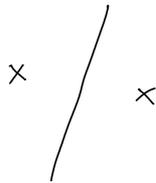


→ casi particolari → ric. semplici

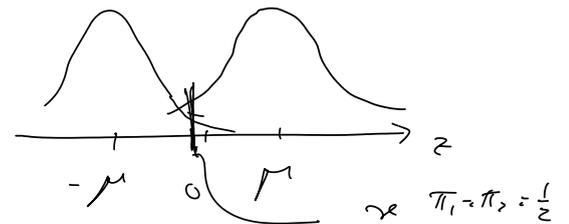


$$\begin{aligned}
 P(e) &= 1 - P(\bar{e}) = 1 - \sum_{i=1}^{\pi} P(e|a_i) \pi_i \\
 &= 1 - \sum_{i=1}^{\pi} \int_{S'_i} f_z(z|a_i) dz \pi_i \\
 &= 1 - \sum_{i=1}^{\pi} \int_{S'_i} N\left(z; \xi_i, \frac{\gamma_0}{2} I_n\right) dz \pi_i
 \end{aligned}$$

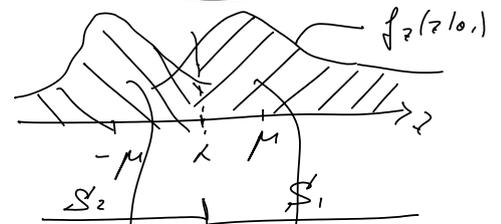
CASO BINARIO



$$\underbrace{\langle z, \xi_1 - \xi_2 \rangle - \frac{1}{2}(\xi_1 - \xi_2)}_z \quad \sum_{\alpha_1}^{\alpha_2} \underbrace{\frac{\gamma_0}{2} dz \frac{\pi_2}{\pi_1}}_{\text{repla } \lambda}$$



$$\begin{aligned}
 f_z(z|a_1) &= N\left(z; \frac{1}{2}d_1^2, \frac{\gamma_0}{2}d_1^2\right) \\
 f_z(z|a_2) &= N\left(z; -\frac{1}{2}d_2^2, \frac{\gamma_0}{2}d_2^2\right)
 \end{aligned}$$

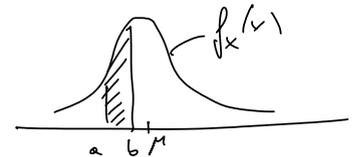
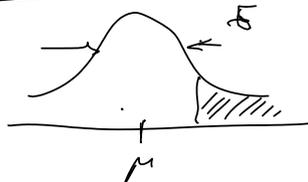


$$P(e) = 1 - P(\bar{e}) = 1 - P(e|a_1)\pi_1 - P(e|a_2)\pi_2$$

$$\begin{aligned}
 &= 1 - \int_{S'_1} f_z(z|a_1) dz \pi_1 - \int_{S'_2} f_z(z|a_2) dz \pi_2 \\
 &= 1 - \int_{-\infty}^{\infty} N(z; \mu, \sigma^2) dz \pi_1 - \int_{-\infty}^{\infty} N(z; -\mu, \sigma^2) dz \pi_2 =
 \end{aligned}$$

$$\begin{aligned}
 &Q\left(\frac{\lambda - \mu}{\sigma}\right) \\
 &= 1 - Q\left(\frac{\lambda + \mu}{\sigma}\right)
 \end{aligned}$$

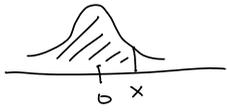
$$\begin{aligned}
 &N(x; \mu, \sigma^2) \\
 &= \frac{1}{\sqrt{2\pi}\sigma} e^{-\frac{(x-\mu)^2}{2\sigma^2}}
 \end{aligned}$$



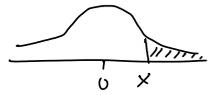
$$\begin{aligned}
 P\{a < X < b\} &= \int_a^b f_x(x) dx = \Phi\left(\frac{b-\mu}{\sigma}\right) - \Phi\left(\frac{a-\mu}{\sigma}\right) \\
 &= Q\left(\frac{a-\mu}{\sigma}\right) - Q\left(\frac{b-\mu}{\sigma}\right)
 \end{aligned}$$

$N(x, 0, 1)$ funzione normalizzata

$$\phi(x) \triangleq \int_{-\infty}^x N(\xi, 0, 1) d\xi = 1 - Q(|x|)$$

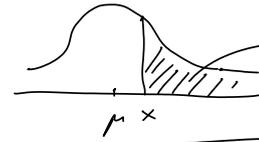


$$\Phi(x) \triangleq \int_{-\infty}^x \mathcal{N}(\xi, 0, 1) d\xi = 1 - Q(|x|)$$

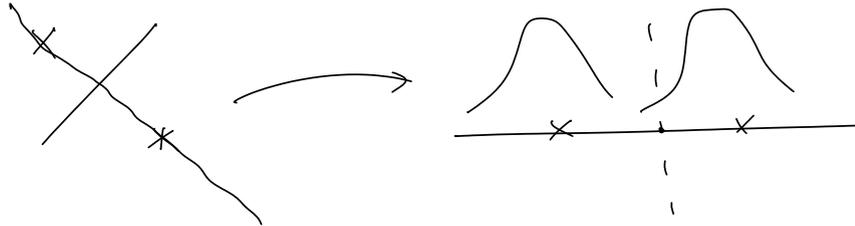


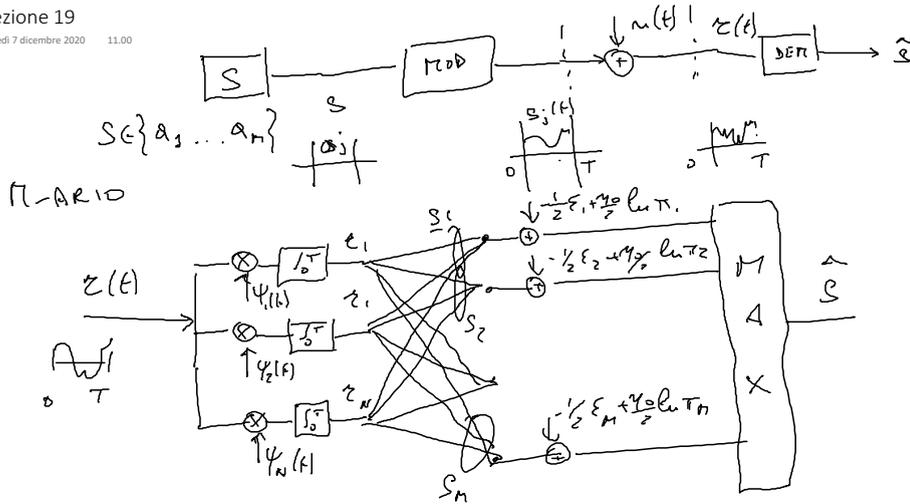
$$Q(x) = \int_x^{\infty} \mathcal{N}(\xi, 0, 1) d\xi = 1 - \Phi(x)$$

$$= Q\left(\frac{a-\mu}{\sigma}\right) - Q\left(\frac{b-\mu}{\sigma}\right)$$

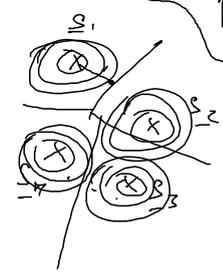


$$Q\left(\frac{x-\mu}{\sigma}\right)$$





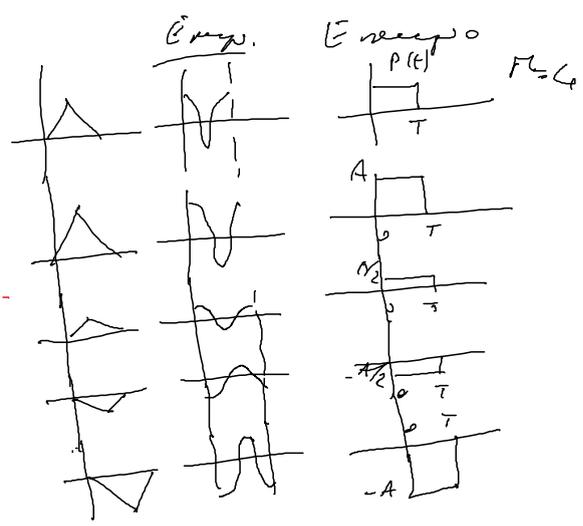
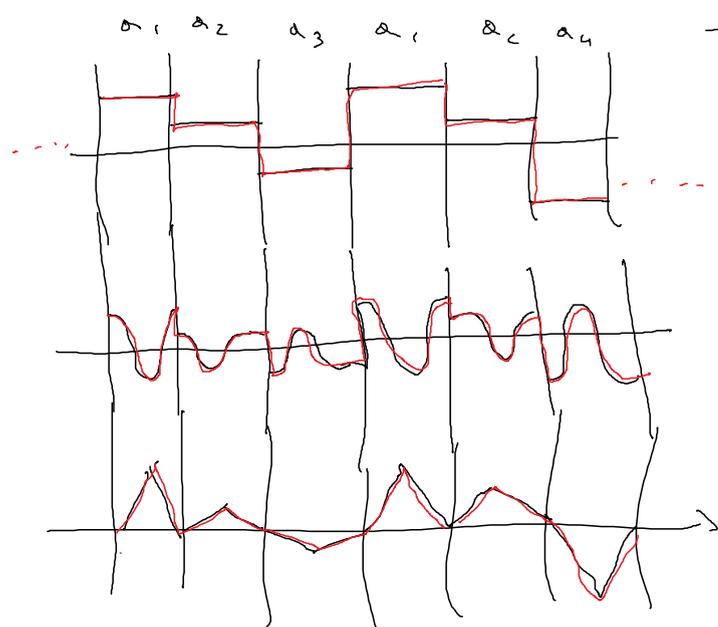
$n(t)$ processo bianco
 $R_n(\tau) = \frac{N_0}{2} \delta(\tau)$
 $P_n(f) = \frac{N_0}{2}$



MODULAZIONE PAM (PULSE AMPLITUDE MODULATION)

$\mathcal{A} = \{a_1, \dots, a_M\} \rightarrow \{s_1(t), s_2(t), \dots, s_M(t)\}$

$s_j(t) = \sum_j p(t)$

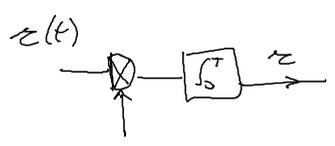


$\omega_j = -\frac{\pi}{2} + j \Delta \quad j = 1, \dots, M$

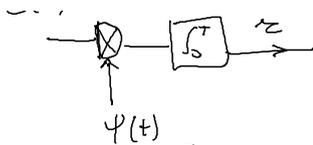
$M=4$
 $-\frac{\pi}{2} + j \Delta$
 $M=7$

$s_j(t) = \sum_j p(t)$

$\psi'(t) = p(t) ; \quad \psi(t) = \frac{\psi'(t)}{\sqrt{\|\psi'\|^2}} = \frac{p(t)}{\sqrt{E_p}}$

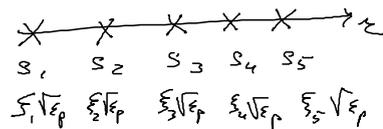


$M=5$



$$s_j = \int_0^T s_j(t) \psi(t) dt = \sum_j \int_0^T p(t) \frac{p(t)}{\sqrt{\epsilon_p}} dt$$

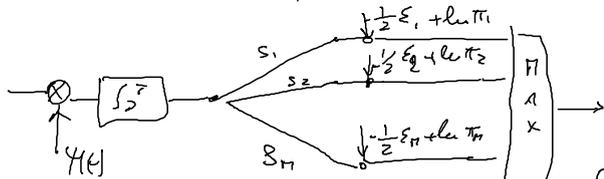
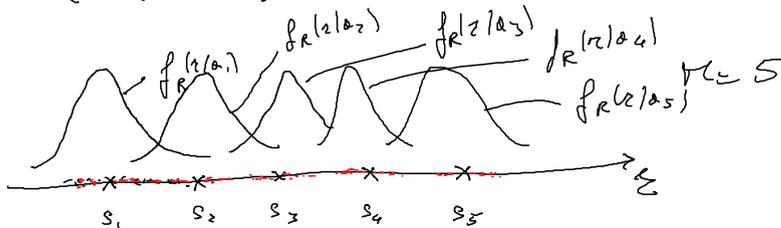
$$= \sum_j \sqrt{\epsilon_p}$$



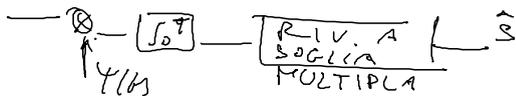
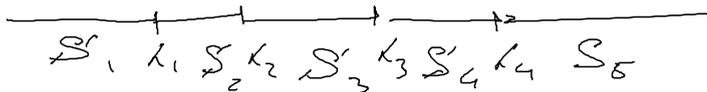
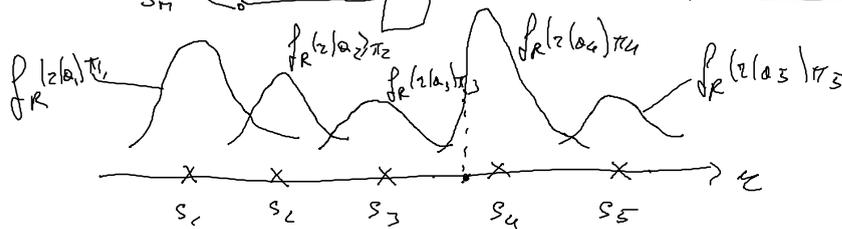
$$z/\alpha_j = \int_0^T z(t) \psi(t) dt = \int_0^T (s_j(t) + u(t)) \psi(t) dt = \frac{\sum_j \sqrt{\epsilon_p}}{s_j} + \int_0^T u(t) \psi(t) dt$$

N processi, media nulla, varianza $\gamma/2$

$$f(z/\alpha_j) = N(z; s_j, \gamma/2) \quad j = 1 \dots M$$

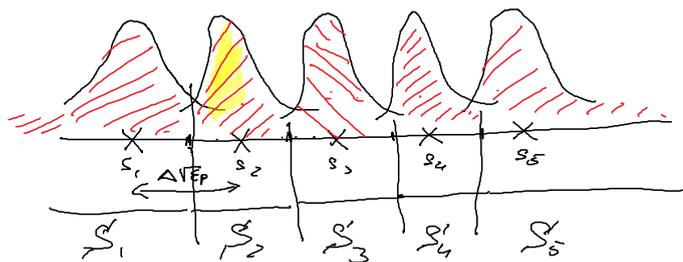


$$f_R(z/\alpha_j) \pi_j > f_R(z/\alpha_i) \pi_i$$



Ric. a P.V. $\pi_j = \frac{1}{M} \quad j = 1 \dots M$

$$f_R(z/\alpha_j) > f_R(z/\alpha_i)$$



copie uniformi

$$P(z) = 1 - P(c) = 1 - \sum_{i=1}^M P(c|\alpha_i) \pi_i = 1 - \sum_{i=1}^M \int_{S'_i} f_R(z/\alpha_i) dz \pi_i$$

$$= 1 - \sum_{i=1}^M \int_{S'_i} N(z; s_i, \gamma/2) dz \pi_i$$

$$= 1 - (\pi_1 + \pi_M) \int N(z; s_1, \gamma/2) - (\pi_2 + \pi_3 \dots + \pi_{M-1}) \int N(z; s_2, \gamma/2) dz$$

$$= \int -(\pi_1 + \pi_M) \int N(z; s_1, \gamma_0/2) - (\pi_2 + \pi_3 + \dots + \pi_{M-1}) \int N(z; s_2, \gamma_0/2) dz$$

$$1 - Q\left(\frac{\sqrt{E_p} \Delta}{2 \sqrt{\gamma_0/2}}\right) = 1 - Q\left(\Delta \sqrt{\frac{E_p}{2\gamma_0}}\right)$$

$$1 - 2Q\left(\frac{\sqrt{E_p} \Delta}{2 \sqrt{\gamma_0/2}}\right)$$

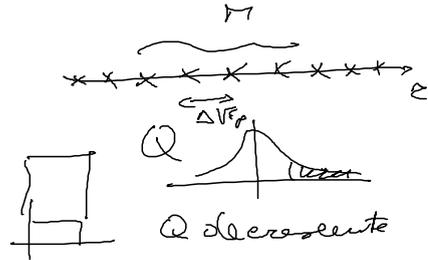
$$1 - 2Q\left(\Delta \sqrt{\frac{E_p}{2\gamma_0}}\right)$$

$$P(e) = 1 - (\pi_1 + \pi_M)(1 - Q(\cdot)) - (\pi_2 + \dots + \pi_{M-1})(1 - 2Q(\cdot))$$

$$\pi_i = \frac{1}{M}$$

$$= 1 - \frac{2}{M}(1 - Q(\cdot)) - \frac{M-2}{M}(1 - 2Q(\cdot))$$

$$P(e) = 2\left(1 - \frac{1}{M}\right) Q\left(\Delta \sqrt{\frac{E_p}{2\gamma_0}}\right)$$



E_p, Δ
 M, T
 γ_0



$$B_{eq} = \frac{\lg M}{T}$$

$\lg_2 M$

Energia non è uniforme
→ segnali non equienergici.

$$\frac{\bar{E}_b}{\gamma_0} = \frac{\text{Energia per bit}}{\gamma_0}$$

$$\Sigma_j = \int_0^T S_j(t) dt = \Sigma_j^2 \int_0^T p(t) dt = \Sigma_j^2 E_p$$

$$\Sigma_j = -\frac{\pi+1}{2} \Delta + j\Delta$$

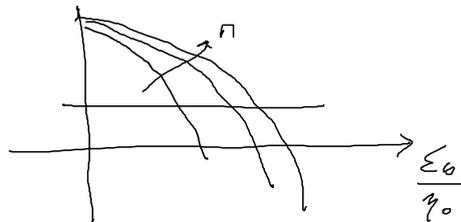
$j = 1 \dots M$

$$\bar{\Sigma} = \frac{1}{M} \sum_{j=1}^M \Sigma_j = \frac{E_p}{M} \sum_{j=1}^M \Sigma_j^2 = \dots = \frac{E_p \Delta^2 (M^2 - 1)}{12}$$

$$\frac{\bar{E}_b}{\gamma_0} = \frac{\bar{\Sigma}}{\lg_2 M \gamma_0} = \frac{E_p \Delta^2 (M^2 - 1)}{12 \lg_2 M \gamma_0} \quad P(e)$$

...

$$P(e) = 2\left(1 - \frac{1}{M}\right) Q\left(\sqrt{\frac{6 \lg_2 M}{M^2 - 1}} \frac{E_b}{\gamma_0}\right)$$



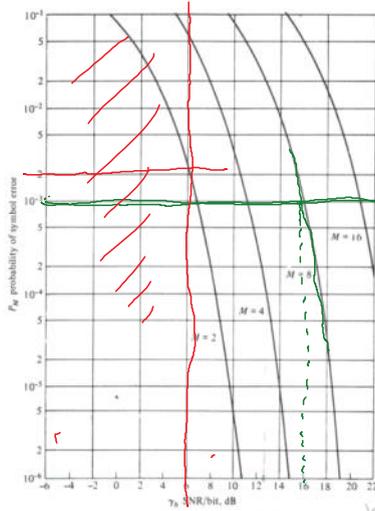


Figure 4.2.24 Probability of a symbol error for PAM.

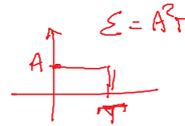
PAM

$M=8 \quad T=1 \mu\text{sec}$

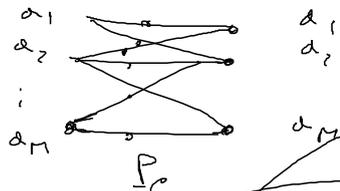
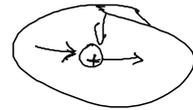
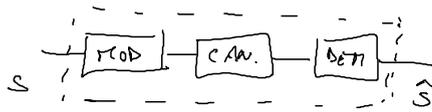
$B_z = \frac{\log_2 M}{T} = \frac{3}{10^{-6}} = 3 \cdot 10^6 = 3 \text{ Mcbit/sec.}$

$P(e) < 10^{-3}$

$\frac{E_b}{N_0} > 16 \text{ dB}$

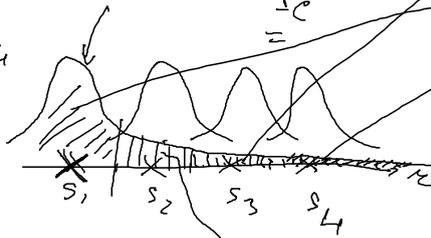


$\sum_j = \int_0^T s_j^2 dt$



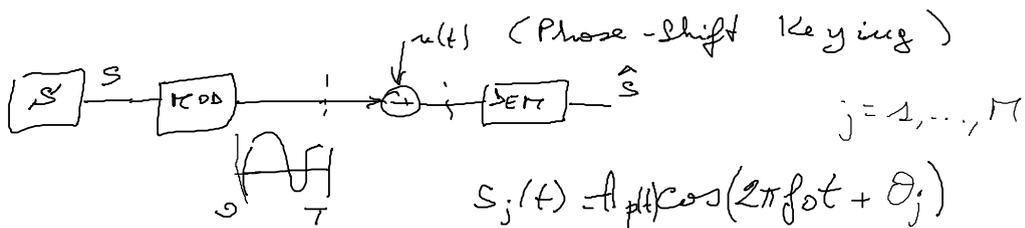
P_e

M=4
MV



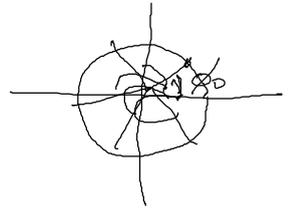
$P_e = \left\{ \begin{matrix} P_2 \{ \hat{s} = a_1 | s = a_1 \} P_1 \{ \hat{s} = a_2 | s = a_1 \} P_2 \{ \hat{s} = a_3 | s = a_1 \} P_1 \{ \hat{s} = a_4 | s = a_1 \} \\ P_1 \{ \hat{s} = a_1 | s = a_2 \} P_2 \{ \hat{s} = a_2 | s = a_2 \} P_1 \{ \hat{s} = a_3 | s = a_2 \} P_2 \{ \hat{s} = a_4 | s = a_2 \} \\ \dots \\ P_1 \{ \hat{s} = a_1 | s = a_4 \} P_2 \{ \hat{s} = a_2 | s = a_4 \} P_1 \{ \hat{s} = a_3 | s = a_4 \} P_2 \{ \hat{s} = a_4 | s = a_4 \} \end{matrix} \right\}$

MODULAZIONE PSK



Base ?

$$\theta_j = \phi_0 + (j-1) \frac{2\pi}{M}$$



$$s_j(t) = A p(t) (\cos \theta_j \cos 2\pi f_0 t - \sin \theta_j \sin 2\pi f_0 t)$$

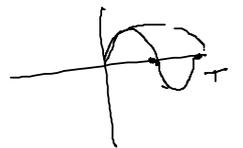
$$\psi_1'(t) = A p(t) \cos 2\pi f_0 t$$

$$\psi_2'(t) = -A p(t) \sin 2\pi f_0 t$$

$$\int \psi_1'(t) \psi_2'(t) dt = -\int_0^T A^2 p^2(t) \cos 2\pi f_0 t \sin 2\pi f_0 t dt$$

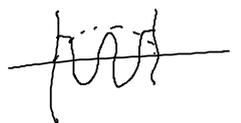
$$= A^2 \int_0^T \frac{1}{2} \sin 4\pi f_0 t p^2(t) dt = 0$$

$$f_0 = \frac{K}{T}$$



$$\langle \psi_1', \psi_1' \rangle = \int_0^T A^2 p^2(t) \cos^2 2\pi f_0 t dt$$

$$f_0 = \frac{K}{T}$$



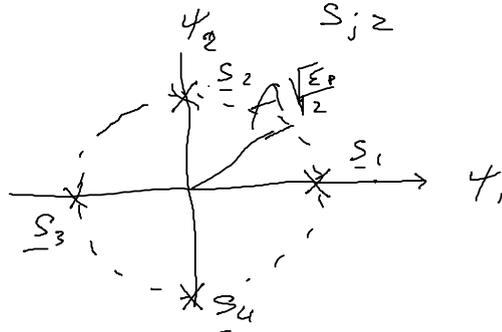
$$= A^2 \left[\frac{1}{2} \int_0^T p^2(t) dt + \frac{1}{2} \int_0^T p^2(t) \cos 4\pi f_0 t dt \right] = \frac{A^2}{2} E_p$$

$$\psi_1(t) = \frac{\sqrt{2} A p(t)}{A \sqrt{E_p}} \cos 2\pi f_0 t = \sqrt{\frac{2}{E_p}} p(t) \cos 2\pi f_0 t$$

$$\psi_2(t) = -\sqrt{\frac{2}{E_p}} p(t) \sin 2\pi f_0 t$$

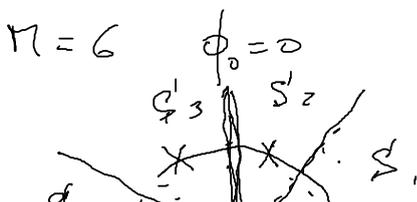
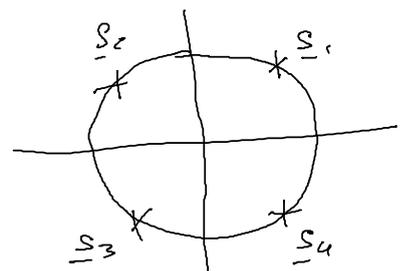
$$s_j(t) = A \sqrt{\frac{E_p}{2}} \cos \theta_j \sqrt{\frac{2}{E_p}} p(t) \cos 2\pi f_0 t - A \sqrt{\frac{E_p}{2}} \sin \theta_j \sqrt{\frac{2}{E_p}} p(t) \sin 2\pi f_0 t$$

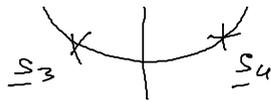
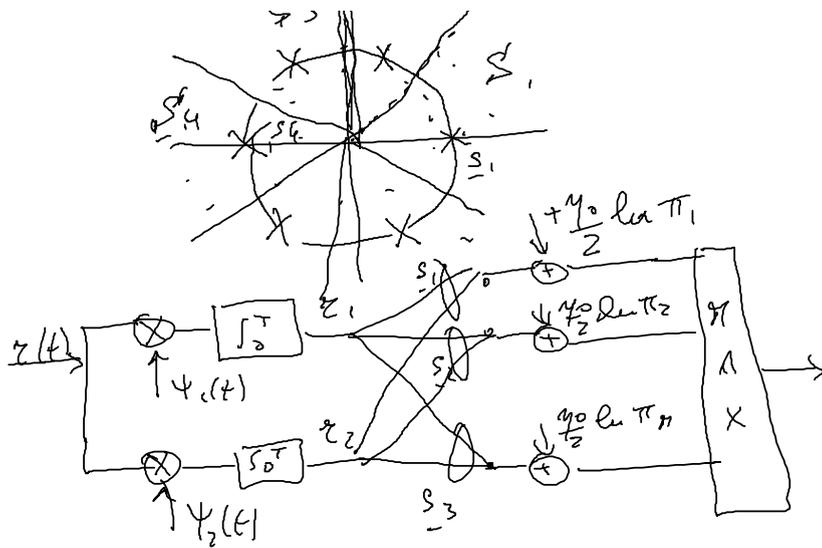
$$= \underbrace{A \sqrt{\frac{E_p}{2}} \cos \theta_j}_{s_{j1}} \psi_1(t) + \underbrace{A \sqrt{\frac{E_p}{2}} \sin \theta_j}_{s_{j2}} \psi_2(t)$$



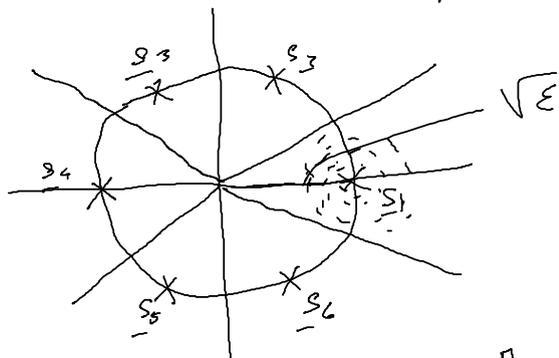
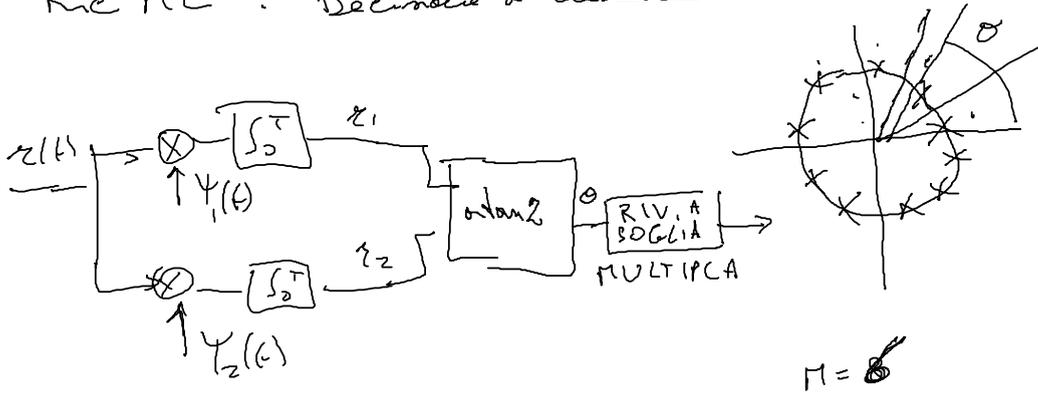
$$M=4, \phi_0=0$$

$$M=4, \phi_0=\pi/4$$





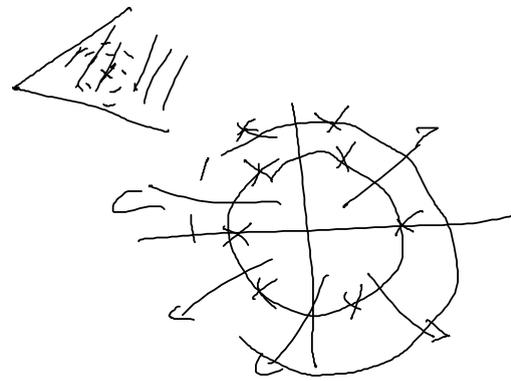
Rice ML: Decisione a minima distanza



$$N(z; s_i, \frac{\gamma_0 I_z}{2})$$

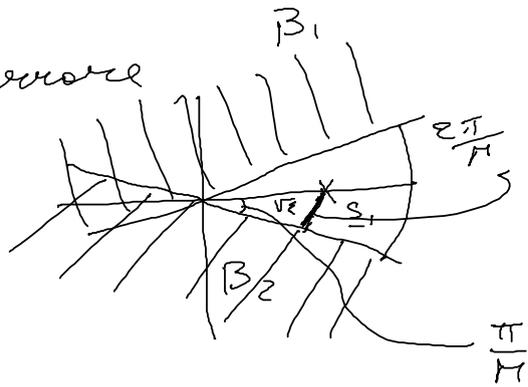
$$P(e) = 1 - P(c) = 1 - \sum_{i=1}^M p(c|a_i) \pi_i = 1 - \sum_{i=1}^M \int_{\mathcal{R}_i} p(z|a_i) dz \pi_i$$

$$= 1 - \int_{\mathcal{R}_1} p(z|a_1) dz \underbrace{\sum_{i=1}^M \pi_i}_1 = 1 - \int_{\mathcal{R}_1} N(z; \sqrt{E}, \frac{\gamma_0 I_z}{2}) dz$$



Appross. della prob. di errore

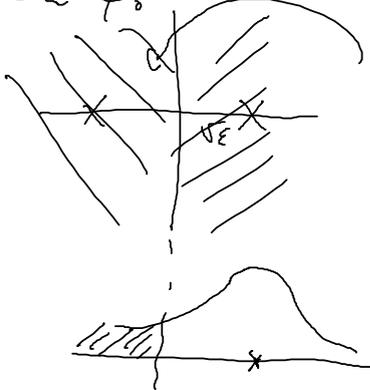
$$P(e) \approx P_1 \{z \in B_1 | a_1\} + P_2 \{z \in B_2 | a_1\}$$



$$Q\left(\frac{\sqrt{E} \sin \frac{\pi}{H}}{\gamma_0/2}\right) = Q\left(\sqrt{\frac{2E}{\gamma_0}} \sin \frac{\pi}{H}\right)$$

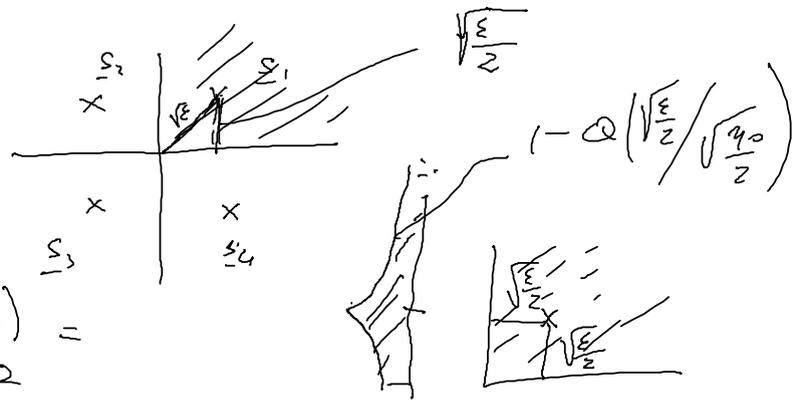
$$Q(\cdot) < P(e) < 2Q(\cdot)$$

$\pi = 2 \quad \phi_0 = 0$



$$P(e) = Q\left(\frac{\sqrt{E}}{\sqrt{\gamma_0/2}}\right) = Q\left(\sqrt{\frac{2E}{\gamma_0}}\right)$$

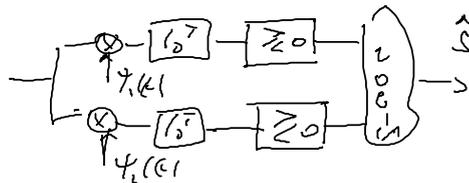
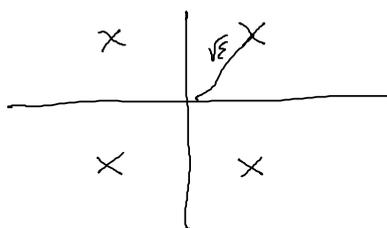
$\pi = 4 \quad \phi_0 = \pi/4$
QPSK

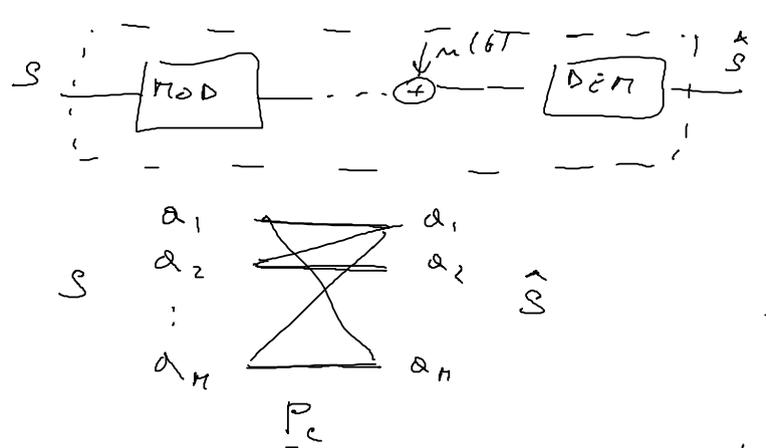
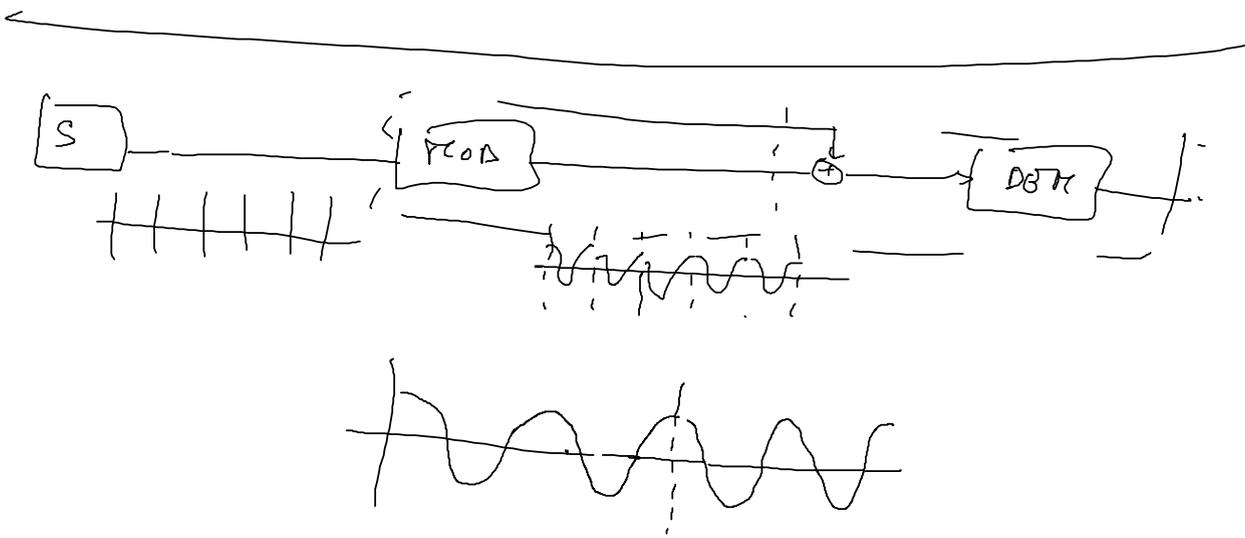


$$P(e) = 1 - P(c) = 1 - P(c|a_1) = 1 - (1 - Q(\sqrt{\frac{E}{\gamma_0}}))^2$$

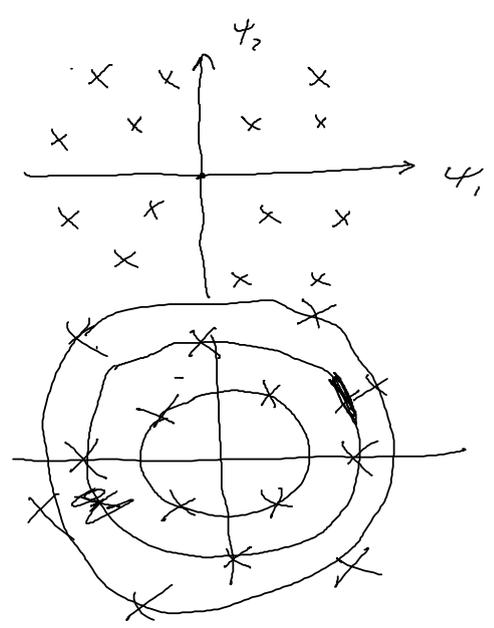
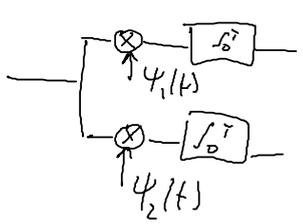
$$= 1 - (1 + Q^2(\cdot) - 2Q)$$

$$= 2Q(1 - Q^2(\cdot)) = 2Q(\cdot)(1 - Q(\cdot)) \approx 2Q(\cdot)$$

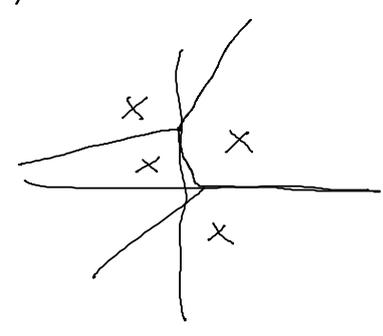
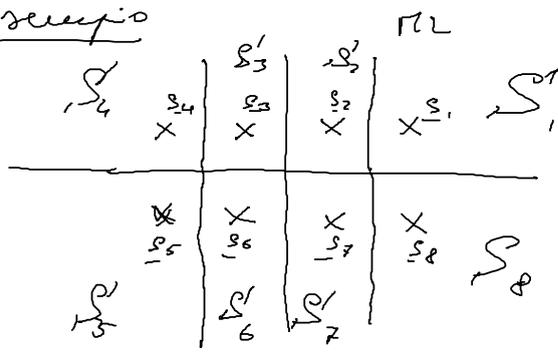




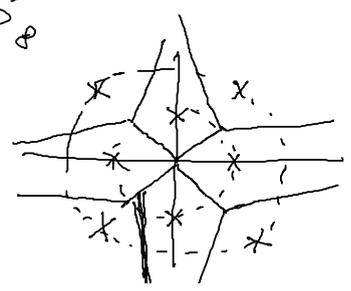
QAM
 QUADRATURE
 AMPLITUDE
 MODULATION



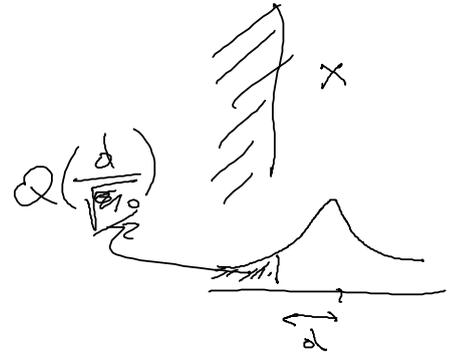
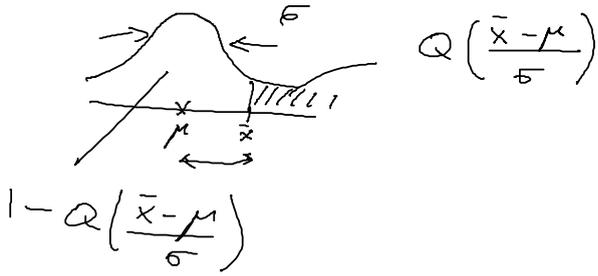
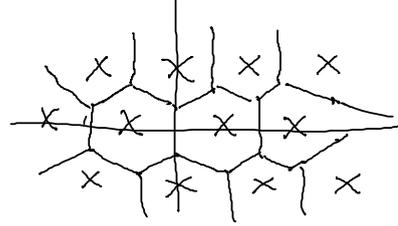
Esempio



X X X X
 X X X X



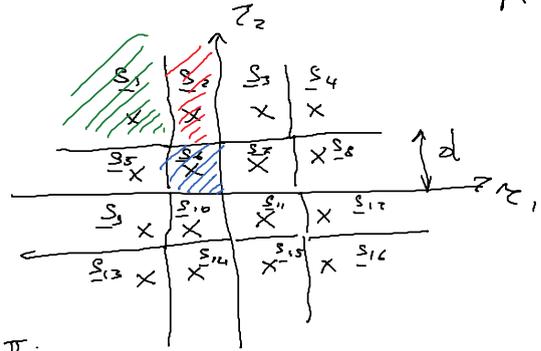
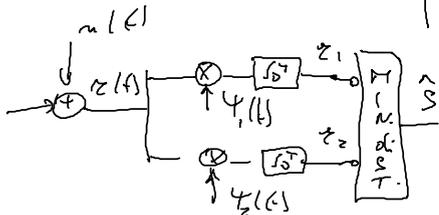
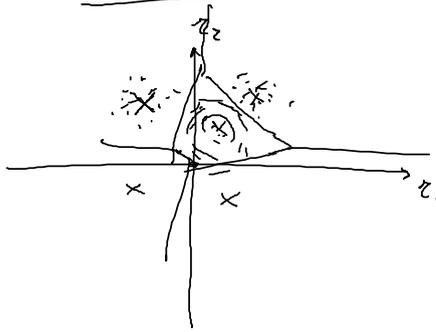
X X X X
 X X X X
 X X X X
 X X X X



MOD. QAM

$$s_j(t) = s_{j,1} \psi_1(t) + s_{j,2} \psi_2(t)$$

$$s_j = \begin{bmatrix} s_{j,1} \\ s_{j,2} \end{bmatrix}$$

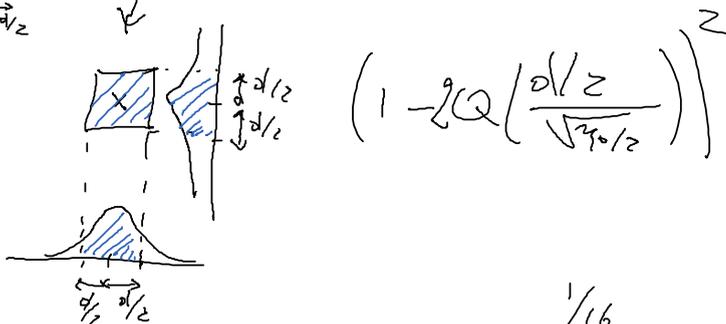
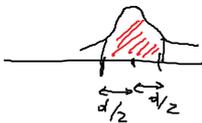
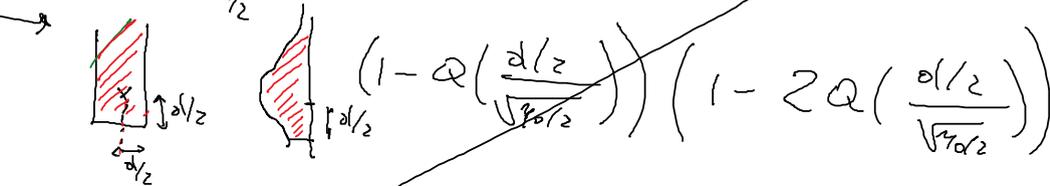
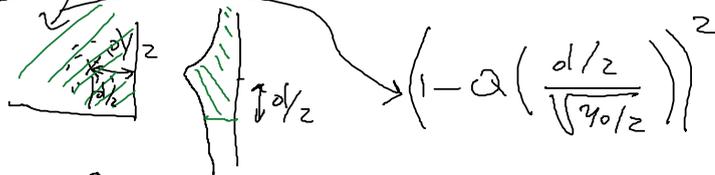


$$P(e) = 1 - P(c) = 1 - \sum_{i=1}^M P(c|\alpha_i) \pi_i$$

$$= 1 - \sum_{i=1}^M \int_{\mathcal{R}} f(z|\alpha_i) dz \pi_i = 1 - \sum_{i=1}^M \int_{\mathcal{S}_i} \mathcal{N}(z; s_i, \eta_0/2 I_2) dz \pi_i$$

$$= 1 - \int_{\mathcal{S}_1} \mathcal{N}(z; s_1, \eta_0/2 I_2) dz (\pi_1 + \pi_4 + \pi_{13} + \pi_{16})$$

$$- \int_{\mathcal{S}_2} \mathcal{N}(z; s_2, \eta_0/2 I_2) dz (\pi_2 + \pi_3 + \pi_5 + \pi_8 + \pi_9 + \pi_{12} + \pi_{14} + \pi_{15}) - \int_{\mathcal{S}_6} \mathcal{N}(z; s_6, \eta_0/2 I_2) dz (\pi_6 + \pi_7 + \pi_{10} + \pi_{11})$$

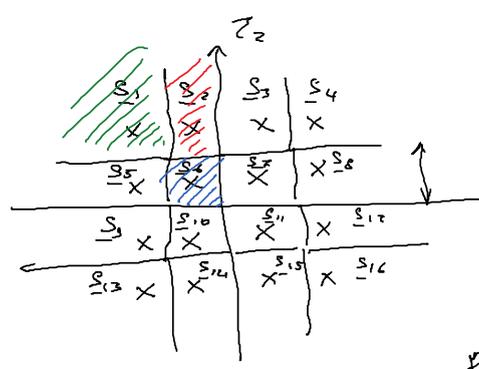


$$\frac{d\sqrt{2}}{2\sqrt{\eta_0}} = \frac{d}{\sqrt{2\eta_0}}$$

1/16

$\frac{d}{2} \quad \frac{d}{2}$

$\frac{1}{16} \quad \frac{2\sqrt{\gamma_0}}{\sqrt{2\gamma_0}} - \sqrt{2\gamma_0}$



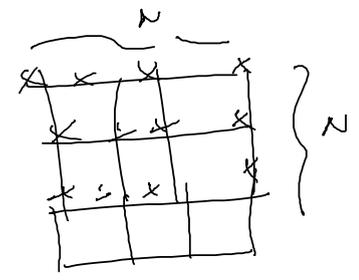
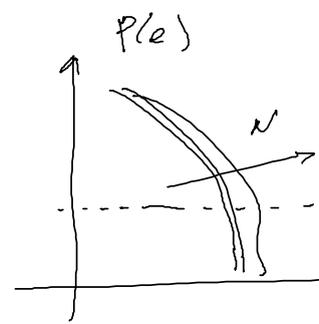
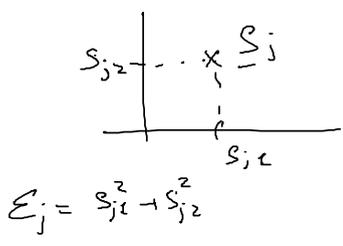
$$P(e) = 1 - (\pi_1 + \pi_4 + \pi_{13} + \pi_{16}) (1 - Q(\cdot))^2$$

$$- (\pi_2 + \pi_3 + \pi_5 + \pi_8 + \pi_9 + \pi_{12} + \pi_{14} + \pi_5) (1 - Q)(1 - 2Q)$$

$$- (\pi_6 + \pi_7 + \pi_{10} + \pi_{11}) (1 - 2Q)^2$$

$\pi_i = 1/16$

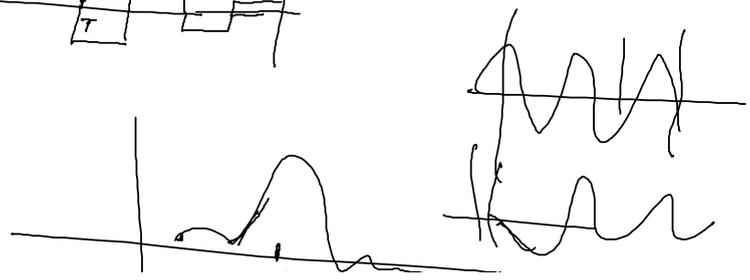
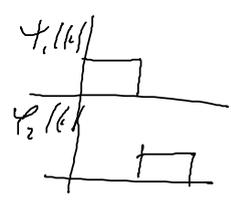
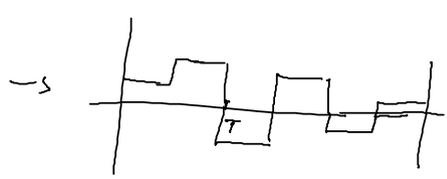
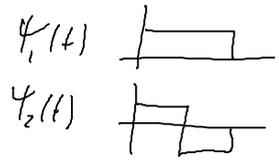
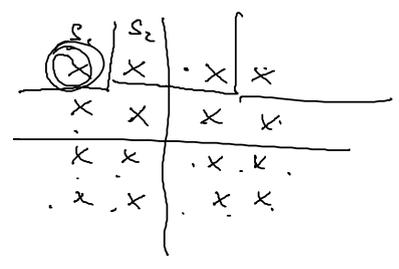
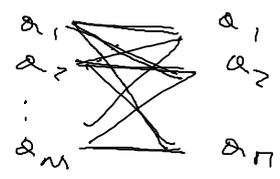
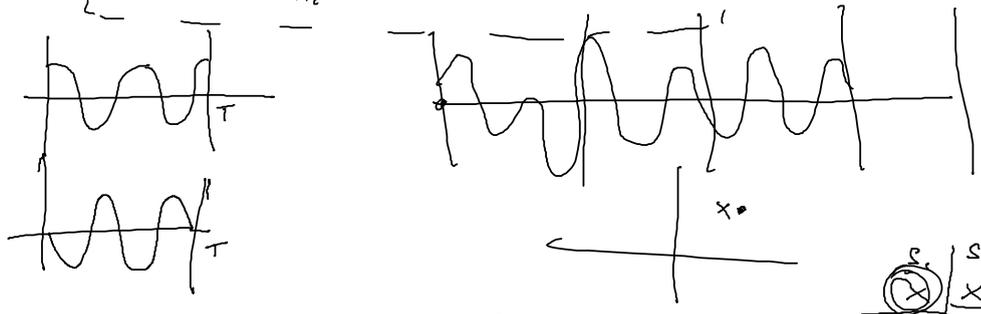
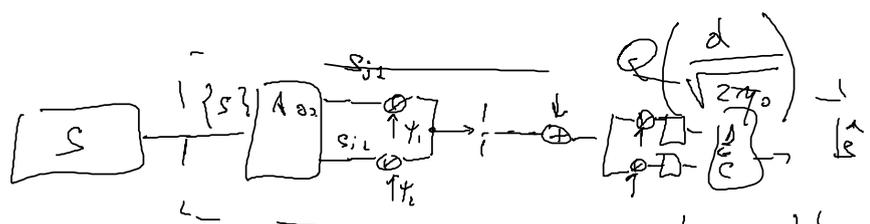
$$= 1 - \frac{1}{4} (1 - Q(\cdot))^2 - \frac{1}{2} (1 - Q)(1 - 2Q) - \frac{1}{4} (1 - 2Q)^2$$

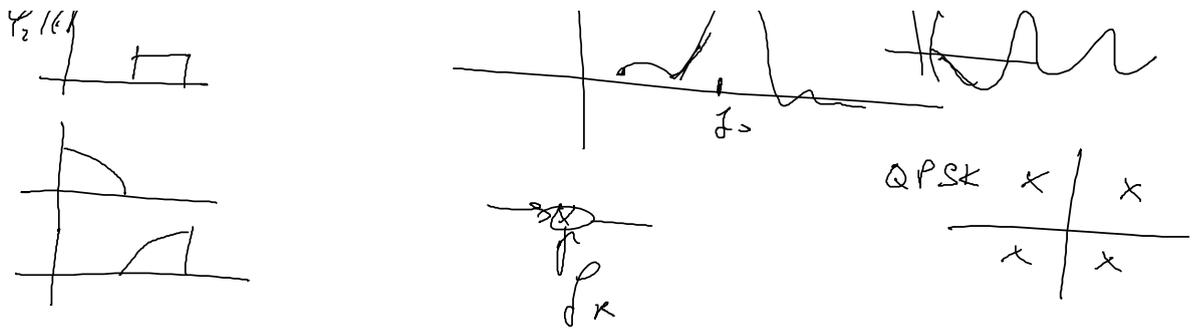


$$\bar{\Sigma} = \frac{d^2 (N^2 - 1)}{6}$$

$$\lg N^2$$

$$\frac{\bar{E}_b}{\gamma_0} = \frac{\bar{\Sigma}}{\lg N^2 \gamma_0} = \frac{d^2 (N^2 - 1)}{12 \lg N^2 \gamma_0}$$

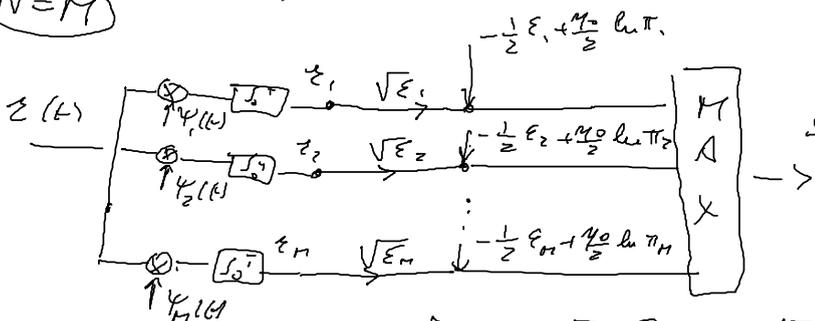




$$S \in \mathcal{R} = \{a_1, \dots, a_M\} \rightarrow \{s_1(t), s_2(t), \dots, s_M(t)\}$$

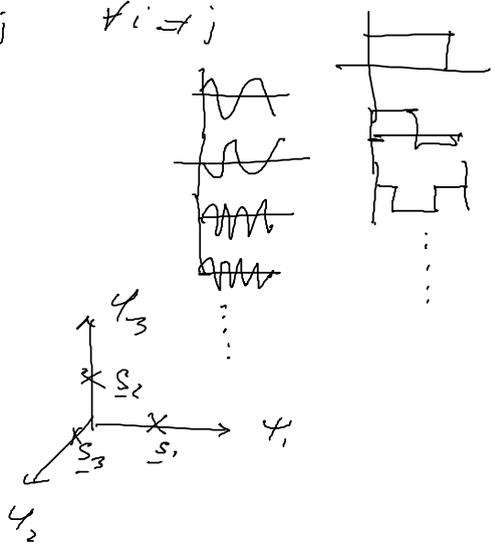
$$\psi_i(t) = \frac{s_i(t)}{\sqrt{E_i}} \quad (i=1, \dots, M) \quad s_i \perp s_j \quad \forall i \neq j$$

$N=M$

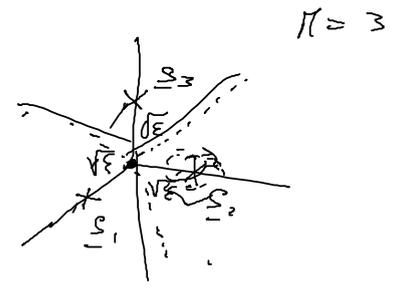
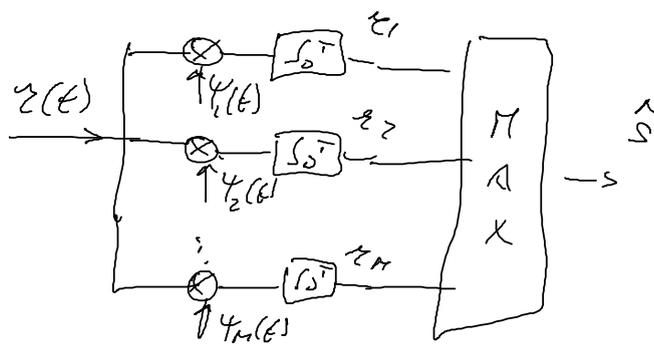


$$\int_0^T s_i(t) \psi_i(t) dt = \int_0^T \frac{s_i^2(t)}{\sqrt{E_i}} dt = \sqrt{E_i}$$

$$s_1 = \begin{bmatrix} \sqrt{E_1} \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} \quad s_2 = \begin{bmatrix} 0 \\ \sqrt{E_2} \\ 0 \\ 0 \\ 0 \end{bmatrix} \quad \dots \quad s_M = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ \sqrt{E_M} \end{bmatrix}$$



Caso particolare: $E_i = E$ (equienergetici); Ric MLC

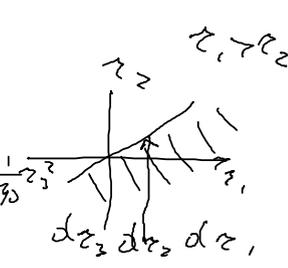


$$P(e) = 1 - P(c) = 1 - \sum_{i=1}^M p(c|a_i) \pi_i = 1 - \sum_{i=1}^M \int_{\mathcal{R}} \mathcal{N}(z; s_i, \frac{E_0}{2} I_3) dz$$

$$= 1 - \int_{\mathcal{S}'_1} \mathcal{N}(z; s_1, \frac{E_0}{2} I_3) dz = 1 - P(c|a_1)$$

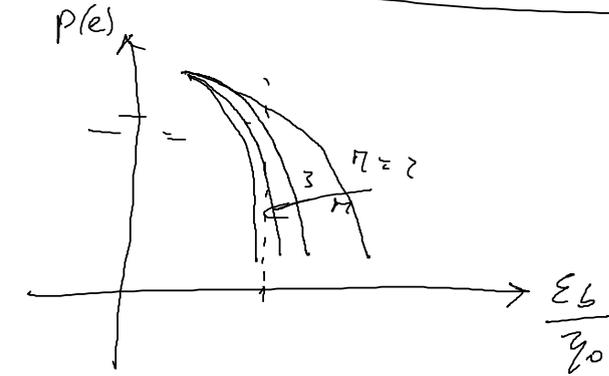
$$= 1 - P\{z_1 > z_2, z_1 > z_3 | a_1\} = 1 - \int_{\mathcal{S}'_1} \mathcal{N}(z; \begin{bmatrix} \sqrt{E} \\ 0 \\ 0 \end{bmatrix}, \frac{E_0}{2} I_3) dz$$

$$= 1 - \int_{-\infty}^{+\infty} \int_{-\infty}^{z_1} \int_{-\infty}^{z_1} \mathcal{N}(z; \begin{bmatrix} \sqrt{E} \\ 0 \\ 0 \end{bmatrix}, \frac{E_0}{2} I_3) dz$$

$$\begin{aligned}
 &= 1 - \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} N\left(\underline{z}; \begin{bmatrix} \sqrt{\epsilon} \\ 0 \\ 0 \end{bmatrix}, \frac{\gamma_0}{2} \underline{I}_3\right) d\underline{z} \\
 &= 1 - \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}(z_1 - \sqrt{\epsilon})^2} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}z_2^2} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}z_3^2} dz_3 dz_2 dz_1 \\
 &= 1 - \int_{-\infty}^{\infty} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}(z_1 - \sqrt{\epsilon})^2} \int_{-\infty}^{\infty} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}z_2^2} dz_2 \int_{-\infty}^{\infty} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}z_3^2} dz_3 dz_1 \\
 &= 1 - \int_{-\infty}^{\infty} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}(z_1 - \sqrt{\epsilon})^2} \left(1 - Q\left(\frac{z_1}{\sqrt{\gamma_0/2}}\right)\right)^2 dz_1
 \end{aligned}$$


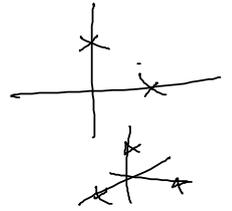
Generalizzazione

$$P(e) = 1 - \int_{-\infty}^{\infty} \frac{1}{\sqrt{\pi\gamma_0}} e^{-\frac{1}{\gamma_0}(z_1 - \sqrt{\epsilon})^2} \left(1 - Q\left(\frac{z_1}{\sqrt{\gamma_0/2}}\right)\right)^{M-1} dz_1$$

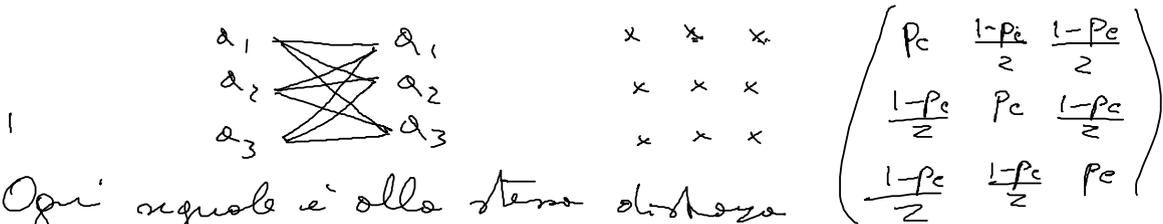


$S_1(t), S_2(t), S_3(t), \dots, S_M(t)$

$$E_b = \frac{E}{\log_2 M}$$



PTT
cresce la richiesta di banda.



Ogni segnale è alla stessa distanza dagli altri

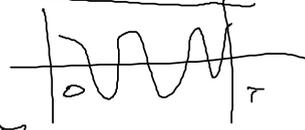
$$\begin{pmatrix} P_c & \frac{1-P_c}{M-1} & \dots & \frac{1-P_c}{M-1} \\ \frac{1-P_c}{M-1} & P_c & \dots & \frac{1-P_c}{M-1} \\ \dots & \dots & \dots & \dots \\ \frac{1-P_c}{M-1} & \dots & \frac{1-P_c}{M-1} & P_c \end{pmatrix}$$

CANALE
UNIFORME

$$\left| \frac{1-p_c}{n-1} \dots \frac{1-p_c}{n-1} \right| p_c$$

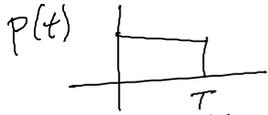
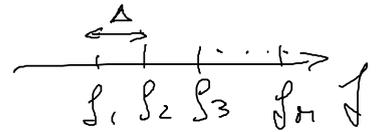
MODULAZIONE DI FREQUENZA ORTOGONALE

$$s_j(t) = A_j \cos 2\pi f_j t$$



$$f_j = f_1 + (j-1)\Delta \quad j=1, \dots, M$$

$s_j(t)$ ortogonali e equieenergetici



$$s_j(t) = A \cos 2\pi f_j t \quad t \in [0, T]$$

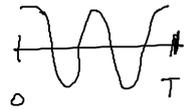
$$E_j = \int_0^T s_j^2(t) dt = \int_0^T A^2 \cos^2 2\pi f_j t dt = A^2 \int_0^T \left(\frac{1}{2} + \frac{1}{2} \cos 4\pi f_j t \right) dt$$

$$= \frac{A^2}{2} T + \frac{A^2}{2} \left[\frac{\sin 4\pi f_j t}{4\pi f_j} \right]_0^T = \frac{A^2}{2} T + \frac{A^2}{2} \frac{\sin 4\pi f_j T}{4\pi f_j}$$

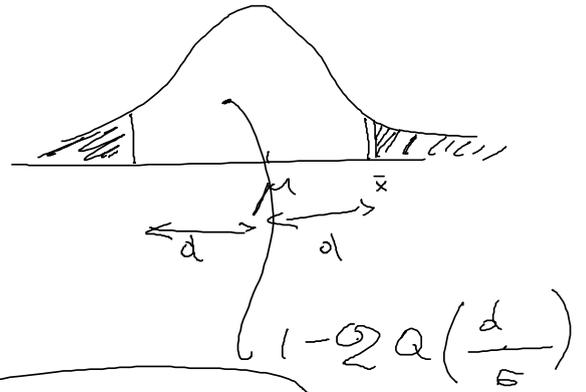
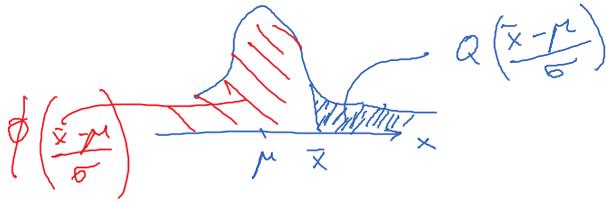
EQUIENERGETICITÀ: $4\pi f_j T = k_j \pi$

ORTOGONALITÀ: $\int_0^T s_j(t) s_i(t) dt \stackrel{?}{=} 0$

$$f_j = \frac{k_j}{4T}$$



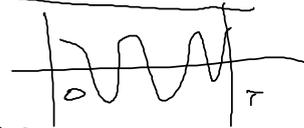
Recap: integrazione delle Gauss



MOD. ORTOGONALE: FSK

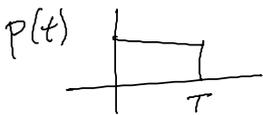
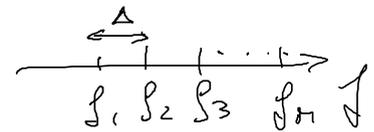
MODULAZIONE DI FREQUENZA ORTOGONALE

$$s_j(t) = A_p(t) \cos 2\pi f_j t$$



$$f_j = f_1 + (j-1)\Delta \quad j=1, \dots, M$$

$s_j(t)$ ortogonali e equiequiprobabili

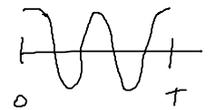


$$s_j(t) = A \cos 2\pi f_j t \quad t \in [0, T]$$

$$\begin{aligned} E_j &= \int_0^T s_j^2(t) dt = \int_0^T A^2 \cos^2 2\pi f_j t dt = A^2 \int_0^T \left(\frac{1}{2} + \frac{1}{2} \cos 4\pi f_j t \right) dt \\ &= \frac{A^2}{2} T + \frac{A^2}{2} \left[\frac{\sin 4\pi f_j t}{4\pi f_j} \right]_0^T = \frac{A^2}{2} T + \frac{A^2}{2} \frac{\sin 4\pi f_j T}{4\pi f_j} \end{aligned}$$

EQUIENERGICITA': $4\pi f_j T = k_j \pi$

$$f_j = \frac{k_j}{4T}$$



ORTOGONALITA': $\int_0^T s_i(t) s_j(t) dt \stackrel{?}{=} 0$

$$\int_0^T s_i(t) s_j(t) dt = \int_0^T A \cos 2\pi f_i t \cdot A \cos 2\pi f_j t dt \quad (i \neq j)$$

$$= \frac{A^2}{2} \int_0^T \cos 2\pi (f_i + f_j) t dt + \frac{A^2}{2} \int_0^T \cos 2\pi (f_i - f_j) t dt$$

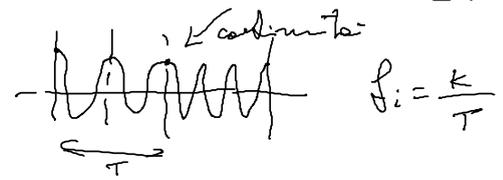
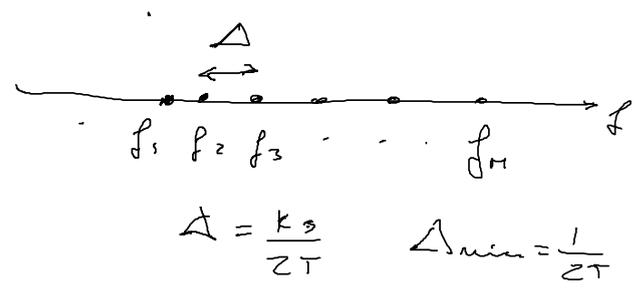
$$= \frac{A^2}{2} \left[\frac{\sin 2\pi (f_i + f_j) t}{f_i + f_j} \right]_0^T + \frac{A^2}{2} \left[\frac{\sin 2\pi (f_i - f_j) t}{f_i - f_j} \right]_0^T$$

$$= \frac{A^2}{2} \left[\frac{\sin 2\pi(f_i + f_j)T}{f_i + f_j} + \frac{A^2}{2} \frac{\sin 2\pi(f_i - f_j)T}{f_i - f_j} \right] = 0$$

$$\approx 2\pi(f_i + f_j)T = k_2 \pi \quad 2\pi(f_i - f_j)T = k_3 \pi$$

$$f_i + f_j = \frac{k_2}{2T} \quad (f_i - f_j) = \frac{k_3}{2T}$$

Equipartitione: $\frac{K_1}{4T}$
 ortogonalita: $f_i + f_j = \frac{k_2}{2T}$
 $f_i - f_j = \frac{k_3}{2T}$



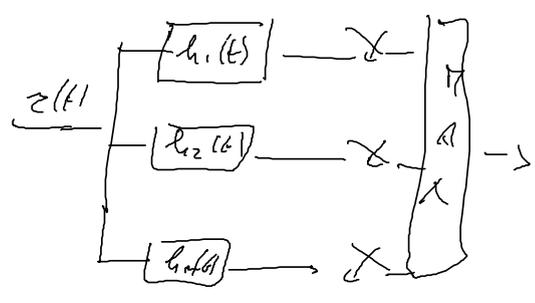
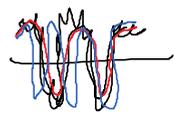
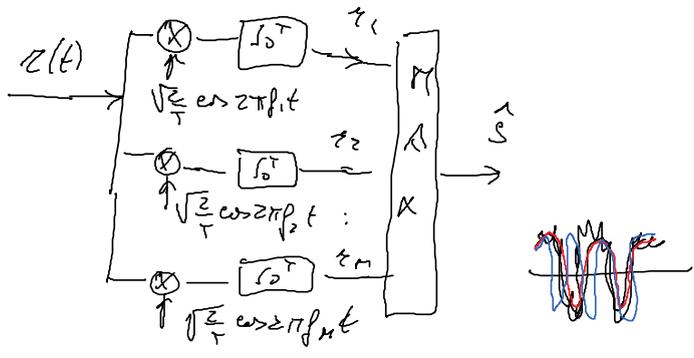
...
RICIPI $f_i = \frac{k}{T}$ \Rightarrow

$$\psi_i(t) = A \cos 2\pi f_i t$$

$$E_i = \frac{A^2 T}{2}$$

$$\psi_i(t) = \frac{A\sqrt{E}}{A\sqrt{T}} \cos 2\pi f_i t$$

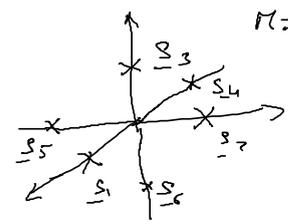
$$\psi_i(t) = \sqrt{\frac{E}{T}} \cos 2\pi f_i t$$



$$h_i(t) = \sqrt{\frac{E}{T}} \psi_i(T-t)$$

$$H_i(f) =$$

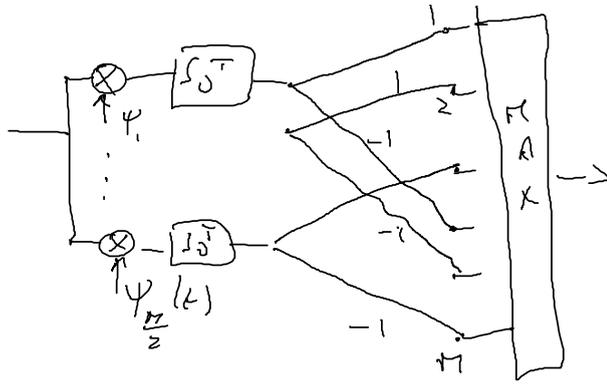
M porzi
 $\{a_1, a_2, \dots, a_M\}$
 \downarrow
 $\{s_1(t), \dots, s_{M/2}(t), -s_1(t), \dots, -s_{M/2}(t)\}$



$$\psi_i(t) = \frac{s_i(t)}{\sqrt{E_i}} \quad i = 1, \dots, \frac{M}{2}$$

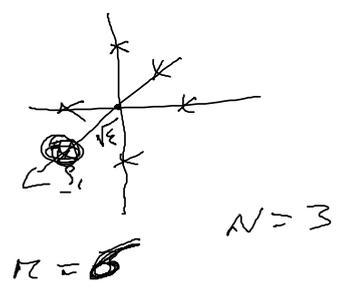


$s_i(t) \perp s_j(t)$
 $i \neq j$
 $i, j = 1, \dots, \frac{M}{2}$



$\sqrt{\epsilon_i}$

Decidi per $a_j \approx |R_j| > |R_i|$ e $R_j > 0$ $j=1, \dots, \frac{M}{2}$
 " " $a_{j+\frac{M}{2}} \approx |R_j| > |R_i|$ e $R_j < 0$ "



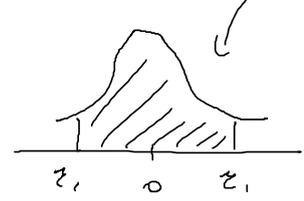
$P(e) = 1 - P(c) = 1 - \sum_{i=1}^M P(c|a_i) \pi_i = 1 - P(c|a_1)$

$P(c|a_1) = P\left\{ R_2 > 0, |R_1| > |R_2|, |R_1| > |R_3| \mid a_1 \right\}$

$= \int_0^\infty \left[\int_{-z_1}^{z_1} \int_{-z_1}^{z_1} \mathcal{N}\left(z; \begin{bmatrix} \sqrt{\epsilon} \\ 0 \\ 0 \end{bmatrix}, \frac{\gamma_0}{2} \mathbf{I}_3\right) dz_2 dz_3 \right] dz_1$

$= \int_0^\infty \int_{-z_1}^{z_1} \int_{-z_1}^{z_1} \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{1}{\gamma_0}(z_1 - \sqrt{\epsilon})^2} \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{1}{\gamma_0} z_2^2} \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{1}{\gamma_0} z_3^2} dz_2 dz_3 dz_1$

$= \int_0^\infty \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{(z_1 - \sqrt{\epsilon})^2}{\gamma_0}} \left[\int_{-z_1}^{z_1} \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{1}{\gamma_0} z_2^2} dz_2 \right] \left[\int_{-z_1}^{z_1} \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{1}{\gamma_0} z_3^2} dz_3 \right] dz_1$



$1 - 2Q\left(\frac{z_1}{\sqrt{\gamma_0/2}}\right)$

$= \int_0^\infty \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{(z_1 - \sqrt{\epsilon})^2}{\gamma_0}} \left(1 - 2Q\left(\frac{z_1}{\sqrt{\gamma_0/2}}\right)\right)^2 dz_1$

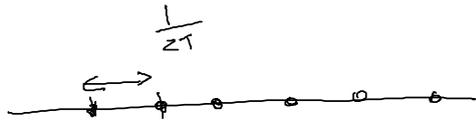
$P(e) = 1 - \int_0^\infty \frac{1}{\sqrt{\pi \gamma_0}} e^{-\frac{(z_1 - \sqrt{\epsilon})^2}{\gamma_0}} \left(1 - Q\left(\frac{z_1}{\sqrt{\gamma_0/2}}\right)\right)^{M-1} dz_1$

FSK ortogonale

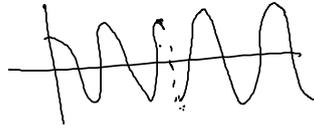
FSB - bi-ortogonale

$$s_i(t) = A \cos 2\pi f_i t \quad i = 1, \dots, \frac{M}{2}$$

$$s_i(t) = -A \cos 2\pi f_i t \quad i = \frac{M}{2} + 1, \dots, M$$

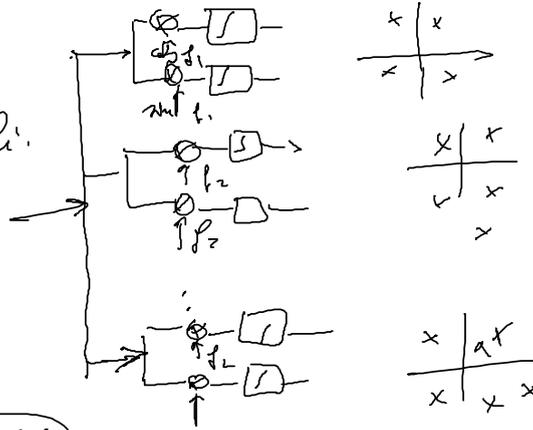
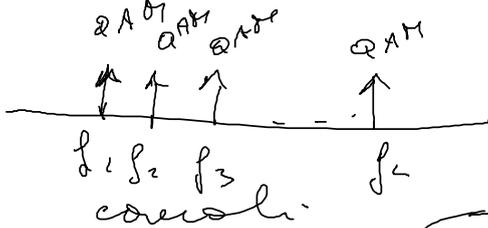


$M \uparrow$ occupazine \uparrow
in banda



QAM

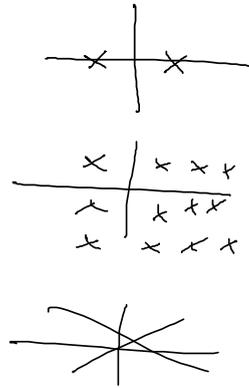
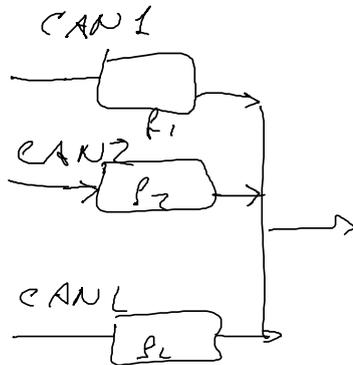
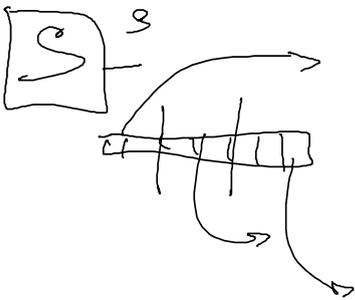
$f_i \rightarrow$ x pendi
ortopendi.



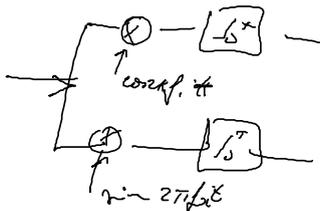
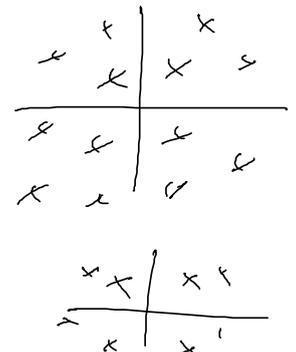
\Rightarrow fluss
dati

OFTM

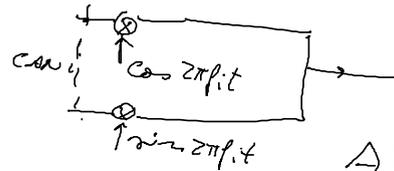
ORTHOGONAL FREQUENCY-DIVISION MULTIPLEX



ADSL

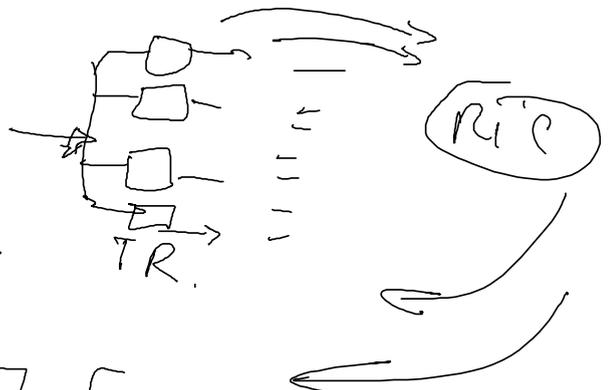
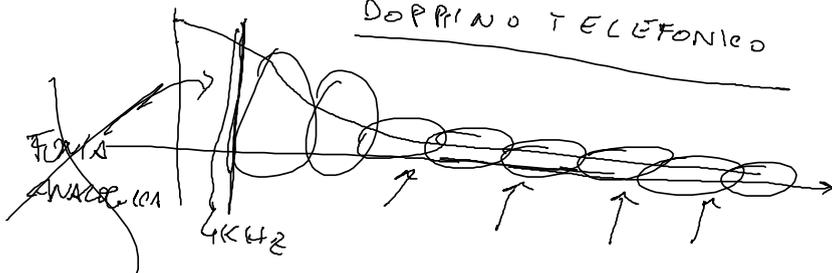


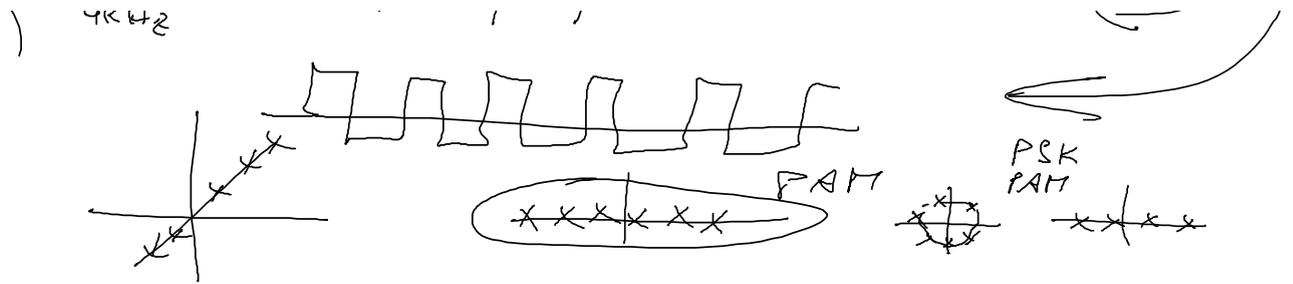
bit \rightarrow



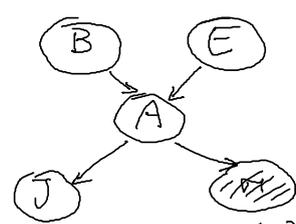
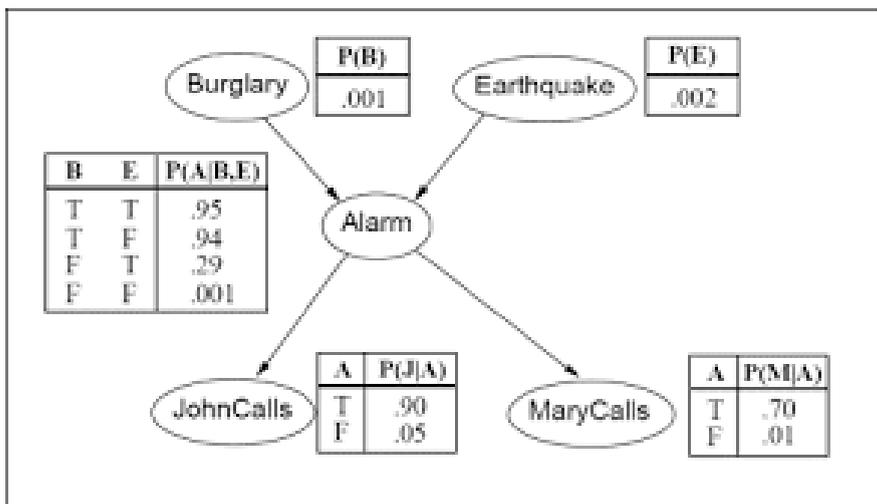
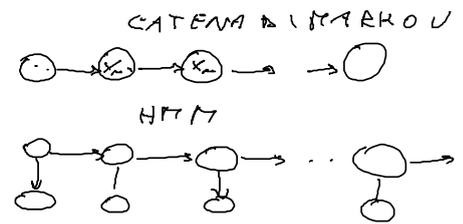
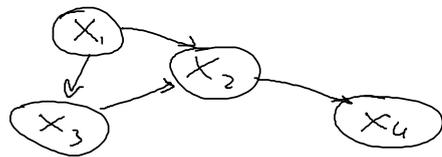
ADSL

DOPPIO TELEFONICO





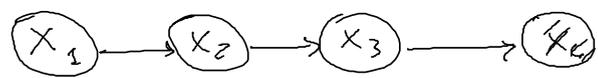
MODELLI BAYESIANI BASATI SU GRAFI



$$p(A, B, E, J, M) = p(A|BE) p(B) p(E) p(M|A) p(J|A)$$

$$p(E) = \sum_{A, B, J} p(A, B, E, J, \bar{M})$$

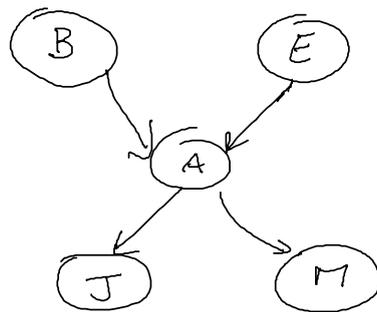
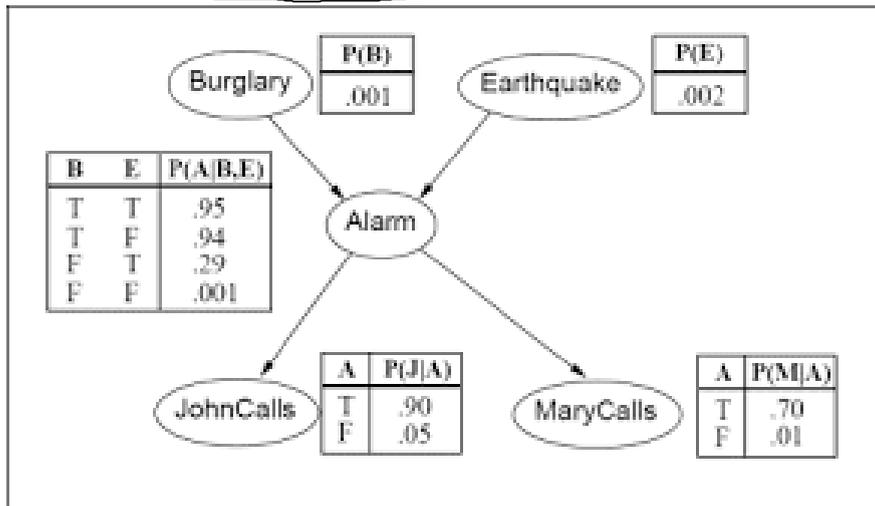
CATENA DI MARKOV



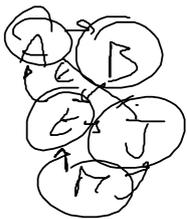
$$p(x_1, x_2, x_3, x_4) = p(x_4|x_3) p(x_3|x_2) p(x_2|x_1) p(x_1)$$

$$p(x_1 | \bar{x}_4) = \sum_{x_2, x_3} p(x_1, x_2, x_3, \bar{x}_4)$$

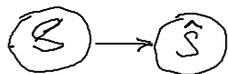
SISTEMI BAYESIANI



$$P(A, B, E, J, M) = P(B)P(E)P(A|BE)P(J|A)P(M|A)$$

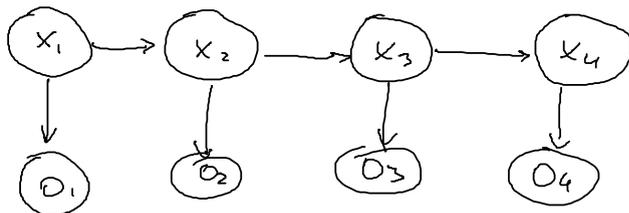


$$P(x_1, x_2, x_3, x_4) = P(x_4|x_3)P(x_3|x_2)P(x_2|x_1)P(x_1)$$



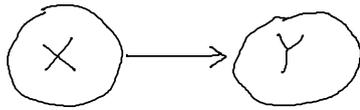
$$P(S, \hat{S}) = P(\hat{S}|S)P(S)$$

Hint



$$P(x_1, x_2, x_3, x_4, o_1, o_2, o_3, o_4) = P(x_1)P(o_1|x_1)P(x_2|x_1)P(o_2|x_2)P(x_3|x_2)P(o_3|x_3)P(x_4|x_3)P(o_4|x_4)$$

$$X \in \mathcal{X} = \{x^1, \dots, x^n\}$$

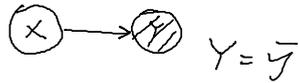
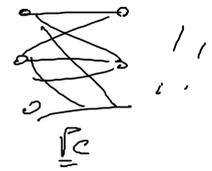
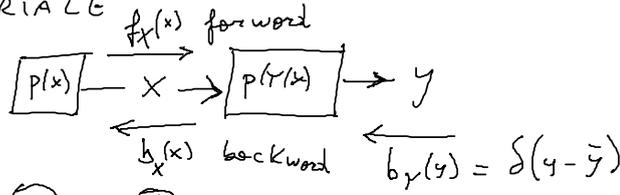


$$X \in \mathcal{X} = \{x^1, \dots, x^n\}$$

$$Y \in \mathcal{Y} = \{y^1, \dots, y^m\}$$

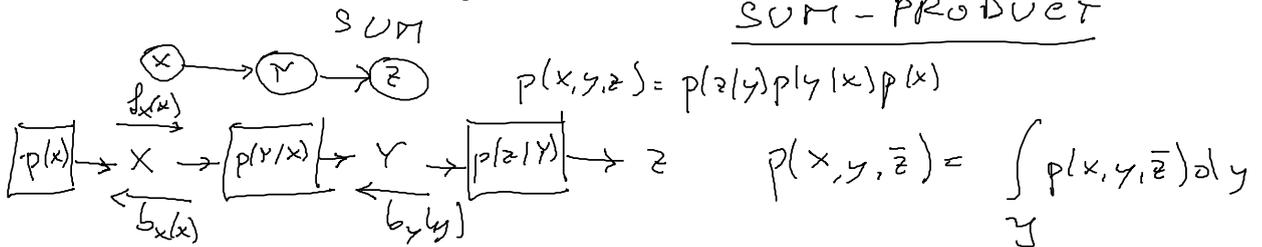
$$p(x, y) = p(y|x) p(x)$$

CRAFO FATTORIALE



$$p(x|Y=\bar{y}) = \frac{p(x, Y=\bar{y})}{p(Y=\bar{y})} = \frac{p(Y=\bar{y}|x) p(x)}{p(Y=\bar{y})} \propto \underbrace{p(Y=\bar{y}|x)}_{b_x(x)} \underbrace{p(x)}_{f_x(x)}$$

$$b_x(x) = \int_{\mathcal{Y}} p(y|x) \delta(y - \bar{y}) dy = p(\bar{y}|x) \quad \downarrow \text{PRODUCT}$$



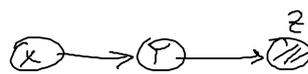
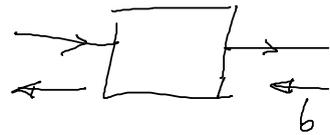
SUM - PRODUCT

$$p(x, y, z) = p(z|y) p(y|x) p(x)$$

$$p(x, y, \bar{z}) = \int_{\mathcal{Y}} p(x, y, \bar{z}) dy$$

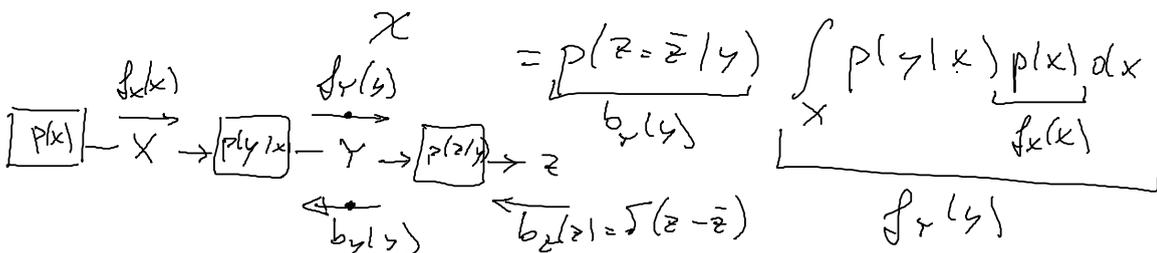
$$= \int_{\mathcal{Y}} p(\bar{z}|y) p(y|x) p(x) dy$$

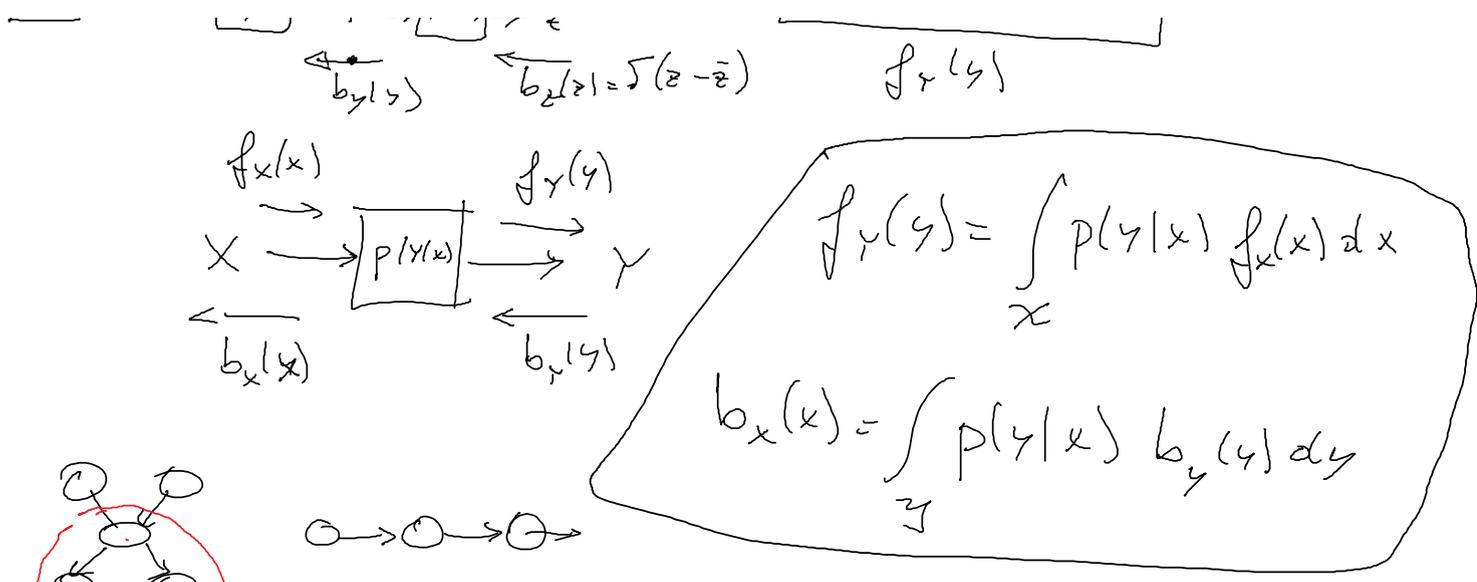
$$= p(x) \underbrace{\int_{\mathcal{Y}} p(\bar{z}|y) p(y|x) dy}_{b_x(x)}$$



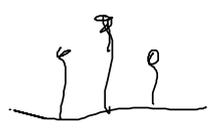
$$p(y|z=\bar{z}) = \frac{p(y, z=\bar{z})}{p(z=\bar{z})} \propto p(y, z=\bar{z}) = \int_{\mathcal{X}} p(x, y, z=\bar{z}) dx$$

$$= \int_{\mathcal{X}} p(z=\bar{z}|y) p(y|x) p(x) dx$$



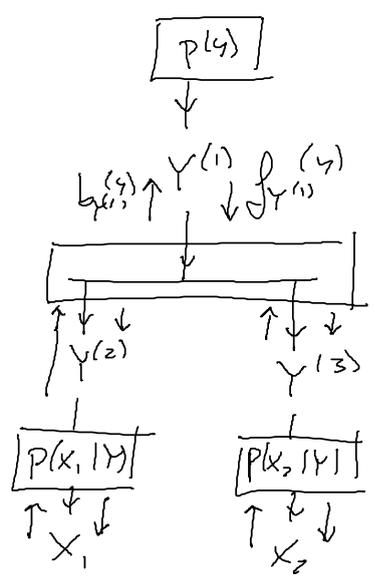


$$p(y|x_1, x_2) = p(x_1|y) p(x_2|y) p(y)$$



$$b_{y(x_2)}(y) = p(x_2|y)$$

$$b_{x_1}(x_1) = \delta(x_1 - \bar{x}_1)$$



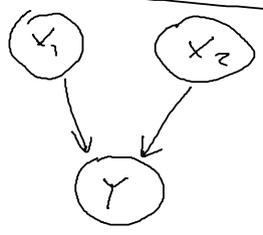
REGOLA DEL PRODOTTO:

messaggi in uscita = prodotto dei messaggi in ingresso

$$b_{y(x_2)}(y) = \int p(x_2|y) U(x_2) dx_2 \propto U(y)$$

$$p(y|x_1 = \bar{x}_1) = \frac{p(y, x_1 = \bar{x}_1)}{p(x_1 = \bar{x}_1)} \propto p(y, x_1 = \bar{x}_1) = \int p(y, x_1 = \bar{x}_1, x_2) dx_2$$

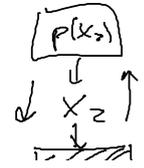
$$= \int p(x_1 = \bar{x}_1 | y) p(x_2 | y) p(y) dx_2 = p(x_1 = \bar{x}_1 | y) p(y) \int p(x_2 | y) dx_2$$



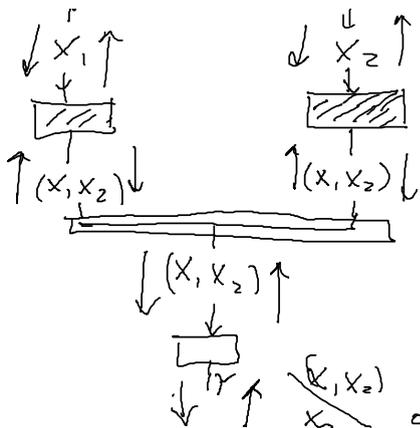
Due genitori

Figlio

$$p(y, x_1, x_2) = p(y|x_1, x_2) p(x_1) p(x_2)$$



$(x_1, x_2) \in$ spazio possibile $\mathcal{X}_x \times \mathcal{X}_z$



$$p(x_1, x_2 | x_1) =$$

$$x_1 \in \mathcal{X}_1 = \{0, 1\}$$

$$x_2 \in \mathcal{X}_2 = \{0, 1\}$$

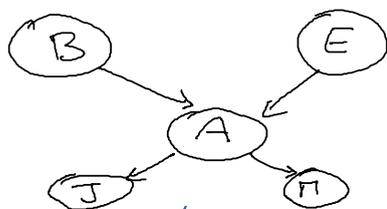
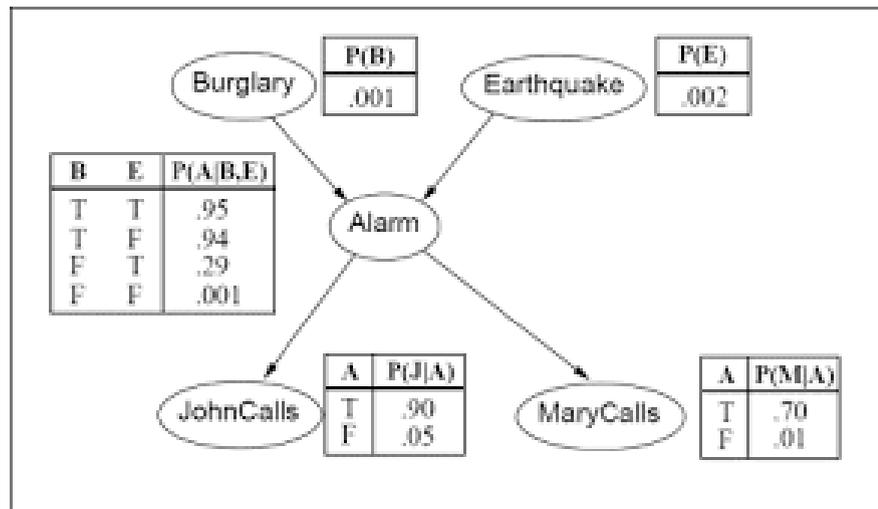
$$\mathcal{X}_1 \times \mathcal{X}_2$$

$$(x_1, x_2) \quad x_1$$

00	0
01	1
10	
11	

	x_2	00	01	10	11
0	x_1	$1/2$	0	$1/2$	0
1		0	$1/2$	0	$1/2$

	x_1	00	01	10	11
0	x_2	$1/2$	$1/2$	0	0
1		0	0	$1/2$	$1/2$



$$p(B, E, A, J, M)$$

$$= p(A|BE) p(B) p(E) p(J|A) p(M|A)$$

$$p(B) = \begin{bmatrix} 0.01 \\ 0.99 \end{bmatrix}$$

$$p(E) = \begin{bmatrix} 0.02 \\ 0.98 \end{bmatrix}$$

BE	TT	TF	FT	FF
B	$1/2$	$1/2$	0	0
E	0	0	$1/2$	$1/2$

BE	TT	TF	FT	FF
E	$1/2$	0	$1/2$	0
B	0	$1/2$	0	$1/2$

$$P(A|BE) =$$

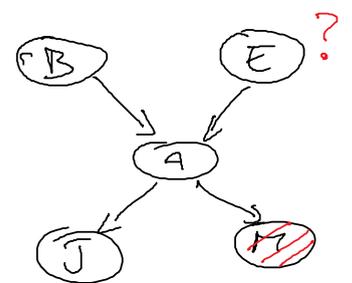
BE	T	F
T	0.95	0.05
F	0.94	0.06
F	0.29	0.61
F	0.01	0.99

$$P(J|A) =$$

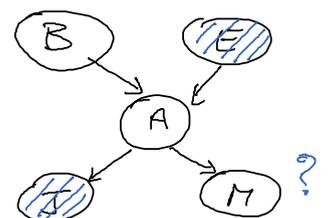
A	T	F
T	0.9	0.1
F	0.01	0.99

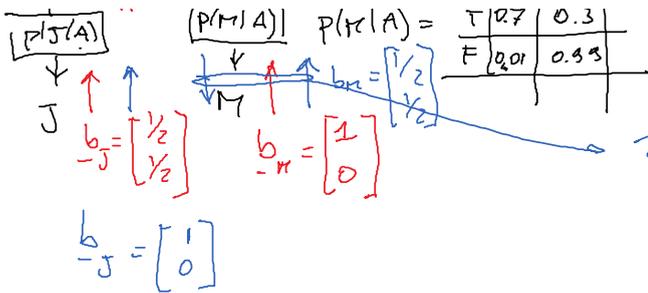
$$P(M|A) =$$

A	T	F
T	0.7	0.3
F	0.01	0.99

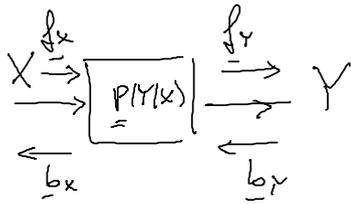


$$p(e|M=Trans)$$



$$P(A) = \begin{array}{c|cc} & T & F \\ \hline T & 0.9 & 0.1 \\ \hline F & 0.01 & 0.99 \end{array}$$


$$P_{Y|X} = \begin{matrix} Y \\ \begin{pmatrix} \dots & \dots & \dots \\ \dots & \dots & \dots \\ \dots & \dots & \dots \end{pmatrix} \end{matrix}$$



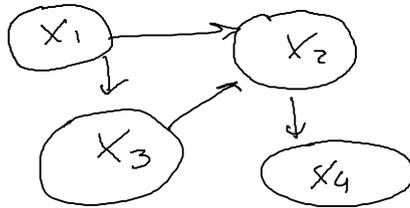
$$f_y = P_{Y|X}^T f_x$$

$$b_x \propto P_{Y|X} b_y$$

$$b_{A^{(3)}} \propto P_{M|A} b_{-m} = \begin{bmatrix} 0.7 & 0.3 \\ 0.01 & 0.99 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 0.7 \\ 0.01 \end{bmatrix}$$

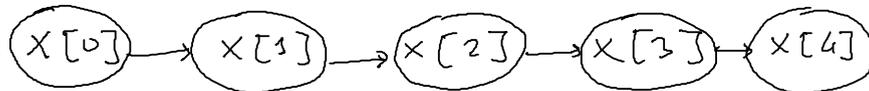
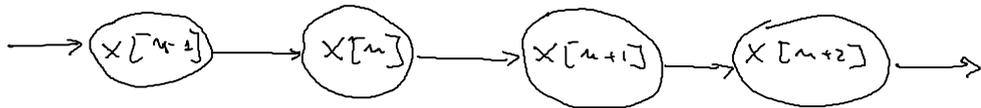
$$b_{A^{(2)}} \propto P_{J|A} b_{-j} = \begin{bmatrix} 0.9 & 0.1 \\ 0.01 & 0.99 \end{bmatrix} \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix} = \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix}$$

RETI BAYESIANE



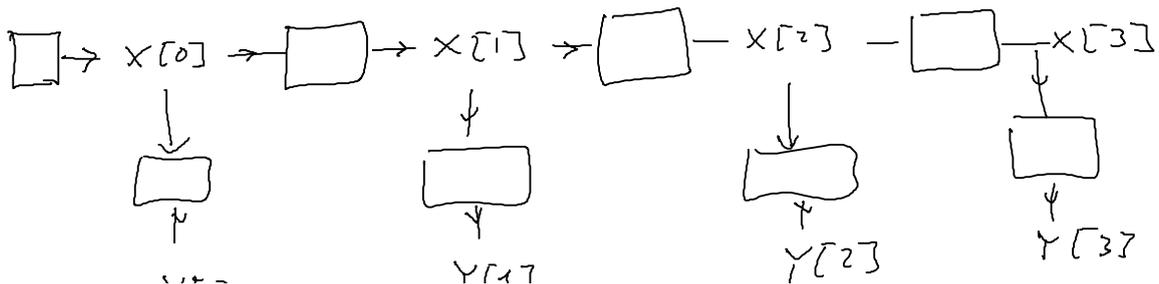
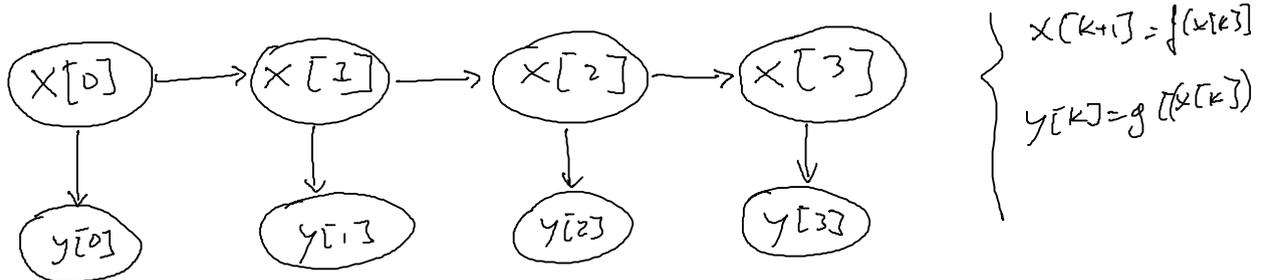
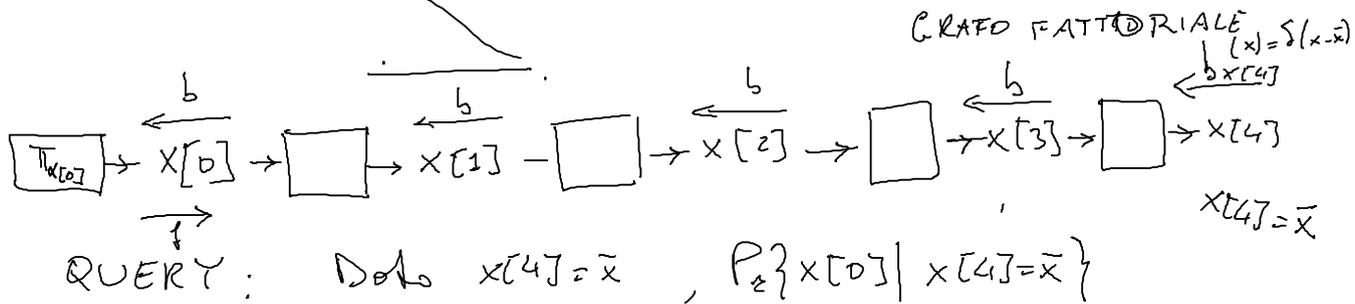
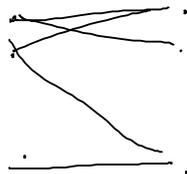
ESEMPIO

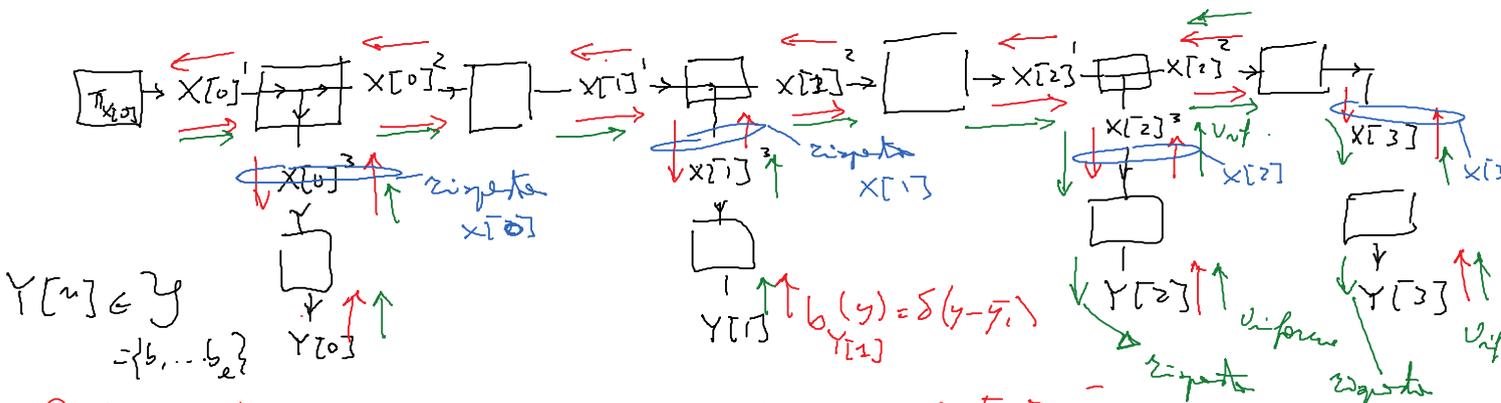
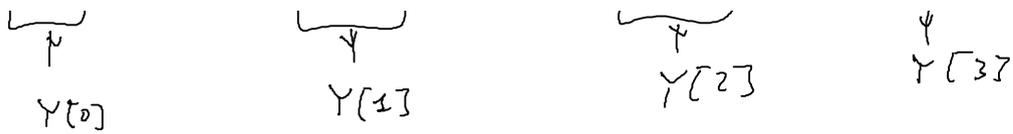
HMM Hidden Markov Model



$X[n] \in \{a_1, a_2, \dots, a_m\}$

$P\{X[n] | X[n-1]\}$





$Y[0] \in \mathcal{Y}$
 $= \{b_1, \dots, b_2\}$

QUERY = $\text{Prob } Y[0] = \bar{y}_0, Y[1] = \bar{y}_1, Y[2] = \bar{y}_2, Y[3] = \bar{y}_3$
 calcolare la sequenza degli stati.
 (prob. a posteriori)

$\Rightarrow P(X[0], X[1], X[2], X[3] | Y[0] = \bar{y}_0, Y[1] = \bar{y}_1, Y[2] = \bar{y}_2, Y[3] = \bar{y}_3)$

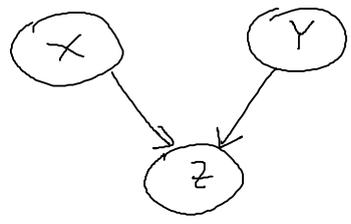
QUERY = $\text{Prob } Y[0] = \bar{y}_0, Y[1] = \bar{y}_1 \Rightarrow P(Y[2] | Y[0] = \bar{y}_0, Y[1] = \bar{y}_1)$

Example

$X \in \{0, 1\}$
 π_x

$Y \in \{0, 1\}$
 π_y

$Z = X + Y$
 $Z \in \{0, 1, 2\}$

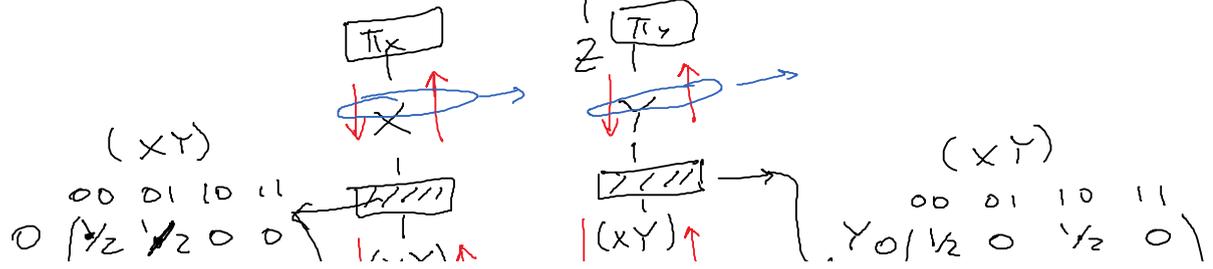
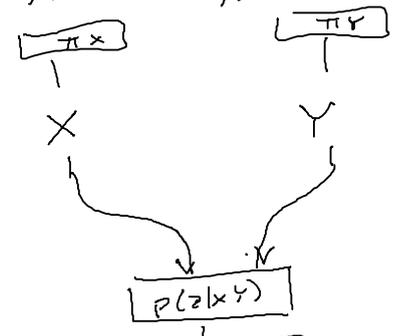


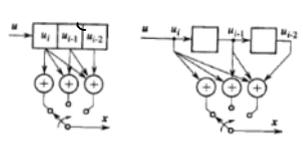
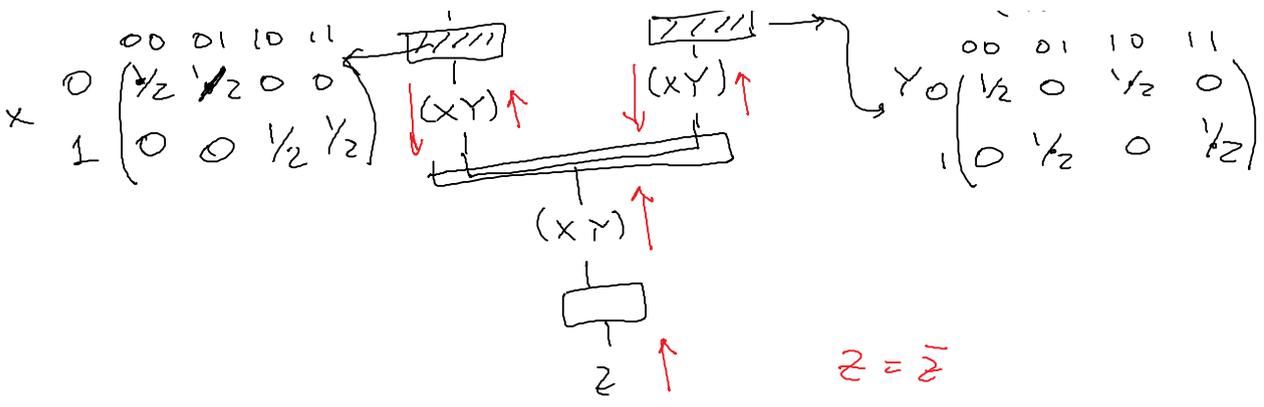
$P(z|xy) =$

$xy \backslash z$	0	1	2
00	1	0	0
01	0	1	0
10	0	1	0
11	0	0	1

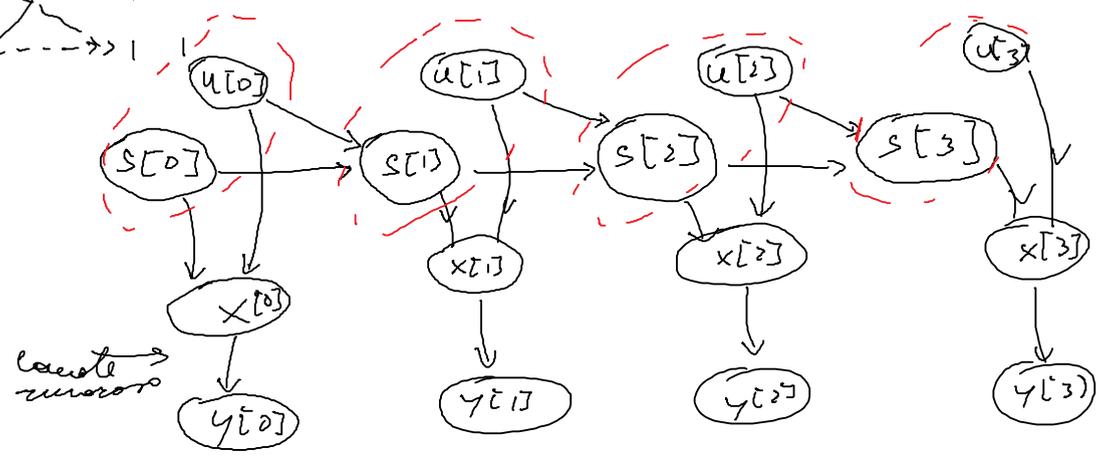
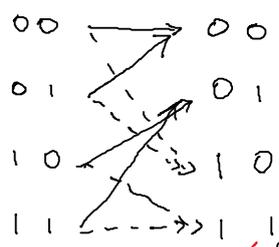
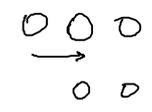
$P(x, y, z) = P(z|x, y) \pi(x) \pi(y)$

$\pi_x = \begin{bmatrix} 0.4 \\ 0.6 \end{bmatrix}$ $\pi_y = \begin{bmatrix} 0.5 \\ 0.5 \end{bmatrix}$

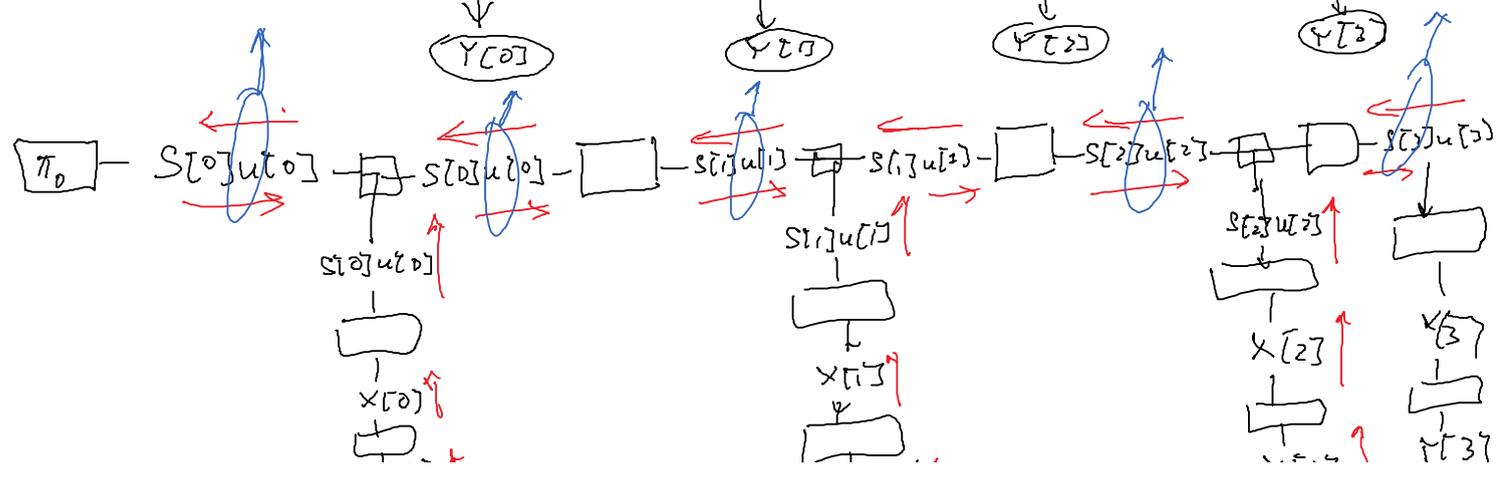
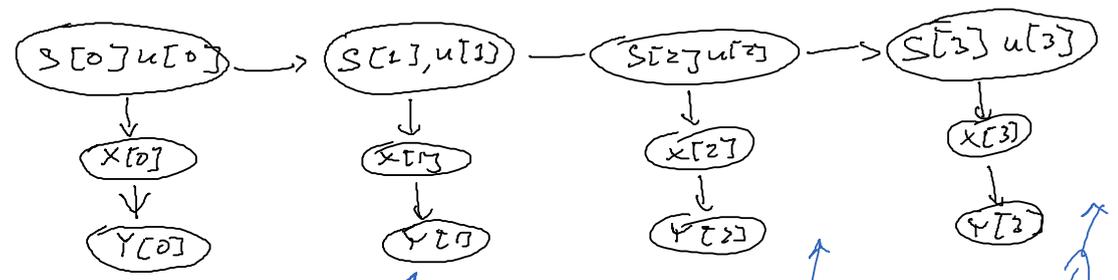


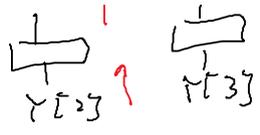
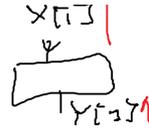
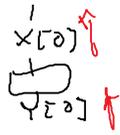


$u \in \{0, 1\}$
 $x \in \{ \text{range di } 3 \text{ bit} \}$



QUERY: $D \rightarrow y[0] = \bar{y}_0, y[1] = \bar{y}_1, y[2] = \bar{y}_2, y[3] = \bar{y}_3$





Cenni su Codifica LDPC
Cenni su Codifica Turbo