

SECONDA UNIVERSITA' DEGLI STUDI DI NAPOLI
 DIPARTIMENTO DI INGEGNERIA INDUSTRIALE E
 DELL'INFORMAZIONE
Scuola Politecnica e delle Scienze di Base

Teoria dei Segnali/Telecomunicazioni 2
 IIa PROVA INTRACORSO
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SOLUZIONI

(1) Si consideri il segnale

$$Y(t) = X(t) \cos^2 2\pi f_0 t + C, \quad (1)$$

dove $X(t)$ è un processo aleatorio SSL avente spettro di potenza $P_X(f) = \Lambda\left(\frac{f}{2B}\right)$ e C una costante aleatoria indipendente da $X(t)$ e distribuita secondo una pdf gaussiana avente media $\mu_C = 1$ e varianza $\sigma_C^2 = 4$. (Si assuma $f_0 >> 2B$). Commentare sulla stazionarietà di $Y(t)$ e valutarne autocorrelazione e spettro di potenza.

(2) Un segnale aleatorio $x(t)$ SSL avente autocorrelazione $R_X(\tau) = \Lambda(\tau)$ è posto all'ingresso di un sistema lineare avente risposta impulsiva $h(t) = \delta(t - 1) + \frac{1}{2}\delta(t - 2)$. Valutare la mutua correlazione uscita-ingresso.

(3) Si consideri una cascata di 3 canali lineari rumorosi identici aventi funzione di trasferimento di energia $|H_i(f)|^2 = \Lambda\left(\frac{f}{B}\right)$, $i = 1, 2, 3$ ($B = 10$ KHz) e rumore additivo all'uscita $n_1(t)$, $n_2(t)$ e $n_3(t)$ rispettivamente, di potenza pari a 9 mV^2 ognuno, e spettro piatto fino alla frequenza B . Si valuti il rapporto segnale(distorto)-rumore in uscita alla catena se all'ingresso è applicato un segnale a banda piatta fino a B con densità spettrale pari a $8 \cdot 10^{-4} \text{ mV}^2/\text{Hz}$.

(4) Un segnale $s(t)$ passa-banda a spettro piatto tra $b = 100$ e $B = 10000$ Hz di valore efficace $P_s^{RMS} = 10 \text{ mV}$ è inviato su un canale distorcere avente risposta armonica di ampiezza $|H_c(f)| = \Lambda\left(\frac{f}{12000}\right)$. Successivamente il segnale è contaminato da rumore additivo avente spettro di potenza piatto da 0 a 10000 Hz e valore efficace di 5 mV . (a) Valutare il rapporto segnale(distorto)/rumore in dB alla fine della catena; (b) Proporre filtri di enfasi e de-enfasi schizzandone l'andamento approssimativo.

(4) Si considerino due sistemi lineari aventi risposte impulsive $h_1(t)$ e $h_2(t)$, e agli ingressi i due segnali aleatori di potenza $X_1(t)$ e $X_2(t)$. I due segnali siano singolarmente e mutuamente stazionari almeno in senso lato. Dimostrare la formula per la mutua correlazione tra le due uscite $R_{y_1 y_2}(\tau)$.

(1)

$$Y(t) = X(t) \cos^2 2\pi f_0 t + C = \frac{X(t)}{2} + \frac{X(t)}{2} \cos 4\pi f_0 t + C$$

$$E[Y(t)] = E[\overset{\circ}{X(t)} \cos^2 2\pi f_0 t + E[C]] = 1$$

$$\begin{aligned} R_Y(t, \tau) &= E[Y(t) Y(t-\tau)] = E\left[\left(\frac{X(t)}{2} + \frac{X(t)}{2} \cos 4\pi f_0 t + C\right)\right. \\ &\quad \left.\left(\frac{X(t-\tau)}{2} + \frac{X(t-\tau)}{2} \cos 4\pi f_0 (t-\tau) + C\right)\right] \\ &= \frac{1}{4} E[X(t) X(t-\tau)] + \frac{1}{4} E[X(t) X(t-\tau)] \cos 4\pi f_0 (t-\tau) + \frac{1}{2} E[X(t)] E[C] \\ &\quad + \frac{1}{4} E[X(t) X(t-\tau)] \cos 4\pi f_0 t + \frac{1}{4} E[X(t) X(t-\tau)] \cos 4\pi f_0 t \cos 4\pi f_0 (t-\tau) \\ &\quad + \frac{1}{2} E[C] E[X(t-\tau)] + \frac{1}{2} E[C] E[X(t-\tau)] \cos 4\pi f_0 (t-\tau) \\ &\quad + E[C^2] \end{aligned}$$

$$\begin{aligned} &= \frac{1}{4} R_X(\tau) + \frac{1}{4} R_X(\tau) \cos 4\pi f_0 (t-\tau) + \frac{1}{4} R_X(\tau) \cos 4\pi f_0 t \\ &\quad + \frac{1}{4} R_X(\tau) \left(\frac{1}{2} \cos 4\pi f_0 (2t-\tau) + \frac{1}{2} \cos 4\pi f_0 \tau \right) + E[C^2] \end{aligned}$$

$$\begin{aligned} &= \frac{1}{4} R_X(\tau) + \frac{1}{4} R_X(\tau) \cos 4\pi f_0 (t-\tau) + \frac{1}{4} R_X(\tau) \cos 4\pi f_0 t \\ &\quad + \frac{1}{8} R_X(\tau) \cos 4\pi f_0 (2t-\tau) + \frac{1}{8} R_X(\tau) \cos 4\pi f_0 \tau + E[C^2] \end{aligned}$$

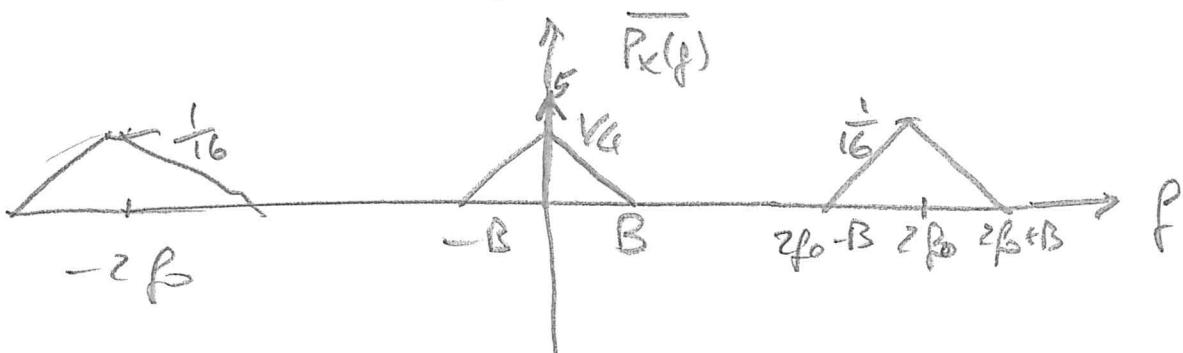
Process cyclostationary; methods will persist

$$\overline{R_X(\tau)} = \langle R_X(t, \tau) \rangle_{\mu} = \frac{1}{4} R_X(\tau) + \frac{1}{8} R_X(\tau) \cos 4\pi f_0 \tau + E[C^2]$$

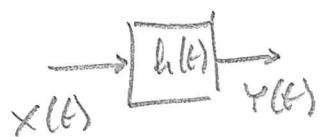
$$\overline{S_c^2} = 4 \quad S_c^2 = E[C^2] - \mu_c^2 \quad E[C^2] = \overline{s_c^2} + \mu_c^2 = \frac{4+1}{16} = \frac{5}{16} \Rightarrow$$

sol II p. - 3 N: 2015-2

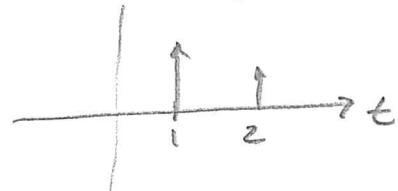
$$\overline{P_x(f)} = \frac{1}{4} P_x(f) + \frac{1}{16} P_x(f-2f_0) + \frac{1}{16} P_x(f+2f_0) + 5 \delta(f)$$



(2)



$$h(t) = \delta(t-1) + \frac{1}{2} \delta(t-2)$$

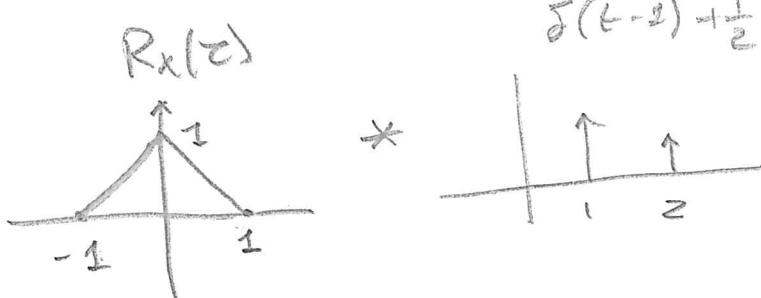


$$R_{yx}(z) = E[Y(t)Y(t-z)] =$$

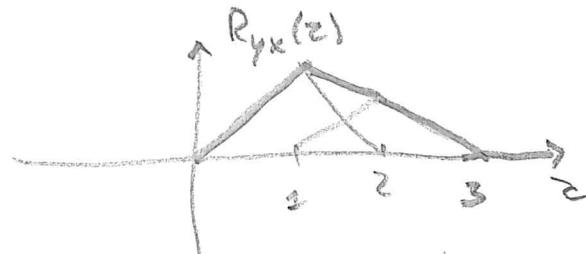
$$= E\left[\int_{-\infty}^{+\infty} h(\eta) x(t-\eta) d\eta \cdot x(t-z)\right] = \int_{-\infty}^{+\infty} h(\eta) E[x(t-\eta)x(t-z)] d\eta$$

$R_x(t-\eta-t+z)$

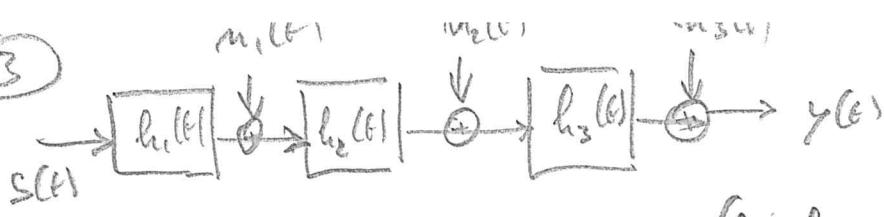
$$= \int_{-\infty}^{+\infty} h(\eta) R_x(z-\eta) d\eta = (h * R_x)(z)$$



$$= R_x(z-1) + \frac{1}{2} R_x(z-2)$$



(3)



Sol II pi - SDic 2015 - 3

$$y(t) = \underbrace{(h_1 * h_2 * h_3 * S)(t)}_{S(t)} + \underbrace{(h_3 * n_1)(t) + (h_3 * n_2)(t) + n_3(t)}_{n_R(t)}$$

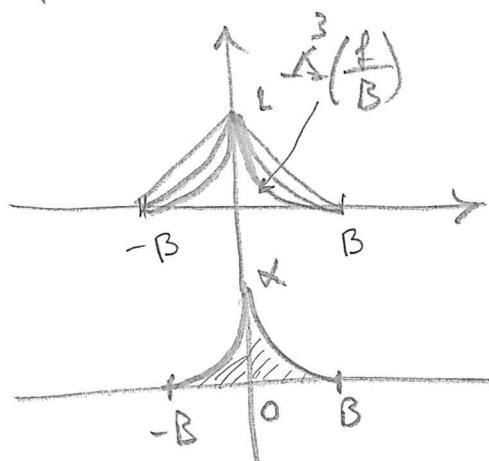
$$P_y(f) = \frac{[H_1(f)]^2 [H_2(f)]^2 [H_3(f)]^2 P_S(f)}{P_{SR}(f)} + \frac{[H_3(f)]^2 [H_2(f)]^2 P_{n_1}(f) + [H_3(f)]^2 P_{n_2}(f) + P_{n_3}(f)}{P_{n_R}(f)}$$

$$P_S(f) = \alpha \pi \left(\frac{f}{2B} \right)^3$$

$\alpha = 8 \cdot 10^{-4} \text{ mV}^2 / \text{Hz}$

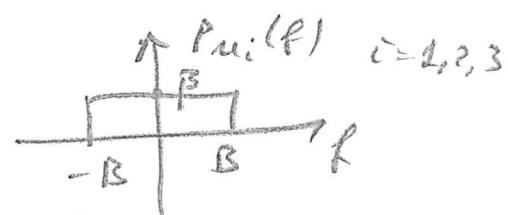
$$P_S = 2B \alpha = 2 \cdot 10000 \cdot 8 \cdot 10^{-4} \text{ mV}^2 = 16 \text{ mV}^2$$

$$P_{SR}(f) = \Delta^3 \left(\frac{f}{2B} \right) \alpha \pi \left(\frac{f}{2B} \right)$$

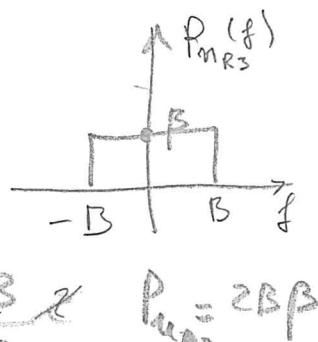
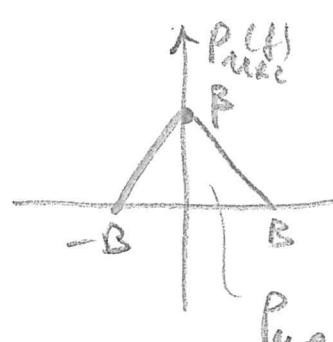
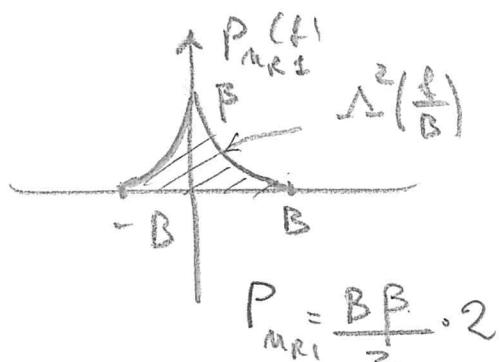


$$P_{SR} = \frac{B \alpha \cdot 2}{4 \pi} = \frac{10000 \cdot 8 \cdot 10^{-4}}{4 \pi} = 4 \text{ mV}^2$$

$$P_{n_1}(f) = P_{n_2}(f) = P_{n_3}(f) = \beta \pi \left(\frac{f}{2B} \right);$$



$$P_{n_i} = 2B\beta; \quad \beta = \frac{P_{n_i}}{2B} = \frac{9 \text{ mV}^2}{20000 \text{ Hz}} = 4.5 \cdot 10^{-4} \text{ mV}^2 / \text{Hz}$$



(3) cont.

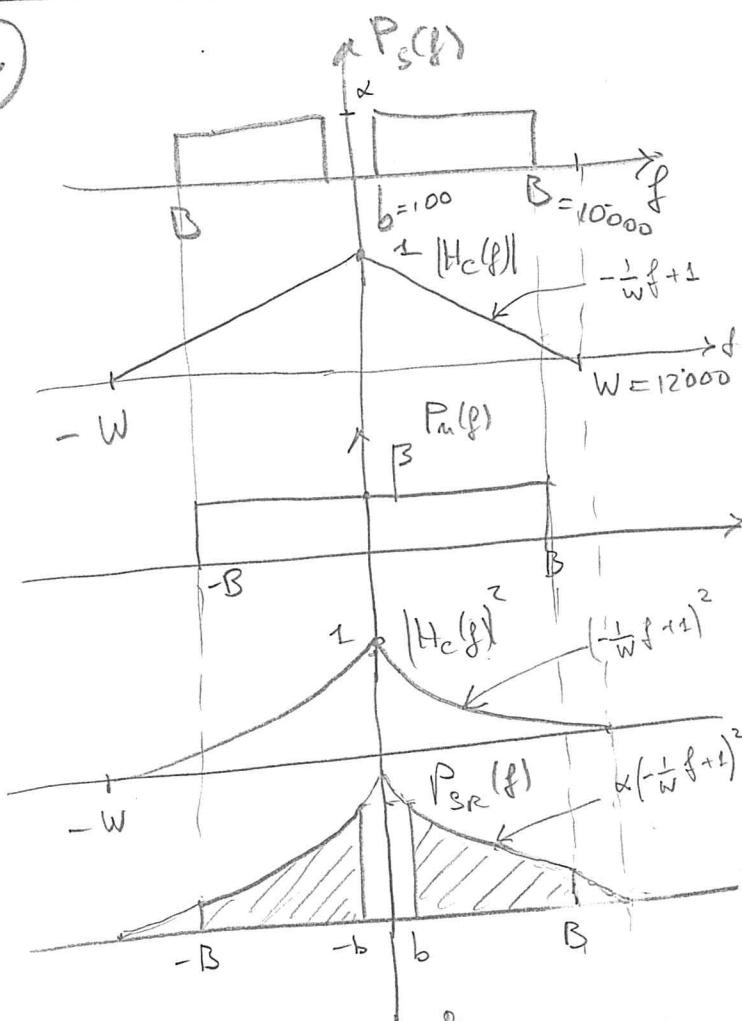
Sol II pi - 3 Dec 2015 - 4

$$\left(\frac{S}{N}\right) = \frac{P_{SR}}{P_{NR}} = \frac{\frac{B\alpha}{2}}{\frac{BP}{3} \cdot 2 + BP + 2BP} = \frac{\frac{1}{\beta}}{\frac{4}{3} + 2 + 4} = \frac{1}{\beta} \frac{1}{\frac{4}{3} + 2 + 4}$$

$$= \frac{1}{\beta} \frac{3}{4+6+12} = \frac{\alpha}{\beta} \frac{3}{22} = \frac{8 \cdot 10^{-4}}{4.5 \cdot 10^{-4}} \frac{3}{22} = 0.24$$

$$\left(\frac{S}{N}\right)_{dB} = 10 \log_{10} 0.24 = -6.15 \text{ dB}$$

(4)



$$P_s^{RMS} = 10 \mu V$$

$$P_s = 100 \mu V^2$$

$$P_s = (B - b)\alpha \cdot 2$$

$$\alpha = \frac{P_s}{2(B - b)} = \frac{100}{2 \cdot 9900}$$

$$= 5.1 \cdot 10^{-3} \frac{\mu V^2}{Hz}$$

$$P_m^{RMS} = 5 \mu V$$

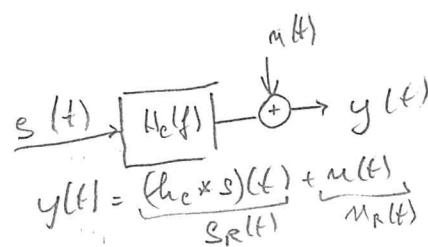
$$P_m = 25 \mu V^2$$

$$P_m = B \cdot \beta \cdot 2$$

$$\beta = \frac{P_m}{B \cdot 2} = \frac{25}{10000 \cdot 2}$$

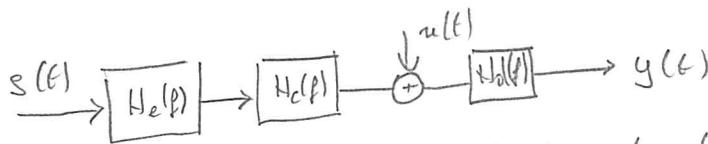
$$= 1.3 \cdot 10^{-3} \frac{\mu V^2}{Hz}$$

$$\begin{aligned} P_{SR} &= 2 \int_b^B \alpha \left(-\frac{1}{w} f + 1 \right)^2 df = 2 \alpha \int_b^B \left(\frac{1}{w^2} f^2 + 1 - \frac{2}{w} f \right) df \\ &= 2 \alpha \left[\frac{1}{w^2} \frac{f^3}{3} \Big|_b^B + (B - b) - \frac{2}{w} \frac{f^2}{2} \Big|_b^B \right] = \\ &= 2 \alpha \left[\frac{B^3 - b^3}{3w^2} + f(B - b) - \frac{B^2 - b^2}{w} \right] = \\ &= 2 \cdot 5.1 \cdot 10^{-3} \left[\frac{(10000)^3 - (100)^3}{3 \cdot (2000)^2} + 9900 - \frac{(10000)^2 - (100)^2}{12 \cdot 000} \right] = 10.2 \cdot 10^{-3} \left(2.32 \cdot 10^3 + 9900 - 8.33 \cdot 10^3 \right) \\ &= 33.60 \mu V^2 \end{aligned}$$



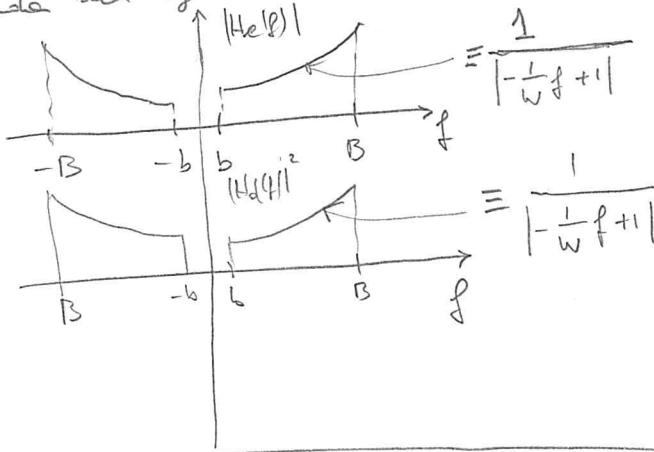
$$\left(\frac{\Sigma}{N}\right) = \frac{39.60}{25.00} = 1.58 \quad \left(\frac{\Sigma}{N}\right)_{dB} = 10 \lg 1.58 = 2.00 \text{ dB}$$

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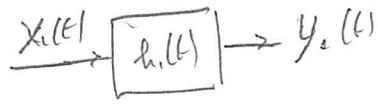


Limitations d'Heureux allo bende del segnale $[S, B]$

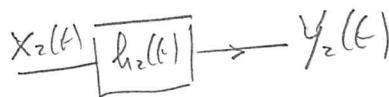
$$\begin{cases} |H_e(f)|^2 = \frac{\alpha K}{|H_e(f)|} \frac{P_{u^2}(f)}{P_{s^2}(f)} \text{ cost} \\ |H_d(f)|^2 = \frac{K}{\alpha |H_c(f)|} \frac{P_s^2(f)}{P_{u^2}(f)} \text{ cost.} \end{cases}$$



⑤



$$R_{y_1 y_2}(z) = (E_{h_1, h_2} * R_{x_1 x_2})(z)$$



$$R_{y_1 y_2}(z) = E[Y_1(t) Y_2^*(t-z)] = E \left[\int_{-\infty}^{+\infty} X_1(\eta) h_1(t-\eta) d\eta \int_{-\infty}^{+\infty} X_2^*(\xi) h_2^*(t-z-\xi) d\xi \right]$$

$$= \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} \underbrace{E[X_1(\eta) X_2^*(\xi)]}_{R_{x_1 x_2}(\eta-\xi)} h_1(t-\eta) h_2^*(t-z-\xi) d\eta d\xi$$

$$= \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} R_{x_1 x_2}(\eta-\xi) h_1(t-\eta) h_2^*(t-z-\xi) d\eta d\xi$$

$$\alpha = \eta - \xi$$

$$= \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} R_{x_1 x_2}(\alpha) h_1(t-\eta) h_2^*(t-z-\eta+\alpha) d\alpha d\eta$$

$$= \int_{-\infty}^{+\infty} R_{x_1 x_2}(\alpha) \int_{-\infty}^{-\infty} h_1(t-\eta) h_2^*(t-z-\eta+\alpha) d\eta d\alpha$$

$$= \int_{-\infty}^{+\infty} R_{x_1 x_2}(\alpha) \int_{-\infty}^{+\infty} h_1(\beta) h_2^*(\beta-z+\alpha) d\beta d\alpha$$

$$= \int_{-\infty}^{+\infty} R_{x_1 x_2}(\alpha) \mathcal{Z}_{h_1, h_2}(z-\alpha) d\alpha = (R_{x_1 x_2} * \mathcal{E}_{h_1, h_2})(z)$$